

**OBJECTIVES:**

- To gain conceptual and basic mathematical understanding of electric and magnetic fields in free space and in materials
- To understand the coupling between electric and magnetic fields through Faraday's law, displacement current and Maxwell's equations
- To understand wave propagation in lossless and in lossy media
- To be able to solve problems based on the above concepts

**UNIT I INTRODUCTION**

12

Electromagnetic model, Units and constants, Review of vector algebra, Rectangular, cylindrical and spherical coordinate systems, Line, surface and volume integrals, Gradient of a scalar field, Divergence of a vector field, Divergence theorem, Curl of a vector field, Stoke's theorem, Null identities, Helmholtz's theorem

**UNIT II ELECTROSTATICS**

12

Electric field, Coulomb's law, Gauss's law and applications, Electric potential, Conductors in static electric field, Dielectrics in static electric field, Electric flux density and dielectric constant, Boundary conditions, Capacitance, Parallel, cylindrical and spherical capacitors, Electrostatic energy, Poisson's and Laplace's equations, Uniqueness of electrostatic solutions, Current density and Ohm's law, Electromotive force and Kirchhoff's voltage law, Equation of continuity and Kirchhoff's current law

**UNIT III MAGNETOSTATICS**

12

Lorentz force equation, Law of no magnetic monopoles, Ampere's law, Vector magnetic potential, Biot-Savart law and applications, Magnetic field intensity and idea of relative permeability, Magnetic circuits, Behaviour of magnetic materials, Boundary conditions, Inductance and inductors, Magnetic energy, Magnetic forces and torques

**UNIT IV TIME-VARYING FIELDS AND MAXWELL'S EQUATIONS**

12

Faraday's law, Displacement current and Maxwell-Ampere law, Maxwell's equations, Potential functions, Electromagnetic boundary conditions, Wave equations and solutions, Time-harmonic fields

**UNIT V PLANE ELECTROMAGNETIC WAVES**

12

Plane waves in lossless media, Plane waves in lossy media (low-loss dielectrics and good conductors), Group velocity, Electromagnetic power flow and Poynting vector, Normal incidence at a plane conducting boundary, Normal incidence at a plane dielectric boundary

**TOTAL: 60 PERIODS****OUTCOMES:**

By the end of this course, the student should be able to:

- Display an understanding of fundamental electromagnetic laws and concepts
- Write Maxwell's equations in integral, differential and phasor forms and explain their physical meaning
- Explain electromagnetic wave propagation in lossy and in lossless media
- Solve simple problems requiring estimation of electric and magnetic field quantities based on these concepts and laws

**TEXT BOOKS:**

1. D.K. Cheng, Field and wave electromagnetics, 2nd ed., Pearson (India), 1989 (UNIT I, II, III, IV, V)
2. W.H. Hayt and J.A. Buck, Engineering electromagnetics, 7th ed., McGraw-Hill (India), 2006 (UNIT I-V)

**REFERENCES**

1. D.J. Griffiths, Introduction to electrodynamics, 4th ed., Pearson (India), 2013
2. B.M. Notaros, Electromagnetics, Pearson: New Jersey, 2011
3. M.N.O. Sadiku and S.V. Kulkarni, Principles of electromagnetics, 6th ed., Oxford (Asian Edition), 2015

## UNIT-I Introduction

Electromagnetic Model, units and constants,  
Review of vector Algebra, Rectangular, cylindrical  
and spherical coordinate systems, Line, Surface  
and volume integrals. Gradient of a scalar field,  
Divergence of vector field, Divergence theorem,  
Curl of a vector field, Stoke's theorem, Null  
Identities, Helmholtz's theorem.

## ⇒ Introduction

Electromagnetics is the study of behaviour of <sup>electric</sup> charges in ~~the free space~~ rest (or) motion

The effect of charges at rest position is called as electrostatic field (or) electric field ('E').

The effect of charges at motion is called as magnetic field. Generally the moving charges produce a current which gives rise to a magnetic field ('H').

A field is a spatial distribution of any physical quantity, which may (or) may not be a function of time.

## ⇒ Electromagnetic Model

It is a mathematical representation of an electromagnetic field.

## ⇒ units and constants

Fundamental units

Quantity	Unit
Length	metres (m)
Mass	Kilogram (Kg)
Time	second (s)
Current	ampere (A)

## Constants

Universal constants	Symbol	Value
Velocity of light	$c$	$3 \times 10^8 \text{ m/s}$
permeability of free space	$\mu_0$	$4\pi \times 10^{-7} \text{ H/m}$
permittivity of free space	$\epsilon_0$	$8.854 \times 10^{-12} \text{ F/m}$

## ⇒ Review of Vector Algebra

### Scalar Quantity

It is a quantity which contains only magnitude.

Ex: - voltage, current, temp etc.

### Vector Quantity

It is a quantity which contains both magnitude and direction.

Ex - force, displacement & acceleration.

$$\vec{F} = m\vec{A}$$

Vector Algebra - Addition, subtraction & multiplication of vectors.

### 1) Addition of vectors

$\vec{A}$  is a vector &  $\vec{B}$  is another vector

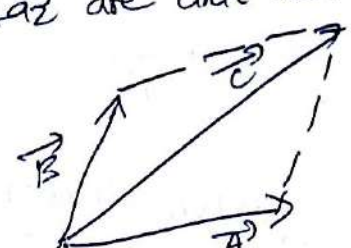
$$\text{In General } \vec{A} = A_x \vec{a}_x + A_y \vec{a}_y + A_z \vec{a}_z$$

$$\vec{B} = B_x \vec{a}_x + B_y \vec{a}_y + B_z \vec{a}_z$$

Where  $A_x, A_y$  &  $A_z$  are magnitudes of vectors  
 $\vec{a}_x, \vec{a}_y$  &  $\vec{a}_z$  are unit vectors.



(a) Two vectors



b) parallelogram rule.



c) Head to tail rule

$$\vec{C} = \vec{A} + \vec{B}$$

Vector Addition obeys commutative & Associative laws.

Commutative law:  $\vec{A} + \vec{B} = \vec{B} + \vec{A}$

Associative law:  $\vec{A} + (\vec{B} + \vec{C}) = (\vec{A} + \vec{B}) + \vec{C}$

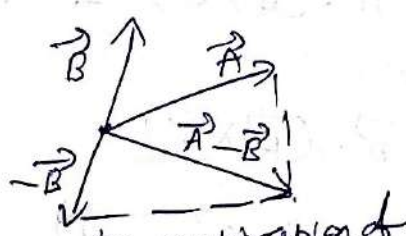
ii) Subtraction of vectors.

$$\vec{A} - \vec{B} = \vec{A} + (-\vec{B})$$

$\vec{A}$  &  $\vec{B}$  are two different vectors.



(a) Two vectors



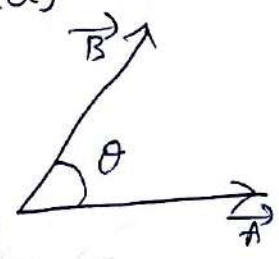
b) subtraction of vectors

iii) Multiplication of vectors

a) Dot product (scalar product)

$$\vec{A} \cdot \vec{B} = |\vec{A}| |\vec{B}| \cos \theta$$

Note:  $\vec{A} \cdot \vec{B} = \vec{B} \cdot \vec{A}$



b) Cross product (vector product)

$$\vec{A} \times \vec{B} = |\vec{A}| |\vec{B}| \sin \theta$$

$$\vec{A} \times \vec{B} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ A_x & A_y & A_z \\ B_x & B_y & B_z \end{vmatrix}$$

Note, -

$$\vec{A} \times \vec{B} = -\vec{B} \times \vec{A}$$

### parallel and perpendicular vectors

a) if  $\vec{A} \times \vec{B} = 0$  then the two vectors are said to be parallel.

b) if  $\vec{A} \cdot \vec{B} = 0$  then the two vectors are said to be perpendicular.

problem

$$\textcircled{1} \quad \vec{A} = \vec{a}_x + \vec{a}_y$$

$$\vec{B} = \vec{a}_x + 2\vec{a}_z$$

$$\vec{C} = 2\vec{a}_y + \vec{a}_z$$

Find  $\vec{A} \cdot (\vec{B} \times \vec{C})$  &  $\vec{A} + \vec{B} + \vec{C}$

Sol:-

i)  $\vec{A} \cdot (\vec{B} \times \vec{C})$

$$\vec{B} \times \vec{C} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ 1 & 0 & 2 \\ 0 & 2 & 1 \end{vmatrix}$$

$$= \vec{a}_x (0 - 4) - \vec{a}_y (1 - 0) + \vec{a}_z (2 - 0)$$

$$\vec{B} \times \vec{C} = -4\vec{a}_x - \vec{a}_y + 2\vec{a}_z$$

$$\vec{A} \cdot (\vec{B} \times \vec{C}) = (\vec{a}_x + \vec{a}_y) \cdot (-4\vec{a}_x - \vec{a}_y + 2\vec{a}_z)$$

$$= -4 - 1$$

$$\begin{aligned} \therefore \vec{a}_x \cdot \vec{a}_x &= 1 \\ \vec{a}_y \cdot \vec{a}_z &= 0 \end{aligned}$$

$$\boxed{\vec{A} \cdot (\vec{B} \times \vec{C}) = -5}$$

$$\begin{aligned}
 \text{ii) } \vec{A} + \vec{B} + \vec{C} &= (\vec{a}_x + \vec{a}_y) + (\vec{a}_x + 2\vec{a}_z) + (2\vec{a}_y + \vec{a}_z) \\
 &= 2\vec{a}_x + 3\vec{a}_y + 3\vec{a}_z.
 \end{aligned}$$

$$\textcircled{2} \text{ Given } \vec{A} = 4\vec{a}_x - 2\vec{a}_y + 2\vec{a}_z$$

$$\vec{B} = -6\vec{a}_x + 3\vec{a}_y - 3\vec{a}_z$$

find whether the two vectors are parallel or perpendicular.

i) parallel condition.

$$\vec{A} \times \vec{B} = 0$$

$$\vec{A} \times \vec{B} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ 4 & -2 & 2 \\ -6 & 3 & -3 \end{vmatrix}$$

$$= \vec{a}_x (6-6) - \vec{a}_y (-12+12) + \vec{a}_z (12-12)$$

$\vec{A} \times \vec{B} = 0$ . Therefore two vectors are parallel.

ii) perpendicular condition.

$$\vec{A} \cdot \vec{B} = 0$$

$$\vec{A} \cdot \vec{B} = (4\vec{a}_x - 2\vec{a}_y + 2\vec{a}_z) \cdot (-6\vec{a}_x + 3\vec{a}_y - 3\vec{a}_z)$$

$$= -24 - 6 - 6$$

$$\vec{A} \cdot \vec{B} = -36 \neq 0$$

Therefore two vectors are not perpendicular.

⇒ Vector Calculations.

There are 4 types

i) Differential operator (or) vector operator

ii) Gradient of a scalar field

iii) Divergence of a vector field

iv) curl of a vector field.

i) Differential operator (or) vector operator

$$\nabla = \frac{\partial}{\partial x} \vec{a}_x + \frac{\partial}{\partial y} \vec{a}_y + \frac{\partial}{\partial z} \vec{a}_z$$

ii) Gradient

Gradient of a scalar field produces vector field.

$$\nabla A = \frac{\partial A}{\partial x} \vec{a}_x + \frac{\partial A}{\partial y} \vec{a}_y + \frac{\partial A}{\partial z} \vec{a}_z$$

Gradient of a scalar is a vector.

iii) Divergence of a vector field.

$$\nabla \cdot \vec{A} = \left( \frac{\partial}{\partial x} \vec{a}_x + \frac{\partial}{\partial y} \vec{a}_y + \frac{\partial}{\partial z} \vec{a}_z \right) \cdot$$

$$(A_x \vec{a}_x + A_y \vec{a}_y + A_z \vec{a}_z)$$

$$\nabla \cdot \vec{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$$

Divergence of a vector is a scalar.



iv) curl of a vector field.

Curl deals with rotation

$$\nabla \times \vec{A} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix}$$

Note:-

if  $\nabla \cdot \vec{A} = 0$  then  $\vec{A}$  is solenoidal.

if  $\nabla \times \vec{A} = 0$  then  $\vec{A}$  is irrotational.

problems

③ Given  $\vec{A} = 3y^4z^2\vec{a}_x + 4x^3z^2\vec{a}_y + 3z^2y^2\vec{a}_z$   
check whether given vector is solenoidal.

Sol:-

$$\nabla \cdot \vec{A} = \left( \frac{\partial}{\partial x} 3y^4z^2 \right) + \left( \frac{\partial}{\partial y} 4x^3z^2 \right) + \left( \frac{\partial}{\partial z} 3z^2y^2 \right)$$

$$\nabla \cdot \vec{A} = 0 + 0 + 6y^2z$$

$$\nabla \cdot \vec{A} = 6y^2z \neq 0$$

so  $\vec{A}$  is not solenoidal.

④ Check whether given vector is irrotational  
 $\vec{A} = 2xy\vec{a}_x + (x^2 + 2yz)\vec{a}_y + (y^2 + 1)\vec{a}_z$

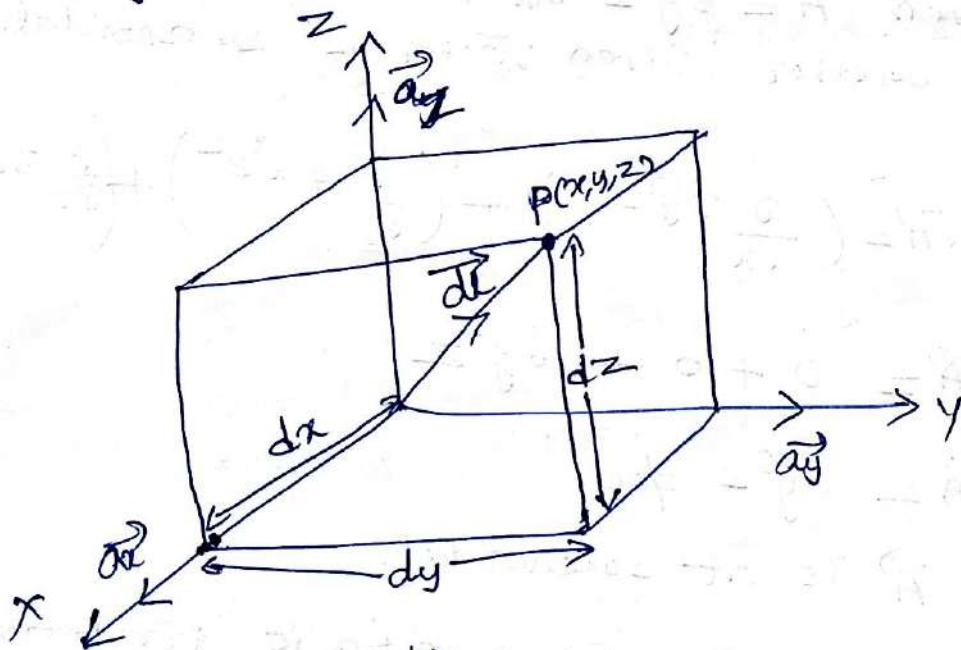
$$\nabla \times \vec{A} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ 2xy & (x^2 + 2yz) & (y^2 + 1) \end{vmatrix}$$

$$\begin{aligned}
 &= \vec{a}_x \left[ \frac{\partial (y^2 + 1)}{\partial y} - \frac{\partial (x^2 + 2yz)}{\partial z} \right] \\
 &\quad - \vec{a}_y \left[ \frac{\partial (y^2 + 1)}{\partial x} - \frac{\partial (2xy)}{\partial z} \right] + \vec{a}_z \left[ \frac{\partial (x^2 + 2yz)}{\partial x} - \frac{\partial (2xy)}{\partial y} \right] \\
 &= \vec{a}_x [2y - 2y] - \vec{a}_y [0 - 0] + \vec{a}_z [2x - 2x]
 \end{aligned}$$

$$= 0$$

$\nabla \times \vec{A} = 0 \quad \therefore$  The given  $\vec{A}$  is irrotational.

(or) Cartesian.  
 $\Rightarrow$  Rectangular Coordinate System.



$\rightarrow$  Differential ~~line~~ elements  
 $dx, dy$  &  $dz$

$\rightarrow$  Differential Surface element  
 along x direction  $dS_x = dy dz \vec{a}_x$   
 along y direction  $dS_y = dx dz \vec{a}_y$   
 along z direction  $dS_z = dx dy \vec{a}_z$

→ Differential volume

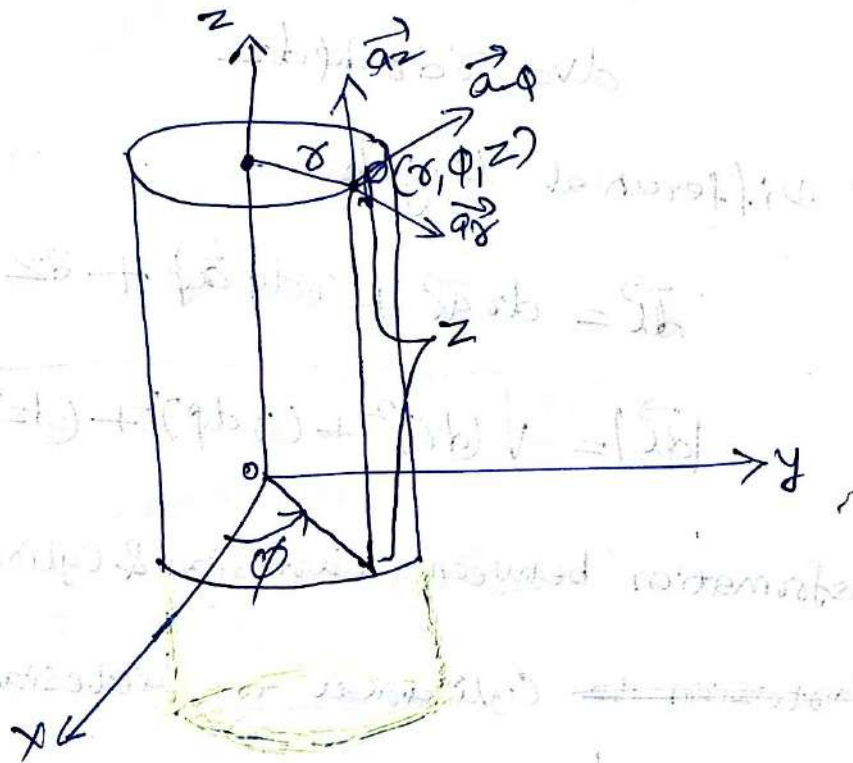
$$dv = dx dy dz$$

→ Differential length

$$d\vec{l} = dx \vec{a}_x + dy \vec{a}_y + dz \vec{a}_z$$

$$|d\vec{l}| = \sqrt{(dx)^2 + (dy)^2 + (dz)^2}$$

⇒ Cylindrical coordinate system.



The ranges of the variables are

$$0 \leq r \leq \infty$$

$$0 \leq \phi \leq 2\pi$$

$$-\infty \leq z \leq \infty$$

where  $r$  - radius of the cylinder with  $z$  axis  
 $\phi$  - angle of the plane w.r.t  $xz$  plane  
 $z$  - height of the plane from origin.

$$\phi = \cos^{-1} \frac{z}{r}$$

→ differential ~~line~~ <sup>line</sup> element  $dr, r d\phi$  &  $dz$

→ Differential surface element

along x direction  $\vec{ds}_r = r d\phi dz \vec{a}_r$

along y direction  $\vec{ds}_\phi = dr dz \vec{a}_\phi$

along z direction  $\vec{ds}_z = r dr d\phi \vec{a}_z$

→ Differential volume

$$dv = r dr d\phi dz$$

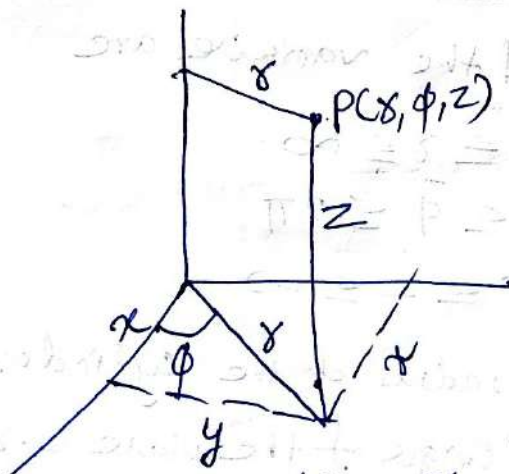
→ Differential length

$$d\vec{l} = dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z$$

$$|d\vec{l}| = \sqrt{(dr)^2 + (r d\phi)^2 + (dz)^2}$$

⇒ Transformation between Cartesian & Cylindrical system

i) ~~Cartesian to~~ Cylindrical to Cartesian



$$\cos \phi = \frac{\text{adj}}{\text{hyp}} = \frac{x}{r}$$

$$\boxed{x = r \cos \phi}$$

$$\sin \phi = \frac{\text{opp}}{\text{Hyp}} = \frac{y}{r}$$

$$\boxed{y = r \sin \phi}$$

$$\boxed{z = z}$$

ii) Cartesian to cylindrical

$$x = r \cos \phi \Rightarrow x^2 = r^2 \cos^2 \phi$$

$$y = r \sin \phi \Rightarrow y^2 = r^2 \sin^2 \phi$$

$$x^2 + y^2 = r^2 [\cos^2 \phi + \sin^2 \phi]$$

$$\therefore r^2 = x^2 + y^2$$

$$\boxed{r = \sqrt{x^2 + y^2}}$$

$$\frac{y}{x} = \frac{r \sin \phi}{r \cos \phi} = \tan \phi$$

$$\therefore \tan \phi = y/x$$

$$\boxed{\phi = \tan^{-1}(y/x)}$$

$$\boxed{z = z}$$

### problems

⑤ At a point  $P(3, 90^\circ, 15)$ . Convert into Cartesian

Soln-  $r = 3$   $\phi = 90^\circ$   $z = 15$

$$x = r \cos \phi = 3 \cos 90^\circ = 0$$

$$y = r \sin \phi = 3 \sin 90^\circ = 3$$

$$z = 15$$

In Cartesian  $P(0, 3, 15)$ .

⑥ At a point  $A(x=2, y=3, z=-1)$  convert into cylindrical.

Sol:-

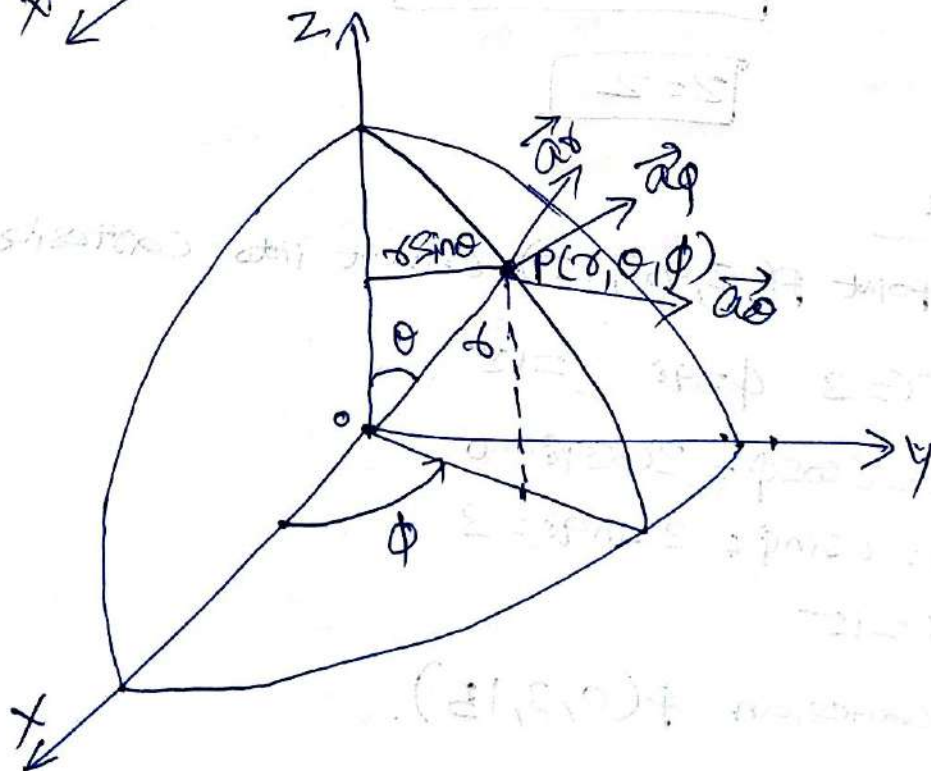
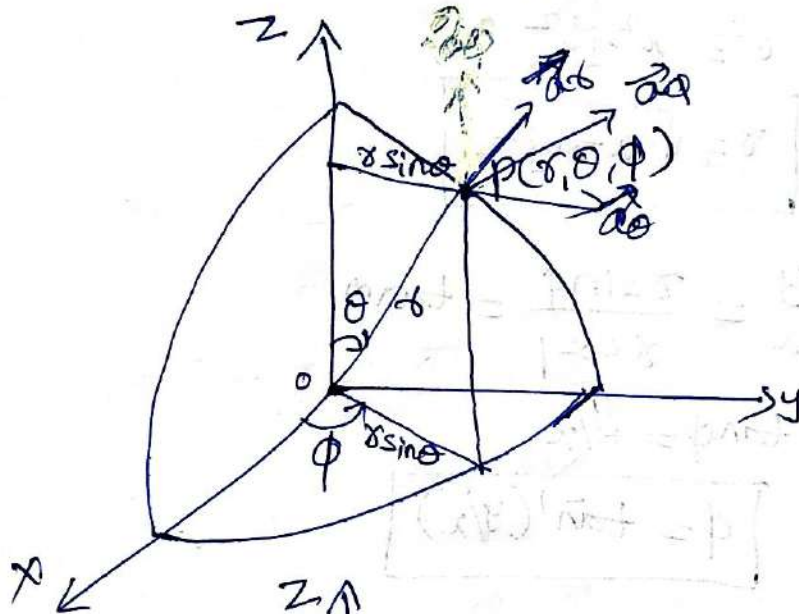
$$r = \sqrt{x^2 + y^2} = \sqrt{4 + 9} = \sqrt{13} = 3.6$$

$$\phi = \tan^{-1}(y/x) = \tan^{-1}(3/2) = 56.3^\circ$$

$$z = -1$$

In cylindrical  $A(r=3.6, \phi=56.3^\circ, z=-1)$

⇒ Spherical coordinate system.



The ranges of the variables are:

$$0 \leq r \leq \infty$$

$$0 \leq \theta \leq \pi$$

$$0 \leq \phi \leq 2\pi$$

where  $r$  - radius of the sphere.

$\theta$  - Half angle of the right circular cone w.r.t z axis

$\phi$  - angle of the plane w.r.t xz plane.

→ Differential line element

$$dr, r d\theta, r \sin\theta d\phi$$

→ Differential surface element

along  $r$  direction  $\vec{ds}_r = r^2 \sin\theta d\theta d\phi \vec{a}_r$

along  $\theta$  direction  $\vec{ds}_\theta = r \sin\theta dr d\phi \vec{a}_\theta$

along  $\phi$  direction  $\vec{ds}_\phi = r dr d\theta \vec{a}_\phi$

→ Differential volume

$$dv = r^2 \sin\theta dr d\theta d\phi$$

→ Differential length

$$\vec{dl} = dr \vec{a}_r + r d\theta \vec{a}_\theta + r \sin\theta d\phi \vec{a}_\phi$$

$$|\vec{dl}| = \sqrt{(dr)^2 + (r d\theta)^2 + (r \sin\theta d\phi)^2}$$

⇒ Transformation between Cartesian & spherical system.

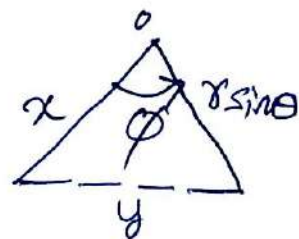
1) Spherical to Cartesian.

$$\cos\phi = \frac{x}{r \sin\theta}$$

$$\Rightarrow x = r \sin\theta \cos\phi$$

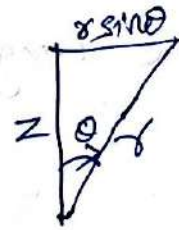
$$\sin\phi = \frac{y}{r \sin\theta}$$

$$\Rightarrow y = r \sin\theta \sin\phi$$



$$\cos \theta = \frac{z}{r}$$

$$\Rightarrow z = r \cos \theta$$



ii) Cartesian to spherical

$$x^2 + y^2 + z^2 = r^2 \sin^2 \theta \cos^2 \phi + r^2 \sin^2 \theta \sin^2 \phi + r^2 \cos^2 \theta$$

$$\Rightarrow r^2 = r^2 \sin^2 \theta [\cos^2 \phi + \sin^2 \phi] + r^2 \cos^2 \theta$$

$$= r^2 [\sin^2 \theta + \cos^2 \theta]$$

$$\Rightarrow r^2 = x^2 + y^2 + z^2$$

$$r = \sqrt{x^2 + y^2 + z^2}$$

$$z = r \cos \theta$$

$$\cos \theta = z/r$$

$$\theta = \cos^{-1}(z/r)$$

$$\frac{y}{x} = \frac{r \sin \theta \sin \phi}{r \sin \theta \cos \phi}$$

$$\frac{y}{x} = \tan \phi$$

$$\phi = \tan^{-1}(y/x)$$





## problems

⑦ A point  $P(6, 4, 2)$  into spherical

$$x=6 \quad y=4 \quad z=2$$

$$r = \sqrt{x^2 + y^2 + z^2} = \sqrt{36 + 16 + 4} = \sqrt{56}$$

$$r = 7.48$$

$$\theta = \cos^{-1}\left(\frac{z}{r}\right)$$

$$\theta = 74.3^\circ$$

$$\phi = \tan^{-1}\left(\frac{y}{x}\right)$$

$$\phi = 33.69^\circ$$

⑧ A point  $P(8, 110^\circ, 60^\circ)$  into Cartesian.

$$x = r \sin\theta \cos\phi = 8 \sin(110^\circ) \cos(60^\circ)$$

$$= 8 \times 0.939 \times 0.5$$

$$x = 3.75$$

$$y = r \sin\theta \sin\phi = 8 \sin(110^\circ) \sin(60^\circ)$$

$$= 8 \times 0.939 \times 0.86$$

$$y = 6.50$$

$$z = r \cos\theta = 8 \cos(110^\circ) = 8 \cos(110^\circ)$$

$$= 8 \times -0.342$$

$$z = -2.73$$

⇒ Divergence Theorem.

The surface integral of the normal component of the vector over the closed surface is equal to the volume integral of a divergence of the vector throughout the volume.

$$\oiint \vec{A} \cdot d\vec{s} = \iiint \nabla \cdot \vec{A} \, dv.$$

Proof

R.H.S

$$\iiint \nabla \cdot \vec{A} \, dv = \iiint \left( \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z} \right) dx dy dz$$

∴ In cartesian coordinate system,

$$\nabla \cdot \vec{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$$

$$dv = dx dy dz$$

$$\iiint \nabla \cdot \vec{A} \, dv = \iiint \frac{\partial A_x}{\partial x} dx dy dz + \iiint \frac{\partial A_y}{\partial y} dx dy dz$$

$$+ \iiint \frac{\partial A_z}{\partial z} dx dy dz$$

$$= \iint A_x dy dz + \iint A_y dx dz + \iint A_z dx dy$$

w.k.T  $\vec{A} = A_x \vec{a}_x + A_y \vec{a}_y + A_z \vec{a}_z$   
 $d\vec{s} = ds_x \vec{a}_x + ds_y \vec{a}_y + ds_z \vec{a}_z$

$$= \iint A_x ds_x + \iint A_y ds_y + \iint A_z ds_z$$

$$= \iint [A_x ds_x + A_y ds_y + A_z ds_z]$$

$$= \oiint \vec{A} \cdot d\vec{s}$$

$\therefore$  R.H.S = L.H.S

Have proved.

Note:-

~~Some~~ Divergence theorem also stated as  $\oiint \vec{A} \cdot n ds = \iiint \nabla \cdot \vec{A} dv$

⇒ Divergence of the vector

→ Cartesian coordinate system

$$\nabla \cdot \vec{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}$$

→ cylindrical coordinate system

$$\nabla \cdot \vec{A} = \frac{1}{r} \frac{\partial (r A_r)}{\partial r} + \frac{1}{r} \frac{\partial A_\phi}{\partial \phi} + \frac{\partial A_z}{\partial z}$$

→ spherical coordinate system

$$\nabla \cdot \vec{A} = \frac{1}{r^2} \frac{\partial (r^2 A_r)}{\partial r} + \frac{1}{r \sin \theta} \frac{\partial (A_\theta \sin \theta)}{\partial \theta} + \frac{1}{r \sin \theta} \frac{\partial A_\phi}{\partial \phi}$$

problem

evaluate  $\int_S \vec{A} \cdot d\vec{s}$  where

- ⑨ using divergence theorem,  $\vec{A} = 2xy \vec{a}_x + y^2 \vec{a}_y + 4yz \vec{a}_z$   
and  $S$  is the surface of the <sup>Cube</sup> bounded by  
 $x=0, x=1, y=0, y=1$  &  $z=0, z=1$ .

Sol:-

By divergence theorem

$$\oint_S \vec{A} \cdot d\vec{s} = \iiint_V \nabla \cdot \vec{A} \, dV$$

$$\nabla \cdot \vec{A} = \frac{\partial Ax}{\partial x} + \frac{\partial Ay}{\partial y} + \frac{\partial Az}{\partial z}$$

$$= \frac{\partial(2xy)}{\partial x} + \frac{\partial(y^2)}{\partial y} + \frac{\partial(4yz)}{\partial z}$$

$$= 2y + 2y + 4y$$

$$\nabla \cdot \vec{A} = 8y$$

$$\therefore \iiint_V \nabla \cdot \vec{A} \, dV = \int_0^1 \int_0^1 \int_0^1 8y \, dx \, dy \, dz$$

$$= \int_0^1 \int_0^1 [8yx]_0^1 \, dy \, dz$$

$$= \int_0^1 \int_0^1 8y \, dy \, dz$$

$$= \int_0^1 \left[ \frac{8y^2}{2} \right]_0^1 \, dz$$

$$= \int_0^1 4 \, dz = 4[z]_0^1 = 4$$

$$\therefore \oint_S \vec{A} \cdot d\vec{s} = 4$$

10) verify divergence theorem, consider the field vector  $\vec{D} = 2xy \vec{a}_x$  and rectangular cube formed by a plane  $x=0, x=1, y=0, y=2$  &  $z=0, z=3$ .

Sol: - Divergence theorem

$$\oiint \vec{D} \cdot d\vec{s} = \iiint \nabla \cdot \vec{D} \, dV$$

R.H.S

$$\nabla \cdot \vec{D} = \frac{\partial D_x}{\partial x} + \frac{\partial D_y}{\partial y} + \frac{\partial D_z}{\partial z}$$

$$= \frac{\partial(2xy)}{\partial x} + \frac{\partial(0)}{\partial y} + \frac{\partial(0)}{\partial z}$$

$$\boxed{\nabla \cdot \vec{D} = 2y}$$

$$\iiint \nabla \cdot \vec{D} \, dV = \int_0^3 \int_0^2 \int_0^1 2y \, dx \, dy \, dz$$

$$= \int_0^3 dx \int_0^2 2y \, dy \int_0^1 dz$$

$$= [x]_0^1 \left[ \frac{2y^2}{2} \right]_0^2 [z]_0^1$$

$$= [1][4][1]$$

$$\boxed{\iiint \nabla \cdot \vec{D} \, dV = 4}$$

L.H.S

$$\oiint \vec{D} \cdot d\vec{s} = \iint D_x \, ds_x + \iint D_y \, ds_y + \iint D_z \, ds_z$$

$$= \int_0^3 \int_0^2 2xy \, dy \, dz + \int_0^3 \int_0^1 x^2 \, dx \, dz$$

$$= 2x \int_0^2 dy \int_0^3 dz + \int_0^1 x^2 dx \int_0^3 dz$$

$$= 2x [y]_0^2 [z]_0^3 + \left[ \frac{x^3}{3} \right]_0^1 [z]_0^3$$

$$= 2x [2][3] + \left[ \frac{1}{3} \right] [3]$$

$$= 12x$$

At  $x=1$

$$\oiint \vec{D} \cdot d\vec{s} = 12$$

(11) Given that  $\vec{D} = \frac{5r^2}{4} \vec{a}_r$  C/m<sup>2</sup>. Evaluate both the sides of divergence theorem for the volume enclosed by  $r=4m$  &  $\theta = \pi/4$ .

Sol:-

$r=4m, \theta = \pi/4$  & let us consider  $\phi = 0$  to  $2\pi$

Divergence theorem

$$\oiint \vec{D} \cdot d\vec{s} = \iiint \nabla \cdot \vec{D} \, dV$$

In spherical coordinate system

$$P(r, \theta, \phi)$$

$$d\vec{s} = r^2 \sin \theta \, d\theta \, d\phi \, \vec{a}_r$$

$$dV = r^2 \sin \theta \, dr \, d\theta \, d\phi$$

R.H.S

$$\nabla \cdot \vec{D} = \frac{1}{r^2} \frac{\partial (r^2 D_r)}{\partial r} \quad \because D_\theta \text{ present in } \vec{D}$$

$$= \frac{1}{r^2} \frac{\partial (r^2 \cdot \frac{5r^2}{4})}{\partial r}$$

$$= \frac{5}{4r^2} \frac{\partial (r^4)}{\partial r} = \frac{5}{4r^2} [4r^3]$$

$$\boxed{\nabla \cdot \vec{D} = 5r}$$

$$\iiint \nabla \cdot \vec{D} \, dv = \iiint 5r (r^2 \sin \theta \, dr \, d\theta \, d\phi)$$

$$= \int_0^{2\pi} \int_0^{\pi/4} \int_0^4 5r^3 \sin \theta \, dr \, d\theta \, d\phi$$

$$= \int_0^4 5r^3 \, dr \int_0^{\pi/4} \sin \theta \, d\theta \int_0^{2\pi} d\phi$$

$$= 5 \left[ \frac{r^4}{4} \right]_0^4 \left[ -\cos \theta \right]_0^{\pi/4} \left[ \phi \right]_0^{2\pi}$$

$$= 5 [64] \left[ 1 - \frac{1}{\sqrt{2}} \right] 2\pi$$

$$= 588.8$$

L.H.S

$$\begin{aligned} \iint \vec{D} \cdot d\vec{S} &= \iint D_r \, ds_r + \iint D_\theta \, ds_\theta + \iint D_\phi \, ds_\phi \\ &= \int_0^{2\pi} \int_0^{\pi/4} \frac{5r^2}{4} \cdot r^2 \sin \theta \, d\theta \, d\phi \end{aligned}$$

$$= \frac{5 \times 10^{-9}}{4} \int_0^{2\pi} \int_0^{\pi/4} 8 \sin \theta \, d\theta \, d\phi$$

$$= \frac{5 \times 10^{-9}}{4} \int_0^{2\pi} 8 \sin \theta \, d\theta \int_0^{\pi/4} d\phi$$

$$= \frac{5 \times 10^{-9}}{4} [-\cos \theta]_0^{\pi/4} (\phi)_0^{2\pi}$$

$$= \frac{5 \times 10^{-9}}{4} \left[ -\frac{1}{\sqrt{2}} + 1 \right] [2\pi]$$

$$\iint \vec{D} \cdot d\vec{S} = \frac{5 \times 10^{-9}}{4} \left[ 1 - \frac{1}{\sqrt{2}} \right] [2\pi]$$

at  $r=4$

$$\iint \vec{D} \cdot d\vec{S} = 5 [64] \left[ 1 - \frac{1}{\sqrt{2}} \right] [2\pi]$$

$$= 588.8$$

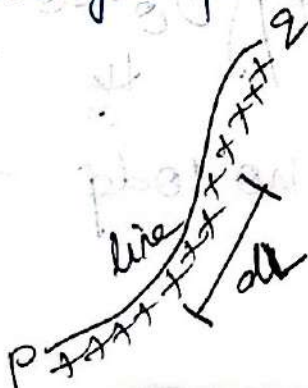
⇒ Types of Integral related to EM Theory.

i) Line Integral

$\rho_L$  is defined as total number of charges presented throughout entire length  $L$ .

Line charge density ( $\rho_L$ )

$$\rho_L = \frac{Q}{L} \text{ coulomb/meter}$$





Consider a small element

$$e_L = \frac{dq}{dL}$$

$$dq = e_L dL$$

$$Q = \int e_L dL$$

## Gauss law

The law states that the electric flux passing through any closed surface is equal to the charge enclosed by the surface.

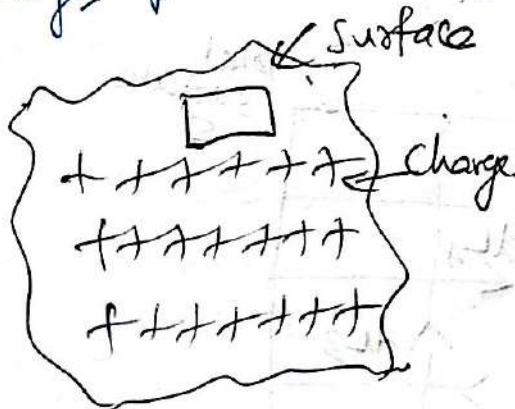
$$\Psi = Q$$

~~using Gauss law~~

~~$\Psi = Q$~~

ii) Surface Integral

$\rho_s$  is defined as total number of charges present throughout the entire surface(s)



Surface Charge Density ( $\rho_s$ )

$$\rho_s = \frac{Q}{S} \text{ Coulomb/m}^2$$

Consider a small element

$$\rho_s = \frac{dq}{ds}$$

$$dq = \rho_s ds$$

$$Q = \iint \rho_s ds$$

(ii) volume Integral

$\rho_v$  is defined as total number of charges present throughout entire volume.

$$\rho_v = \frac{Q}{V} \frac{\text{Coulomb}}{\text{m}^3}$$

Consider small element

$$\rho_v = \frac{dQ}{dV}$$

$$dQ = \rho_v dV$$

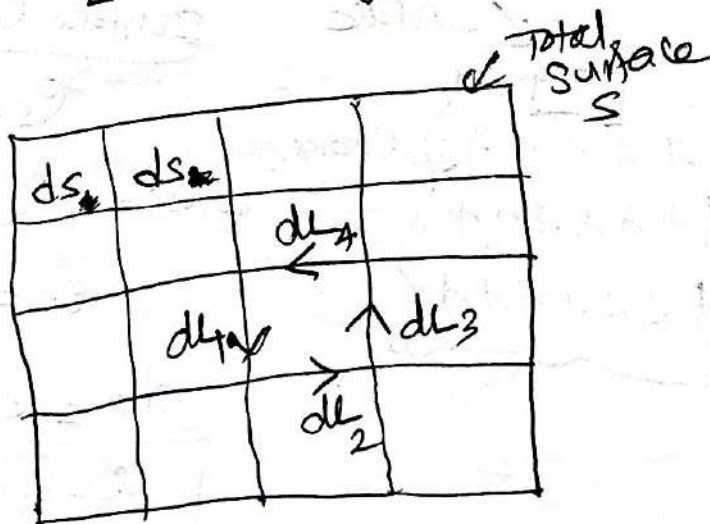
$$Q = \iiint \rho_v dV$$

⇒ Stokes's Theorem.

The line integral of any vector around a closed path  $L$  is equal to the surface integral of the curl of the vector over a open surface enclosed by the closed path  $L$ .

$$\oint_L \vec{F} \cdot d\vec{l} = \iint_S (\nabla \times \vec{F}) \cdot d\vec{S}$$

proof



$$\oint \vec{F} \cdot d\vec{L}_1 + \oint \vec{F} \cdot d\vec{L}_2 + \oint \vec{F} \cdot d\vec{L}_3 + \oint \vec{F} \cdot d\vec{L}_4$$

$$= \iint (\nabla \times \vec{F}) \cdot d\vec{s}_1 + \iint (\nabla \times \vec{F}) \cdot d\vec{s}_2$$

$$+ \iint (\nabla \times \vec{F}) \cdot d\vec{s}_3$$

According to ~~the~~ definition, Amount of rotation is done by dividing ~~the~~ total surface into small element surface area  $ds_1, ds_2, ds_3$ . Here the vector  $\vec{F}$  is moving on a closed path has shown in fig which forms as curl.

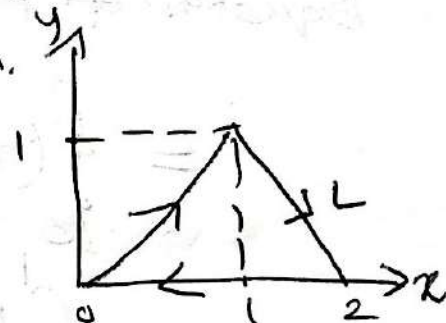
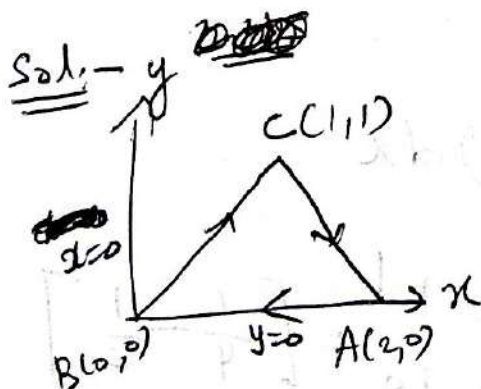
→ formula.

Combining the entire sequences

$$\boxed{\oint \vec{F} \cdot d\vec{L} = \iint (\nabla \times \vec{F}) \cdot d\vec{s}}$$

Problem

- (12) Given that  $\vec{F} = x^2y \vec{a}_x - y \vec{a}_y$   
verify Stokes's theorem.



L. 14.5

$$\int \vec{F} \cdot d\vec{r} = \int_{AB} \vec{F} \cdot d\vec{r} + \int_{BC} \vec{F} \cdot d\vec{r} + \int_{CA} \vec{F} \cdot d\vec{r}$$

$$\int_{AB} \vec{F} \cdot d\vec{r} = \int_2^0 (x^2 y \vec{a}_x - y \vec{a}_y) \cdot dx \vec{a}_x$$

$$= \int_2^0 x^2 y dx$$

$$\int_{AB} \vec{F} \cdot d\vec{r} = 0 \quad \because y=0 \text{ for path AB.}$$

$$\int_{BC} \vec{F} \cdot d\vec{r} = \int_{BC} (x^2 y \vec{a}_x - y \vec{a}_y) \cdot (dx \vec{a}_x + dy \vec{a}_y)$$

$$= \int_{BC} (x^2 y dx - y dy)$$

Equation of path BC is  $y=x$  i.e.  $dy=dx$

$$= \int_{BC} (x^3 dx - x dx)$$

$$= \int_0^1 (x^3 - x) dx$$

$$= \left[ \frac{x^4}{4} - \frac{x^2}{2} \right]_0^1 = \left[ \frac{1}{4} - \frac{1}{2} \right]$$

$$= -\frac{(2-4)}{8} = -\frac{2}{8} = -\frac{1}{4}$$

$$\int_{CA} \vec{f} \cdot d\vec{l} = \int_{CA} (x^2 y \vec{a}_x - y \vec{a}_y) \cdot (dx \vec{a}_x + dy \vec{a}_y)$$

$$= \int_{CA} (x^2 y dx - y dy)$$

Equation of path CA       $C(x_1, y_1)$      $A(x_2, y_2)$

$$\frac{y - y_1}{y_2 - y_1} = \frac{x - x_1}{x_2 - x_1}$$

$$\frac{y - 1}{0 - 1} = \frac{x - 1}{2 - 1}$$

$$y - 1 = -x + 1$$

$$\boxed{y = 2 - x}$$

$$\int_{CA} (x^2 y dx - y dy) = \int_1^2 x^2 (2 - x) dx - \int_1^0 y dy$$

$$= \int_1^2 (2x^2 - x^3) dx - \int_1^0 y dy$$

$$= \left[ \frac{2x^3}{3} - \frac{x^4}{4} \right]_1^2 - \left[ \frac{y^2}{2} \right]_1^0$$

$$= \left[ \frac{16}{3} - \frac{16}{4} \right] - \left[ \frac{2}{3} - \frac{1}{4} \right] - \left[ 0 - \frac{1}{2} \right]$$

$$= \left[ \frac{16 - 12}{3} \right] - \left[ \frac{8 - 3}{12} \right] + \frac{1}{2}$$

$$= \left[ \frac{4}{3} - \frac{5}{12} \right] + \frac{1}{2}$$

$$= \left[ \frac{16-5}{12} \right] + \frac{1}{2}$$

$$= \frac{11}{12} + \frac{1}{2}$$

$$= \frac{11+6}{12}$$

$$\oint_{CA} \vec{F} \cdot d\vec{r} = \frac{17}{12}$$

$$\begin{aligned} \oint \vec{F} \cdot d\vec{r} &= 0 - \frac{1}{4} + \frac{17}{12} \\ &= \frac{17-3}{12} = \frac{14}{12} \end{aligned}$$

R.H.S

$$\nabla \times \vec{F} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ x^2y & -y & 0 \end{vmatrix}$$

$$= \vec{a}_x \left( 0 + \frac{\partial y}{\partial z} \right) - \vec{a}_y \left( 0 - \frac{\partial (x^2y)}{\partial z} \right)$$

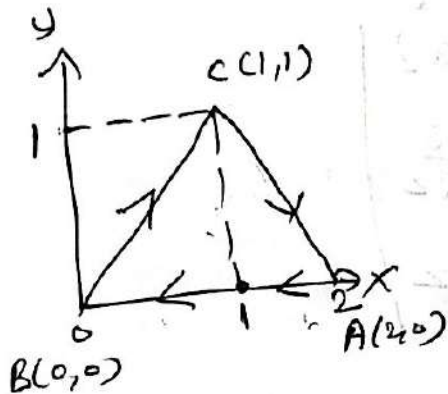
$$+ \vec{a}_z \left( \frac{\partial (-y)}{\partial x} - \frac{\partial (x^2y)}{\partial y} \right)$$

$$= 0 - 0 + \vec{a}_z (-x^2)$$

$$\nabla \times \vec{F} = -x^2 \vec{a}_z$$

$$\iint (\nabla \times \vec{F}) \cdot d\vec{S} = \iint -x^2 \vec{a}_z \cdot dx dy \vec{a}_z$$

$$= \iint -x^2 dx dy$$



Now split the area into two <sup>right angled</sup> triangles.

for first triangle, the equation of line is  $y=x$ , hence use  $dy=dx$  &  $x$  varies from 1 to 0.

for second triangle, the equation of line is  $y=2-x$ , hence use  $dy=-dx$  &  $x$  varies from 2 to 1.

$$\iint (\nabla \times \vec{F}) \cdot d\vec{S} = \int_{x=1}^0 -x^2 \cdot dx + \int_{x=2}^1 -x^2 (2-x) dx$$

$$= \int_1^0 -x^2 dx + \int_2^1 (-2x^2 + x^3) dx$$

$$= \left[ -\frac{x^3}{3} \right]_1^0 + \left[ -\frac{2x^3}{3} + \frac{x^4}{4} \right]_2^1$$

$$= \left[ 0 + \frac{1}{3} \right] + \left[ \left( -\frac{2}{3} + \frac{1}{4} \right) - \left( -\frac{16}{3} + \frac{16}{4} \right) \right]$$

$$= \frac{1}{3} + \left[ \frac{(3-8)}{12} - \frac{(-64+48)}{12} \right]$$

$$= \frac{1}{3} + \left[ \frac{(-5)}{12} - \frac{(-16)}{12} \right] = \frac{1}{3} + \left[ \frac{16-5}{12} \right]$$

$$= \frac{1}{3} + \frac{11}{12} = \frac{4+11}{12} = \frac{15}{12}$$

⇒ curl of the vectors

→ Cartesian coordinate system.

$$\nabla \times \vec{F} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ F_x & F_y & F_z \end{vmatrix}$$

→ cylindrical.

$$\nabla \times \vec{F} = \frac{1}{r} \begin{vmatrix} \vec{a}_r & r\vec{a}_\phi & \vec{a}_z \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ F_r & F_\phi & F_z \end{vmatrix}$$

→ Spherical.

$$\nabla \times \vec{F} = \frac{1}{r^2 \sin \theta} \begin{vmatrix} \vec{a}_r & r\vec{a}_\theta & r\sin \theta \vec{a}_\phi \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \theta} & \frac{\partial}{\partial \phi} \\ F_r & rF_\theta & r\sin \theta F_\phi \end{vmatrix}$$

⇒ Null Identities.

There are two vector identities based on curl & divergence of the field, which are called null vector identities.

i)  $\nabla \times (\nabla V) = 0$

ii)  $\nabla \cdot (\nabla \times \vec{A}) = 0$



→ The curl of the gradient of any scalar field is identically zero.

→ The divergence of the curl of any vector field is identically zero.

⇒ Helmholtz's theorem.

It is based on curl & divergence of vector field. It states that a vector field is ~~uniquely~~ uniquely defined within an additive constant by specifying its divergence and curl.

$$\vec{B} = -\nabla U + (\nabla \times \vec{A})$$

$U \rightarrow$  scalar field

$\vec{A} \rightarrow$  vector field.

$\vec{B}$  can be divided into two components

i) Gradient of scalar field  $U$

ii) Curl of the vector field  $\vec{A}$ .

Divergence of  $\vec{B}$

$$\nabla \cdot \vec{B} = \nabla \cdot (-\nabla U) + \nabla \cdot (\nabla \times \vec{A})$$

According to ~~second~~ Null Identities (ii)

$$\nabla \cdot \vec{B} = \nabla \cdot (-\nabla U) \quad \because \nabla \cdot (\nabla \times \vec{A}) = 0$$

$$\boxed{\nabla \cdot \vec{B} = e}$$

$$\because \nabla \cdot (-\nabla U) \neq 0$$

$$\text{put } \nabla \cdot (-\nabla U) = e$$

This is non solenoidal as divergence is non-zero

## curl of $\vec{B}$

$$\nabla \times \vec{B} = \nabla \times (-\nabla U) + \nabla \times (\nabla \times \vec{A})$$

According to Vec Identities (i)

$$\nabla \times (\nabla U) = 0$$

$$\text{then } \nabla \times \vec{B} = \nabla \times (\nabla \times \vec{A})$$

$$\boxed{\nabla \times \vec{B} = \vec{J}}$$

$$\text{put } \nabla \times (\nabla \times \vec{A}) = \vec{J}$$

This is rotational as the curl is not equal to zero

According to Helmholtz's theorem, four types of fields are defined

i) field is non solenoidal & rotational

$$\nabla \cdot \vec{B} = \rho \quad \& \quad \nabla \times \vec{B} = \vec{J}$$

ii) field is non solenoidal & irrotational

$$\nabla \cdot \vec{B} = \rho \quad \& \quad \nabla \times \vec{B} = 0$$

iii) field is solenoidal & rotational

$$\nabla \cdot \vec{B} = 0 \quad \& \quad \nabla \times \vec{B} = \vec{J}$$

iv) field is solenoidal & irrotational

$$\nabla \cdot \vec{B} = 0 \quad \& \quad \nabla \times \vec{B} = 0$$

$$\boxed{\vec{B} = \vec{J} \cdot \vec{r}}$$

## UNIT-II ELECTROSTATICS

Coulomb's law.

Coulomb's law states that the force of attraction or repulsion between two point charges is directly proportional to the product of charge ( $Q_1, Q_2$ ) and inversely proportional to the square of distance between them.



$$F \propto \frac{Q_1 Q_2}{r^2}$$

$$F = k \frac{Q_1 Q_2}{r^2} \quad k = \frac{1}{4\pi\epsilon_0}$$

$$F = \frac{1}{4\pi\epsilon_0} \frac{Q_1 Q_2}{r^2}$$

in Newton

$\epsilon_0$  - permittivity  $\epsilon$

$$\epsilon = \epsilon_0 \epsilon_r$$

$\epsilon_0$  = permittivity in free space

$$= 8.85 \times 10^{-12} \text{ F/m}$$

$\epsilon_r$  = relative permittivity in medium

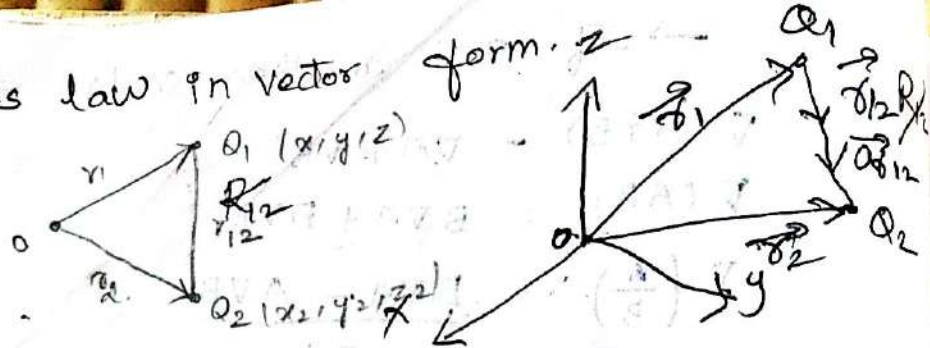
In air medium  $\epsilon_r = 1$ .

$$F = \frac{1}{4\pi\epsilon_0 \epsilon_r} \frac{Q_1 Q_2}{r^2} \quad \epsilon_r = 1$$

In air medium

$$F = \frac{Q_1 Q_2}{4\pi\epsilon_0 r^2}$$

Coulomb's law in vector form.



Distance of vector  $Q_1$  and  $Q_2$

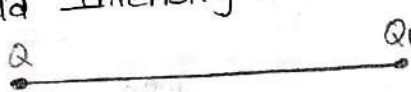
$$\vec{F} = \frac{Q_1 Q_2}{4\pi\epsilon_0 r_{12}^2} \hat{a}_{12}$$

$$\vec{a}_{12} = \frac{\vec{r}_{12}}{|\vec{r}_{12}|}$$

$$\vec{r}_{12} = \vec{r}_2 - \vec{r}_1$$

$$\vec{F} = \frac{Q_1 Q_2}{4\pi\epsilon_0 r_{12}^2} \frac{\vec{r}_2 - \vec{r}_1}{|\vec{r}_2 - \vec{r}_1|}$$

Electric Field Intensity :- (E)



$$E = \frac{F}{Q_t}$$

Electric Field Intensity is defined as force experienced by test charge <sup>(or) none</sup> due to main charge. It is known as Electric field Intensity.

Unit is  $\frac{N}{C}$

$$E = \frac{Q Q_2}{4\pi\epsilon_0 r^2}$$

$$E = \frac{Q}{4\pi\epsilon_0 r^2} = \frac{C}{m} \text{ or } \frac{V}{m}$$

## Charge distribution:

→ point charge

Point charge may  $+q$  (or)  $-q$  is located in free space  $\vec{E} = \frac{Q}{4\pi\epsilon_0 r^2} \vec{a}_r \rightarrow \textcircled{1}$

→ Line charge

It is defined as total number of charges presented throughout the entire length ( $l$ ) line charge becomes

$$P_l = \frac{Q}{l}$$

Let us consider a small elemental length

$$P_l = \frac{dQ}{dl}$$

$$dQ = P_l dl$$

$$Q = \int P_l dl \rightarrow \textcircled{2}$$

$$\vec{E} = \frac{\int P_l dl \vec{a}_r}{4\pi\epsilon_0 r^2}$$

Unit  $\rightarrow \text{C/m}$

Surface charge

It is defined as ratio of total no. of charges presented throughout the surface area.

$$P_s = \frac{Q}{S}$$

consider small elemental surface

$$P_s = \frac{dQ}{dS}$$

$$dQ = P_s dS$$

$$Q = \int P_s dS = \oint P_s dS \rightarrow \textcircled{4}$$

sub eqn ④ in ①

$$\vec{E} = \frac{\iint P_s ds \vec{a}_r}{4\pi \epsilon_0 r^2}$$

unit  $\Rightarrow C/m^2$

Volume charge

It is defined as ratio of total charges located throughout the entire volume

$$\rho_v = \frac{Q}{V}$$

Consider a small element

$$\rho_v = \frac{dQ}{dV}$$

$$dQ = \rho_v dV$$

$$Q = \iiint \rho_v dV$$

$$Q = \int_V \rho_v dV \rightarrow \textcircled{5}$$

Sub eqn ⑤ in ①

$$\vec{E} = \frac{\iiint \rho_v dV \vec{a}_r}{4\pi \epsilon_0 r^2}$$

Electric ~~Field~~ <sup>Flux</sup> Density

It is defined as total number of lines of force in a electric field is called as Electric flux density  $\vec{D}$

$$D = \frac{Q}{S}$$

$$S = \frac{Q}{D}$$

Total no. of charges per unit area

$$D \propto \vec{E}$$

$$\vec{D} = \epsilon_0 \epsilon_r \vec{E}$$

In air medium  $\epsilon_r = 1$

$$\vec{D} = \epsilon_0 \vec{E}$$

$$= \epsilon_0 \left( \frac{Q}{4\pi\epsilon_0 r^2} \vec{a}_r \right)$$

$$\vec{D} = \frac{Q}{4\pi r^2} \vec{a}_r$$

✓ Properties of Electric flux Density.

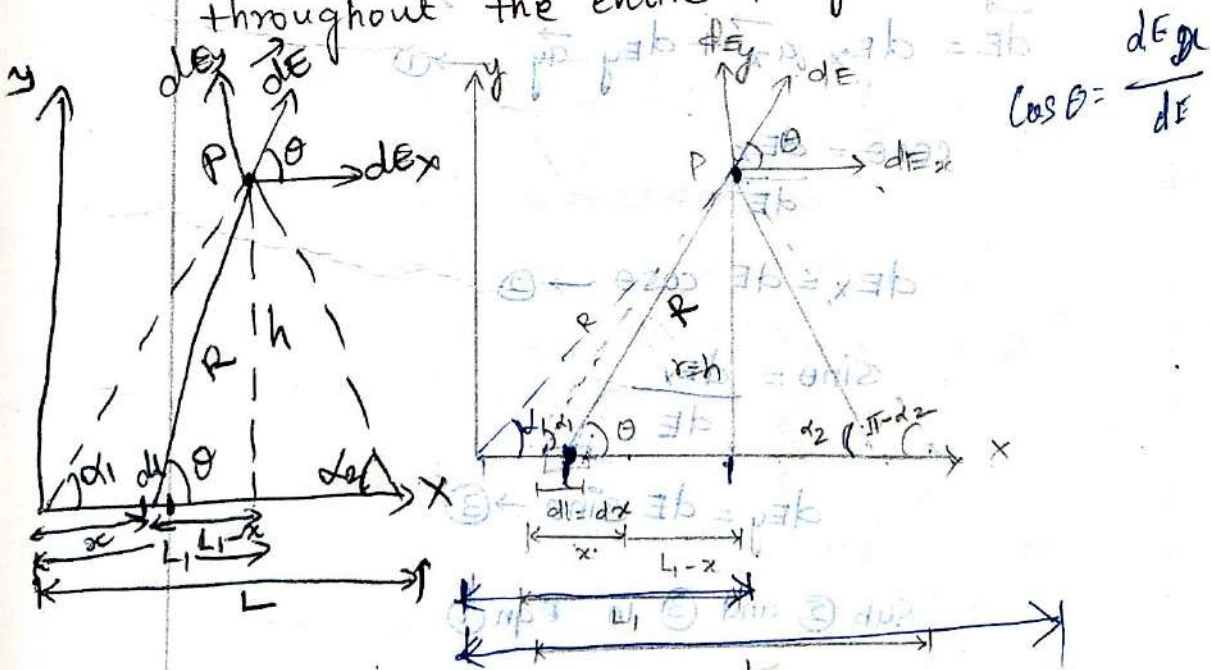
The force of lines are parallel to each other

The force of lines are never cross each other

The electric flux lines are entering or leaving normally charged surface.

The electric flux line are strong. the electric field intensity become strong.

⊗ Electric field due to line charge distribution.  
 (or) To find the Electric field intensity at a point 'P' where the charges are distributed throughout the entire length



Let us consider the charges are presented throughout entire length due to point charge in free space.

Here Force Experienced by test charge or point charge due to main charge electric field intensity is Produced.

Hence to find out the electric field intensity in small elemental length  $dl$ .

$\theta$   $\rightarrow$  angle between point charge and main charge

$r$   $\rightarrow$  Indicate the equi potential surface in free space

$\alpha_1$   $\rightarrow$  angle btw starting of the length to point charge.

$(\pi - \alpha_2)$   $\rightarrow$  angle btw end point and point charge.

Case (i)

$$d\vec{E} = dE_x \vec{a}_x + dE_y \vec{a}_y \rightarrow \textcircled{1}$$

$$\cos\theta = \frac{dE_x}{dE}$$

$$dE_x = dE \cos\theta \rightarrow \textcircled{2}$$

$$\sin\theta = \frac{dE_y}{dE}$$

$$dE_y = dE \sin\theta \rightarrow \textcircled{3}$$

Sub  $\textcircled{2}$  and  $\textcircled{3}$  in Eqn  $\textcircled{1}$

$$d\vec{E} = dE \cos\theta \vec{a}_x + dE \sin\theta \vec{a}_y \rightarrow \textcircled{4}$$



According to Electric field intensity in line charge distribution

$$E = \frac{\int \rho_l dl}{4\pi\epsilon_0 R^2} \vec{a}_R$$

from diagram  $dl = dx$

$$\int \vec{E} = dE$$

$$dE = \frac{\rho_l dl}{4\pi\epsilon_0 R^2} \vec{a}_R \rightarrow \textcircled{5}$$

Sub eqn ⑤ in ④.

$$d\vec{E} = \frac{\rho_l dx}{4\pi\epsilon_0 R^2} \cos\theta \vec{a}_x +$$

$$\frac{\rho_l dx}{4\pi\epsilon_0 R^2} \sin\theta \vec{a}_y \rightarrow \textcircled{6}$$

Here there are three different parameters

$R$ ,  $x$ ,  $\theta$  it may be varied due to that we need a single line integral convert this 3 parameters into one parameter

Let us consider a triangle from the diagram.

$$\cot\theta = \frac{L_1 - x}{h}$$

$$L_1 - x = h \cot\theta$$

$$-dx = h [-\operatorname{cosec}^2\theta d\theta]$$

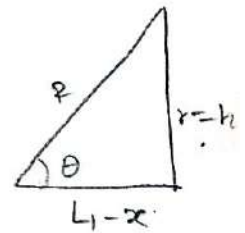
$$dx = h \operatorname{cosec}^2\theta d\theta \rightarrow \textcircled{7}$$

$$\operatorname{cosec}\theta = \frac{R}{h}$$

$$R = h \operatorname{cosec}\theta \rightarrow \textcircled{8}$$

Sub ⑦ & ⑧ in eqn ⑥

$$dE = \frac{\rho_l (h \operatorname{cosec}^2\theta d\theta)}{4\pi\epsilon_0 (h^2 \operatorname{cosec}^2\theta)} \cos\theta \vec{a}_x +$$



$$\frac{\rho_l (h \operatorname{cosec}^2 \theta) \sin \theta \vec{a}_y}{4\pi\epsilon_0 (h^2 \cos^2 \theta)}$$

$$d\vec{E} = \frac{\rho_l}{4\pi\epsilon_0 h} [\cos \theta d\theta \vec{a}_x + \sin \theta d\theta \vec{a}_y]$$

Integrate on BS

$$\vec{E} = \frac{\rho_l}{4\pi\epsilon_0 h} \left[ \int_{\alpha_1}^{\pi-\alpha_2} \cos \theta d\theta \vec{a}_x + \int_{\alpha_1}^{\pi-\alpha_2} \sin \theta d\theta \vec{a}_y \right]$$

$$= \frac{\rho_l}{4\pi\epsilon_0 h} \left[ (\sin \theta)_{\alpha_1}^{\pi-\alpha_2} \vec{a}_x + (-\cos \theta)_{\alpha_1}^{\pi-\alpha_2} \vec{a}_y \right]$$

$$= \frac{\rho_l}{4\pi\epsilon_0 h} \left\{ (\sin(\pi-\alpha_2) - \sin \alpha_1) \vec{a}_x + (-\cos(\pi-\alpha_2) + \cos \alpha_1) \vec{a}_y \right\}$$

For finite length

$$\vec{E} = \frac{\rho_l}{4\pi\epsilon_0 h} \left\{ \sin(\pi-\alpha_2) - \sin \alpha_1 \right\} \vec{a}_x + \left\{ -\cos(\pi-\alpha_2) + \cos \alpha_1 \right\} \vec{a}_y$$

$$\vec{D} = \epsilon \vec{E}$$

For free space

$$\vec{D} = \epsilon_0 \vec{E} \quad \epsilon_0 = 1$$

$$\vec{D} = \frac{\rho_l}{4\pi\epsilon_0 h} \left\{ \sin(\pi-\alpha_2) - \sin \alpha_1 \right\} \vec{a}_x + \left\{ -\cos(\pi-\alpha_2) + \cos \alpha_1 \right\} \vec{a}_y$$

Case (ii)

Let us consider an infinite length

$$\alpha_1 = 0 \quad \alpha_2 = 0 \quad \text{So,}$$

$$\begin{aligned} \sin(180^\circ - \theta) &= \sin \theta \\ \cos(180^\circ - \theta) &= -\cos \theta \\ \sin(\pi - \alpha_1) &= \sin \alpha_1 \end{aligned}$$

$$\vec{E} = \frac{\rho_l}{4\pi\epsilon_0 h} \left\{ \sin(\pi - \alpha_2) - \sin \alpha_1 \right\} \vec{a}_x + \left\{ -\cos(\pi - \alpha_2) + \cos \alpha_1 \right\} \vec{a}_y$$

$$= \frac{\rho_l}{4\pi\epsilon_0 h} \left\{ (\sin \alpha_2 - \sin \alpha_1) \vec{a}_x + (-\cos \alpha_2 + \cos \alpha_1) \vec{a}_y \right\}$$

$$\alpha_2 = 0 \quad \alpha_1 = 0$$

$$= \frac{Pl}{4\pi\epsilon_0 h} \left\{ (0 - 0) \vec{a}_x + (\cancel{-1} + 1) \vec{a}_y \right\}$$

$$= \frac{Pl}{4\pi\epsilon_0 h} \{ 0 \vec{a}_x + 2 \vec{a}_y \}$$

$$= \frac{Pl}{4\pi\epsilon_0 h} 2 \vec{a}_y$$

$$\sin(\pi - \theta) = \sin \theta$$

$$\cos(\pi - \theta) = -\cos \theta$$

$$\vec{E} = \frac{Pl}{2\pi\epsilon_0 h} \vec{a}_y \rightarrow \textcircled{11}$$

$$\epsilon_0 \left( \frac{Pl}{2\pi\epsilon_0 h} \right) = \frac{Pl}{2\pi h}$$

$$D = \epsilon_0 \vec{E} \text{ in free space } \epsilon_0$$

$$\vec{D} = \frac{Pl}{2\pi h} \vec{a}_y \rightarrow \textcircled{12}$$

Case (iii)

Let us consider the point charge 'P' at mid position. Hence  $\alpha_1 = \alpha_2 = \alpha$

From eqn (4)

$$\vec{E} = \frac{Pl}{4\pi\epsilon_0 h} \left\{ (\sin(\pi - \alpha_2) - \sin \alpha_1) \vec{a}_x + (-\cos(\pi - \alpha_2) + \cos \alpha_1) \vec{a}_y \right\}$$

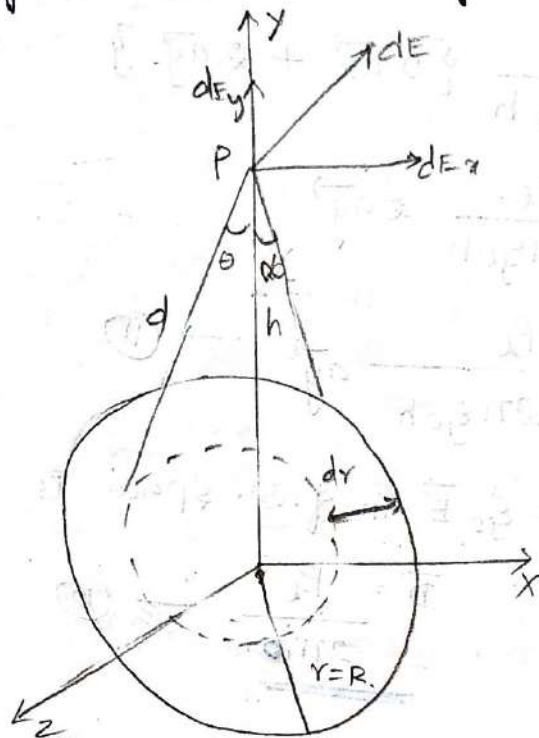
$$\vec{E} = \frac{Pl}{4\pi\epsilon_0 h} \left\{ (\sin(\pi - \alpha) - \sin \alpha) \vec{a}_x + (-\cos(\pi - \alpha) + \cos \alpha) \vec{a}_y \right\}$$

$$= \frac{Pl}{4\pi\epsilon_0 h} \left\{ 0 + (\cos \alpha + \cos \alpha) \vec{a}_y \right\}$$

$$= \frac{Pl}{4\pi\epsilon_0 h} (2 \cos \alpha) \vec{a}_y$$

$$\vec{E} = \frac{Pl}{2\pi\epsilon_0 h} \cos \alpha \vec{a}_y \rightarrow \textcircled{13}$$

To find Electric field intensity and density on the circular axis. Here the charges are uniformly distributed throughout the circular disc.



$$\cos \theta = \frac{dE_y}{dE}$$

$$dE_y = dE \cos \theta \rightarrow (1)$$

According to surface charge distribution

$$\vec{E} = \frac{\iint \rho_s \cdot d\vec{s}}{4\pi \epsilon_0 R^2} \vec{a}_R \rightarrow (2)$$

diff on B.S

$$d\vec{E} = \frac{\rho_s ds}{4\pi \epsilon_0 R^2} \vec{a}_R \rightarrow (3)$$

Sub eqn (3) in (4)

$$d\vec{E}_y = \frac{\rho_s ds}{4\pi \epsilon_0 R^2} \cos \theta \vec{a}_y$$

ds → differential surface

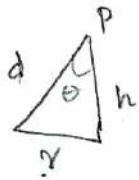
$$\text{In circle } S = A = \pi r^2$$

$$A_s = dA = 2\pi r dr$$

$$d\vec{E}_y = \frac{\rho_s (2\pi r \cdot dr) \cos\theta}{4\pi \epsilon_0 d^2} \vec{a}_y \rightarrow \textcircled{4}$$

$d \rightarrow$  distance btw point charge and inner

circle



$$\tan\theta = \frac{r}{h}$$

$$\cos\theta = \frac{h}{d}$$

$$r = h \tan\theta$$

$$d = \frac{h}{\cos\theta}$$

$$dr = h \sec^2\theta d\theta$$

$$d = h \sec\theta$$

$$d\vec{E}_y = \frac{\rho_s 2\pi (h \tan\theta) (h \sec^2\theta d\theta) \cos\theta}{4\pi \epsilon_0 (h^2 \sec^2\theta)} \vec{a}_y$$

$$\vec{dE}_y = \frac{\rho_s \sin\theta}{2\epsilon_0} \vec{a}_y \rightarrow \textcircled{5}$$

Integrate on  $\theta$ 's

$$\vec{E}_y = \int_0^\alpha \frac{\rho_s \sin\theta}{2\epsilon_0} \vec{a}_y$$

$$= \frac{\rho_s}{2\epsilon_0} [-\cos\theta]_0^\alpha \vec{a}_y$$

$$= \frac{\rho_s}{2\epsilon_0} [1 - \cos\alpha] \vec{a}_y$$

$$2\theta = \alpha \\ \theta = \frac{\alpha}{2}$$

$$= \frac{\rho_s}{2\epsilon_0} \left[ 2 \sin^2 \frac{\alpha}{2} \right] \vec{a}_y \quad \sin^2\theta = \frac{1 - \cos 2\theta}{2}$$

$$\vec{E}_y = \frac{\rho_s}{\epsilon_0} \sin^2 \frac{\alpha}{2} \vec{a}_y \rightarrow \textcircled{6}$$

$$\sin^2 \frac{\alpha}{2} = \frac{1 - \cos\alpha}{2}$$

$$2 \sin^2 \frac{\alpha}{2} = 1 - \cos\alpha$$

$$\vec{D} = \epsilon_0 \vec{E}$$

In free space  $\vec{D} = \epsilon_0 \vec{E}$

$$\vec{D}_y = \rho_s \sin^2 \frac{\alpha}{2} \vec{a}_y \rightarrow \textcircled{7}$$

Case (ii)

The charges are distributed infinitely from circle to plane of sheet

$$\alpha = 90^\circ$$

$$\begin{aligned}\vec{E}_y &= \frac{\rho_s}{\epsilon_0} \sin^2 \frac{90}{2} \vec{a}_y \\ &= \frac{\rho_s}{\epsilon_0} \left(\frac{1}{\sqrt{2}}\right)^2\end{aligned}$$

$$\boxed{\vec{E}_y = \frac{\rho_s}{2\epsilon_0} \vec{a}_y}$$

$$D = \epsilon_0 \vec{E} \text{ in free}$$

$$D_y = \frac{\rho_s}{2} \vec{a}_y \rightarrow \textcircled{1}$$

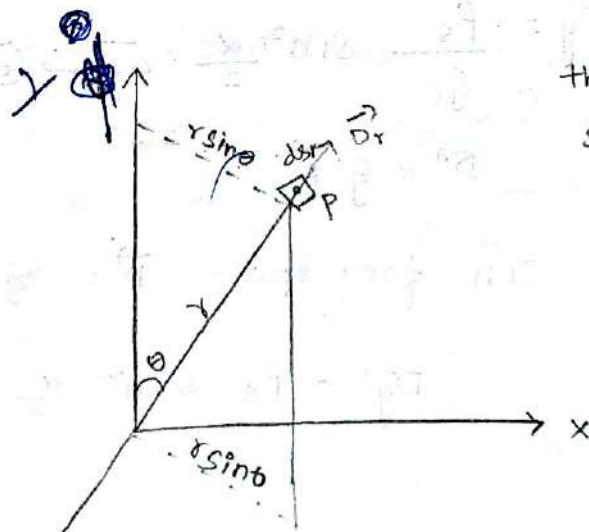
Gauss law

The total number of lines of force leaving from the charging surface area is equal to the total number of charges enclosed by the surface

$$\boxed{Q = \psi}$$

Where  $Q \rightarrow$  Total no. of charges  
 $\psi \rightarrow$  flux lines.

Proof:-



through  $\vec{r}$  direction  
So  $ds_r, \vec{D}_r$

Let us consider spherical coordinate s/y.

$$Q = \psi$$

AS per electric flux density

$$D = \frac{Q}{S} = \frac{dQ}{ds} \quad [Q = \psi]$$

$$D = \frac{\psi}{S} = \frac{d\psi}{ds}$$

$$d\psi = D_r ds_r$$

From diagram

$$d\psi = D_r ds_r \rightarrow \textcircled{1}$$

Let us consider the flux lines are produced outwards from the surface

By point charge

$$\vec{E} = \frac{Q}{4\pi\epsilon_0 r^2} \vec{a}_r \quad D = \epsilon_0 \vec{E}$$

$$\vec{D}_r = \frac{Q}{4\pi r^2} \vec{a}_r$$

In spherical coordinate s/m

$$ds_r = r^2 \sin\theta \, d\theta \, d\phi \, \vec{a}_r \rightarrow \textcircled{3}$$

sub  $\textcircled{2}$  &  $\textcircled{3}$  into  $\textcircled{1}$

$$d\psi = \frac{Q}{4\pi r^2} \vec{a}_r \cdot r^2 \sin\theta \, d\theta \, d\phi \, \vec{a}_r$$

$$d\psi = \frac{Q}{4\pi} \sin\theta \, d\theta \, d\phi$$

$$\psi = \frac{Q}{4\pi} \left[ \int_0^\pi \sin\theta \, d\theta \int_0^{2\pi} d\phi \right]$$

$$\psi = \frac{Q}{4\pi} \left[ (-\cos\theta)_0^\pi (\phi)_0^{2\pi} \right]$$

$$= \frac{Q}{4\pi} [4\pi]$$

$$\boxed{\psi = Q}$$

Hence Gauss law proved

Application:

Gauss law is applied to the surface if the following conditions are satisfied

- surface is enclosed
- electric flux density  $\vec{D}$  is either normal or tangential to the surface at each point of the surface
- $\vec{D}$  is constant over the point P out of the surface where  $\vec{D}$  is normal.

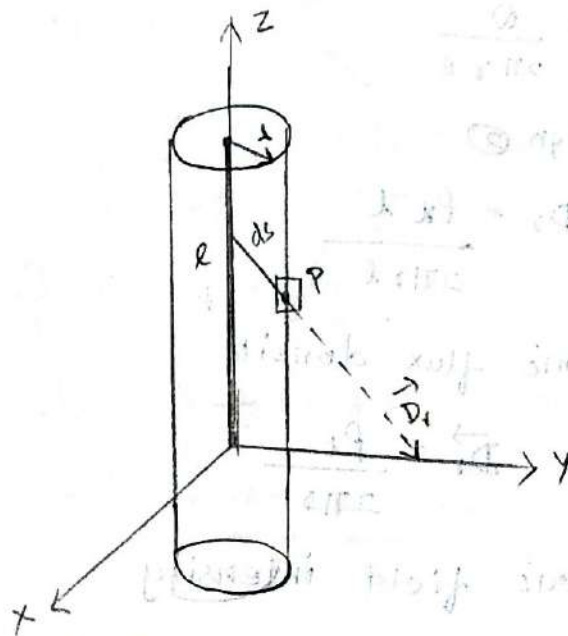
To find electric field intensity and density by using Gauss law. The charges are uniformly distributed throughout the infinite line <sup>line charge</sup>

Assume that infinite line is in Gaussian surface

consider cylindrical coordinate system  
↳ Line charge distribution

$$\left[ \begin{matrix} \pi r \\ 0 \\ 0 \end{matrix} \right] \left[ \begin{matrix} \pi r \\ 0 \\ 0 \end{matrix} \right] \frac{Q}{\pi r}$$





In cylindrical.

$$ds = r d\phi dz \vec{a}_r$$

Due to line charge distribution the charge density

$$\rho_e = \frac{Q}{l} = \frac{dQ}{dl} \rightarrow \textcircled{1}$$

$$dQ = \rho_e dl \quad \int d \rightarrow \text{cancel.}$$

$$Q = \int \rho_e dl$$

$$Q = \rho_e l \rightarrow \textcircled{2}$$

Electric flux density

$$\vec{D} = \frac{Q}{s} = \frac{dQ}{ds} = \frac{d\psi}{ds}$$

$$dQ = \vec{D} ds = \rho_e ds$$

$$Q = \iint \vec{D}_r ds_r$$

$$Q = \iint \vec{D}_r [r d\phi dz \vec{a}_r]$$

$$= \int_0^l \int_0^{2\pi} D_r r d\phi dz$$

$$= D_r r [\phi]_0^{2\pi} [z]_0^l$$

$$= D_r r (2\pi) l$$

$$Q = 2\pi r l D_r$$

Sub eqn ②

$$D_r = \frac{\rho_l l}{2\pi r l}$$

Electric flux density

$$\vec{D}_r = \frac{\rho_l}{2\pi r}$$

Electric field intensity

$$\vec{E} = \frac{\vec{D}_r}{\epsilon_0}$$

$$\vec{E} = \frac{\rho_l}{2\pi r \epsilon_0}$$

To find Electric field intensity and density when the charges in Cartesian rectangular box or pill box. Assume the pill box is infinite sheet of charges

Surface charge distribution consider

Cartesian coordinate system.

Electric flux Density

$$D = \frac{Q}{S} = \frac{dq}{ds}$$

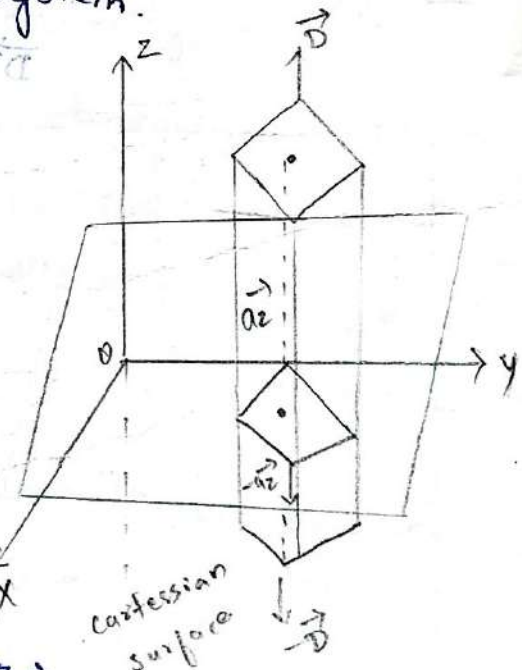
$$dq = D \cdot ds$$

$$Q = \iint \vec{D} \cdot d\vec{s}$$

Let us consider electric flux density along z direction

$$Q = \int_{\text{Top}} \vec{D}_z \cdot d\vec{s}_z + \int_{\text{side}} \vec{D}_z \cdot d\vec{s}_z +$$

$$\int_{\text{Bottom}} \vec{D}_z \cdot d\vec{s}_z$$



Let  $\int_{\text{Side}} \vec{D}_z ds_z = 0$

$$Q = \iint_{\text{Top}} \vec{D}_z ds_z + \iint_{\text{Bot}} \vec{D}_z ds_z$$

$$= \iint_{\text{Top}} \vec{D}_z \vec{a}_z [dx dy \vec{a}_z] + \iint \vec{D}_z - \vec{a}_z [dx dy (-\vec{a}_z)]$$

$$= \iint \vec{D}_z dx dy + \iint \vec{D}_z dx dy$$

$$= 2 \iint \vec{D}_z dx dy$$

$$= 2 \vec{D}_z [x] [y]$$

$$= 2 \vec{D}_z (xy)$$

$$Q = 2 \vec{D}_z (A) \rightarrow \textcircled{1}$$

$$P_s = \frac{Q}{A}$$

$$P_s A = Q \rightarrow \textcircled{2}$$

Sub  $\textcircled{2}$  in  $\textcircled{1}$

$$P_s A = 2 \vec{D}_z A$$

$$P_s = 2 \vec{D}_z$$

$$\boxed{\vec{D}_z = \frac{P_s}{2}}$$

Electric field Intensity

$$E = \frac{D}{\epsilon_0} \hat{i}_n$$

free space

$$\boxed{\vec{E} = \frac{P_s}{2 \epsilon_0}}$$

Absolute Electric Potential (V)

Electric potential is defined as work done moving a unit positive charge  $Q$  from infinite point to the given point is known as potential denoted by  $V$

$$E = \frac{Q}{4\pi\epsilon_0 r^2} \text{ Volt}$$

$$E = \frac{F}{q}$$

$$\frac{F}{q}$$

$$W = F \times d$$

$$W = F \cdot d$$

$$W = (E \cdot q) \cdot d$$

work done on unit positive charge also called as potential

$$V = \frac{W}{q} \text{ unit J/C}$$

$$V = \frac{Q}{4\pi\epsilon_0 r}$$

$$V = -\int_{r_1}^{r_2} E \, ds$$

$$E = \frac{Q}{4\pi\epsilon_0 r^2}$$

$$V = -\int_{r_1}^{r_2} E \, ds$$

$$= -\int_{r_1}^{r_2} \frac{Q}{4\pi\epsilon_0 r^2} \, dr$$

$$= \frac{-Q}{4\pi\epsilon_0} \int_{r_1}^{r_2} \frac{1}{r^2} \, dr$$

$$= \frac{-Q}{4\pi\epsilon_0} \left[ -\frac{1}{r} \right]_{r_1}^{r_2}$$

$$\frac{r^{-2+1}}{-2+1} = -\frac{1}{r}$$

$$= \frac{-Q}{4\pi\epsilon_0} \left[ \frac{1}{r_2} + \frac{1}{r_1} \right]$$

$$V = \frac{Q}{4\pi\epsilon_0} \left[ \frac{1}{r_2} - \frac{1}{r_1} \right]$$

$$V = -\int E \, dl \rightarrow \textcircled{2}$$

Absolute Potential (V)  $\Delta V = -E \Delta r$

$$E = -\frac{\Delta V}{\Delta l}$$

$$E = -\text{grad } V$$

$$E = -\nabla V \rightarrow \textcircled{3}$$

$$E = \frac{V}{d} \text{ or } \frac{V}{l}$$

d - distance

## Potential Difference:-

It is defined as Work done moving a unit positive charge from one point to another point

$$V_A = \frac{+Q}{4\pi\epsilon_0 r_A}$$

$$V_B = \frac{-Q}{4\pi\epsilon_0 r_B}$$

$$V_{AB} = V_A - V_B$$

$$= \frac{Q}{4\pi\epsilon_0 r_A} - \frac{-Q}{4\pi\epsilon_0 r_B}$$

$$V_{AB} = \frac{Q}{4\pi\epsilon_0 r_A} + \frac{Q}{4\pi\epsilon_0 r_B}$$

Find  $\vec{E}$  at  $(1,1,1)$  if potential  $V = xyz^2 + x^2yz + xy^2z$

$$\vec{E} = -\nabla V$$

$$\vec{E} = -\left[ \frac{\partial V}{\partial x} \vec{a}_x + \frac{\partial V}{\partial y} \vec{a}_y + \frac{\partial V}{\partial z} \vec{a}_z \right]$$

$$= -\left[ yz^2 \vec{a}_x + x^2 z \vec{a}_y + \right]$$

$$= -\frac{\partial}{\partial x} (xyz^2 + x^2yz + xy^2z) + \frac{\partial}{\partial y}$$

$$= -\left[ yz^2 + 2x yz + y^2 z \right] + \left[ xz^2 + x^2 z + 2xy z \right] \vec{a}_y$$

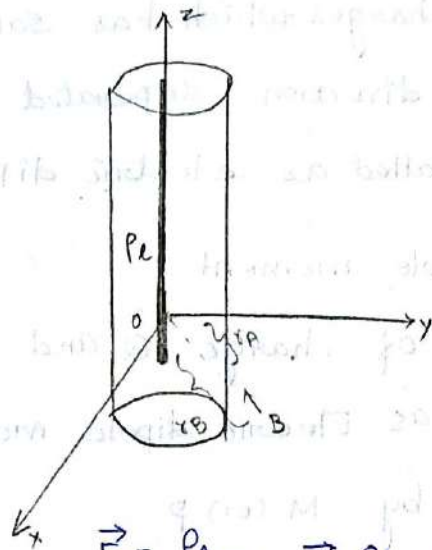
$$+ \left[ 2xyz + x^2 y + xy^2 \right] \vec{a}_z$$

$$= -\left[ 1 + 2 + 1 \right] + \left[ 1 + 1 + 2 \right] \vec{a}_y + \left[ 4 \right] \vec{a}_z$$

$$= -4 \vec{a}_x - 4 \vec{a}_y - 4 \vec{a}_z$$

$$= -4 \left[ \vec{a}_x + \vec{a}_y + \vec{a}_z \right] \text{ V/m}$$

# Potential Difference for different configuration



$$\vec{E} = \frac{\rho_l}{2\pi\epsilon_0 r} \vec{a}_r \text{ in } xy \text{ plane}$$

(line charge distribution along cylindrical coordinate system)

$$V = -\int \vec{E} \cdot d\vec{l}$$

For cylindrical coordinate system

$$d\vec{l} = dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z$$

$$V = -\int \frac{\rho_l}{2\pi\epsilon_0 r} \vec{a}_r [dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z]$$

$$= -\int_{r_B}^{r_A} \frac{\rho_l}{2\pi\epsilon_0 r} dr$$

$$= -\frac{\rho_l}{2\pi\epsilon_0} \int_{r_B}^{r_A} \frac{1}{r} dr$$

$$= -\frac{\rho_l}{2\pi\epsilon_0} [\log r]_{r_B}^{r_A}$$

$$= -\frac{\rho_l}{2\pi\epsilon_0} [\log r_A - \log r_B]$$

$$= -\frac{\rho_l}{2\pi\epsilon_0} \left( \log \left( \frac{r_A}{r_B} \right) \right)$$

$$\Rightarrow \left[ \frac{1}{r_B} - \frac{1}{r_A} \right] = \frac{\rho_l}{2\pi\epsilon_0} \log \left( \frac{r_B}{r_A} \right)$$

2m

Electric dipole:

Two point charges which has same magnitude but opposite direction separated by small distance is called as electric dipole.

2m

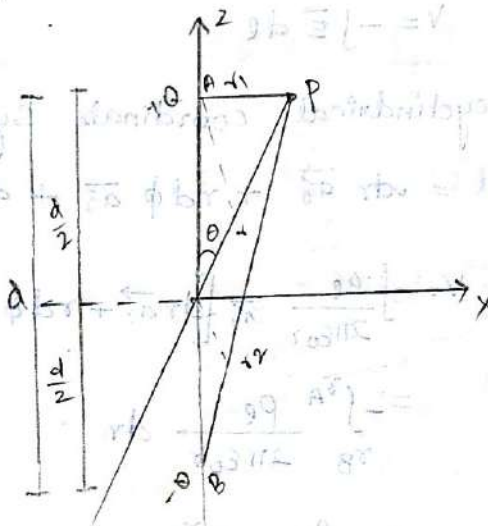
Electric dipole moment

The product of charge  $Q$  and displacement  $d$  is called as Electric dipole moment.

It denoted by  $M$  (or)  $P$

$$M = Qd$$

Relationship between Electric dipole moment and potential



$$V_1 = \frac{Q}{4\pi\epsilon_0 r_1}$$

$$V_2 = \frac{-Q}{4\pi\epsilon_0 r_2}$$

$$V = V_1 + V_2$$

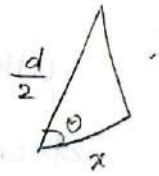
$$= \frac{Q}{4\pi\epsilon_0 r_1} + \frac{-Q}{4\pi\epsilon_0 r_2}$$

$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{1}{r_1} - \frac{1}{r_2} \right] \rightarrow \text{①}$$

same magnitude  
↔ direction.

$$\cos \theta = \frac{x}{d/2}$$

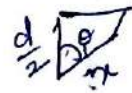
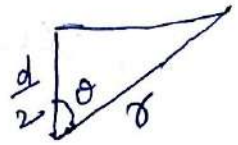
$$x = \frac{d}{2} \cos \theta$$



$$r_1 = r - x$$

$$r_2 = r + x$$

$$V = \frac{Q}{4\pi\epsilon_0} \left[ \frac{1}{r-x} - \frac{1}{r+x} \right]$$



$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{r+x-r+x}{r^2-x^2} \right]$$



$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{2x}{r^2-x^2} \right]$$

$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{2 \cdot \frac{d}{2} \cos \theta}{r^2 - \frac{d^2}{4} \cos^2 \theta} \right]$$

$$\frac{d^2}{4} \ll 1$$

$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{d \cos \theta}{r^2} \right]$$

$$= \frac{(Qd) \cos \theta}{4\pi\epsilon_0 r^2}$$

$$V = \frac{P \cos \theta}{4\pi\epsilon_0 r^2}$$

6m

### Continuity Equation

According to law of conservation energy of charges cannot be created nor destroyed

Let us assume cubical box the charges are present in the cubical box

The charges are slowly move from box to the outwards, the charges are

decreasing hence

$$I = -\frac{dq}{dt}$$



Where  $(-)$   $\rightarrow$  indicates the charges are decreasing

Current Density :-

$$\text{Current Density } J = \frac{I}{A} = \frac{dI}{dA} = \frac{dI}{ds}$$

$$dI = J \cdot ds$$

$$I = \int_S \vec{J} \cdot ds \rightarrow \textcircled{1}$$

By using divergence theorem

$$\oint_S \vec{J} \cdot ds = \int_V (\nabla \cdot \vec{J}) dv$$

$$\rho_v = \frac{Q}{V} = \frac{dQ}{dV}$$

$$dQ = \rho_v dv$$

$$Q = \int_V \rho_v dv \rightarrow \textcircled{2}$$

By continuity eqn

$$I = -\frac{dQ}{dt} = \int_S \vec{J} \cdot ds \rightarrow \textcircled{3}$$

$$-\frac{dQ}{dt} = \int_V (\nabla \cdot \vec{J}) dv$$

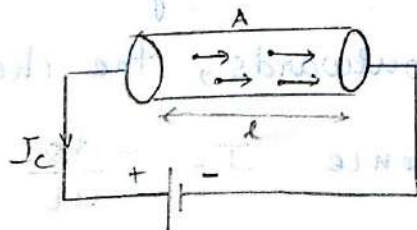
Sub  $\textcircled{2}$  in where

$$-\frac{d}{dt} \left( \int_V \rho_v dv \right) = \int_V \nabla \cdot \vec{J} dv$$

$$-\frac{\partial \rho_v}{\partial t} = \nabla \cdot \vec{J}$$

$$\boxed{\nabla \cdot \vec{J} = -\frac{\partial \rho_v}{\partial t}}$$

Current and current density :



$$I = \frac{Q}{t} = \frac{dQ}{dt}$$

$$I = \frac{V}{R}$$

$$n = \frac{I}{A}$$

By ohm's law

$$I_c = \frac{V}{R}$$

WKT  $R = \frac{\rho l}{A}$

$\frac{1}{\rho} = \text{conductivity}$

$$I_c = \frac{VA}{\rho l}$$

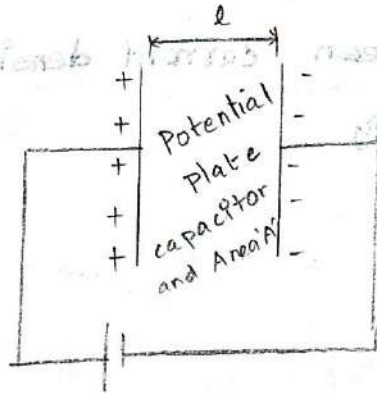
$$\frac{I_c}{A} = \frac{V}{\rho l}$$

$$\frac{V}{l} = \vec{E}$$

$$\frac{I_c}{A} = \frac{1}{\rho} \vec{E}$$

$$\boxed{J = \sigma \vec{E}}$$

Displacement current and Displacement current Density



Displacement current:

It is defined as current flowing through the capacitor when the

Voltage is applied across the capacitor is called Displacement current  $I_D$ .

$$I_D = \frac{Q}{t} = \frac{dQ}{dt} \rightarrow \textcircled{1}$$

$$C = \frac{Q}{V}$$

$$Q = CV \rightarrow \textcircled{2}$$

$$I_D = \frac{d(CV)}{dt}$$

$$I_D = c \frac{dv}{dt} \rightarrow \textcircled{3}$$

we know that  $c = \frac{\epsilon A}{d}$  and  $E = \frac{V}{d}$

$$I_D = \frac{\epsilon A}{d} \frac{dv}{dt}$$

$$= \frac{\epsilon A}{d} \frac{d(Ed)}{dt}$$

$$= \frac{\epsilon A}{d} d \frac{dE}{dt}$$

$$I_D = \epsilon A \frac{dE}{dt}$$

$$\frac{I_D}{A} = \epsilon \frac{dE}{dt}$$

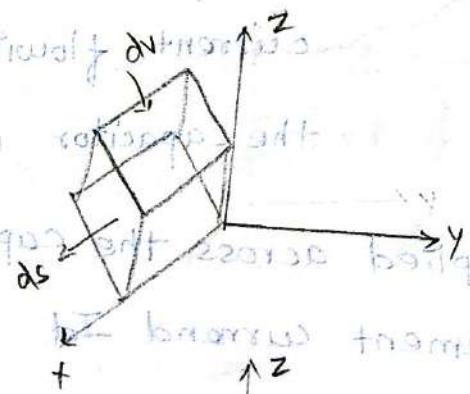
electric field flux  $\vec{D} = \epsilon \vec{E}$

$$J_D = d(\epsilon E) / dt$$

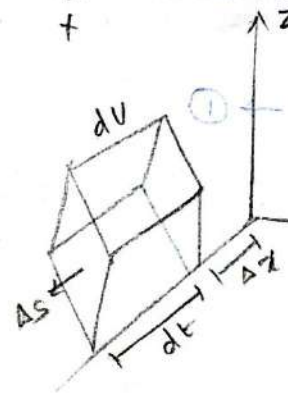
$$J_D = \frac{dD}{dt}$$

$J_D = J_{dis}$

### Relationship between current density and Volume charge density



$$\text{Potential} = \frac{x}{t}$$



$$\frac{\partial \rho}{\partial t} = \frac{\partial \rho}{\partial t} = \rho$$

$$c = \frac{\epsilon}{v}$$

$$\rho = c v = \rho$$

$$I_D = \frac{d(\rho v)}{dt}$$

## Volume charge distribution

$$\rho_v = \frac{Q}{V} = \frac{dQ}{dV}$$

$$dQ = \rho_v dV \rightarrow \textcircled{1}$$

$$I = \frac{Q}{t} = \frac{dQ}{dt} \rightarrow \textcircled{2}$$

$$\Delta V = \Delta s \cdot \Delta x \rightarrow \textcircled{3}$$

Sub ③ in ①

$$dV = \Delta V$$

$$dQ = \rho_v (\Delta s \cdot \Delta x) \quad ds = \Delta s$$

$$I = \frac{\rho_v (\Delta s \cdot \Delta x)}{dt}$$

$$I = \rho_v \Delta s \left( \frac{\Delta x}{dt} \right) \quad \frac{dx}{dt} = v_x$$

$$I = \rho_v \Delta s (v_x) \rightarrow \textcircled{4}$$

$$J = \frac{I}{A} = \frac{I}{\Delta s} = \frac{I}{\Delta s} \text{ from dia}$$

$$I = J \cdot \Delta s \rightarrow \textcircled{5}$$

Equate ④ & ⑤

$$J \cdot \Delta s = \rho_v \Delta s (v_x)$$

$$J = \rho_v v_x$$

8m \*

## Poisson's and Laplace Equations

Poisson's Eqn:-

Assume point form to Gauss law

$$\nabla \cdot D = \rho_v$$

WKT

$$D = \epsilon E$$

$$\nabla (\epsilon E) = \rho_v$$

Relationship between  $E$  and  $V$

$$E = -\nabla V$$

$$\nabla (\epsilon (-\nabla V)) = \rho V$$

$$-\epsilon (\nabla^2 V) = \rho V$$

$$\nabla^2 V = -\frac{\rho V}{\epsilon}$$

Cartesian coordinate

$$\nabla^2 V = \frac{\partial V}{\partial x} \vec{a}_x + \frac{\partial V}{\partial y} \vec{a}_y + \frac{\partial V}{\partial z} \vec{a}_z = -\frac{\rho V}{\epsilon}$$

Cylindrical coordinate

$$\nabla^2 V = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V}{\partial r} \right) + \frac{1}{r^2} \left( \frac{\partial^2 V}{\partial \phi^2} \right) + \frac{\partial^2 V}{\partial z^2} = -\frac{\rho V}{\epsilon}$$

Spherical coordinate

$$\nabla^2 V = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V}{\partial \theta} \right)$$

$$+ \frac{1}{r^2 \sin^2 \theta} \left( \frac{\partial^2 V}{\partial \phi^2} \right) = -\frac{\rho V}{\epsilon}$$

Laplace Equation:

In steady state condition there is no rate of flow of charges which means there is no current

By continuity equation

$$\nabla \cdot J = -\frac{d\rho}{dt}$$

current = 0  $\frac{d\rho}{dt} = 0$   
 $I = 0$

$$\nabla \cdot J = 0$$

$$J = \sigma E$$

$$\nabla (\sigma E) = 0$$

$$\sigma (\nabla E) = 0$$

Relationship btw  $E$  and  $V$

$$E = -\nabla V$$

$$\sigma (\nabla \cdot (-\nabla V)) = \rho$$

$$\sigma (-\nabla^2 V) = \rho$$

where

$$\sigma \neq 0 ; -\nabla^2 V = 0$$

$$\nabla^2 V = 0$$

Cartesian :-

$$\nabla^2 V = \frac{\partial^2 V}{\partial x^2} \vec{a}_x + \frac{\partial^2 V}{\partial y^2} \vec{a}_y + \frac{\partial^2 V}{\partial z^2} \vec{a}_z = 0$$

cylindrical :-

$$\nabla^2 V = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V}{\partial r} \right) + \frac{1}{r^2} \left( \frac{\partial^2 V}{\partial \phi^2} \right) +$$

spherical :-

$$\nabla^2 V = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \left( \frac{\partial^2 V}{\partial \phi^2} \right) = 0$$

Applications of Poisson's and Laplace equation

Poisson and Laplace equation are used to solve the electrostatic problem involving a set of conductors maintaining at different potential.

Both Poisson and Laplace are used to find the electric field intensity and capacitance value in different conditions.

Point form (differential form) and Integral form of Ohm's law

$$V \propto I$$

$$V = IR$$

$$R = \frac{\rho l}{A} = \frac{l}{\sigma} \cdot \frac{l}{A}$$

$$\rho = \frac{1}{\sigma}$$

$$V = \frac{I l}{\sigma A}$$

$$\frac{I}{A} = J$$

$$\frac{V}{l} = \frac{I}{\sigma A}$$

$$E = \frac{J}{\sigma}$$

$$\boxed{J = E \sigma}$$

Point form differential form of ohm's law

$$J = \frac{I}{A} = \frac{I}{s} = \frac{dI}{ds}$$

$$dI = J ds$$

$$\boxed{I = \iint J \cdot ds}$$

$$\boxed{I = \int_S J \cdot ds}$$

Integral form  
of ohm's law

### Properties of Conductor

\* The conductor surface is an equipotential surface

\* The charge density is always zero within the conductor.

\* The charge can exist on the surface of the conductor which gives rise to the surface charge density.

\* The conductivity of ideal conductor is infinite (not with respect to) Hence it is known as superconductor

\* No charges and no electric field can exist at any point within the conductor.

\* The conductivity of the material depends on the temperature

$$\sigma \propto \frac{1}{T}$$

$$J = \sigma E = \frac{J}{A}$$

For aluminium:

$$\sigma = 3.82 \times 10^{17}$$

For copper

$$\sigma = 5.8 \times 10^{27}$$

The conductors of the material which have no forbidden gap btw valence band and conduction band

In perfect conduction

$$E = 0 \text{ (inside the conductor)}$$

$$E = \infty \text{ (surface of conductor)}$$

Properties of Dielectrics:

\* Dielectrics does not contain any free electrons which all the charges are well bounded and cannot be in motion easily

\* The charges in dielectric medium are bounded by finite force Hence it is called bounded charges

\* In dielectric medium the charges are bounded there is no free electrons so they cannot contribute to the conduction process

\* Dielectrics are the material for which have large forbidden gap btw valence and conduction band



\* In perfect dielectric material the conductivity  $\sigma = 0$

\* The dielectric does not contain any current which oppose the flow of current

\* The volume charge density  $\rho_v = 0$

\* Applied the electric field the bounded charges becomes slowly breakdown

\* It produces the free charges it slowly starts to change their position. Hence dielectric store the charges.

Qm (X)

### Dielectric strength

The minimum value of applied electric field at which dielectric breakdown occurs is called dielectric strength of the material

\* Dielectric become conducting due to dielectric breakdown.

\* The electric field outside and inside the dielectrics gets modified due to the induced electric dipole.

Dielectric is classified into two types

\* Polar Dielectric

\* Non polar Dielectric

#### Polar Dielectric

In a polar type both  $(+ve)$  and  $(-ve)$  charges are separated by small distance 'd'

which act as a dipole, where exist a dipole moment  $m$  (or)  $p$ .

Non-polar:

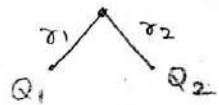
In a non-polar dielectric both (+)ve and (-)ve charges are coincides hence there is no dipole and dipole moment.

For such a material is placed in a electric field the centre of positive and negative are displacement by small distance now there exist a dipole moment.

Since non-polar dielectric material becomes polar dielectric material

Find the Force on a charge  $Q_1$  20  $\mu\text{C}$  at  $(0, 1, 2)$  m due to  $Q_2$  300  $\mu\text{C}$  at  $(2, 0, 0)$  m

$$\vec{r} = \vec{r}_2 - \vec{r}_1$$



$$\vec{r} = (2-0)\vec{a}_x + (0-1)\vec{a}_y + (0-2)\vec{a}_z$$

$$\vec{r} = 2\vec{a}_x - \vec{a}_y - 2\vec{a}_z$$

$$F = \frac{Q_1 Q_2}{4\pi\epsilon_0 r^2} \vec{a}_r$$

$$\vec{a}_r = \frac{\vec{r}}{|\vec{r}|}$$

$$\vec{a}_r = \frac{2\vec{a}_x - \vec{a}_y - 2\vec{a}_z}{\sqrt{(2)^2 + (-1)^2 + (-2)^2}}$$

$$= \frac{2\vec{a}_x - \vec{a}_y - 2\vec{a}_z}{\sqrt{9}}$$

$$= \frac{2\vec{a}_x - \vec{a}_y - 2\vec{a}_z}{3}$$

$$\vec{a}_r = \frac{2\vec{a}_x - \vec{a}_y - 2\vec{a}_z}{3}$$

$$F = \frac{(20 \times 10^{-6}) (300 \times 10^{-6})}{4\pi (8.85 \times 10^{-12}) \times (3)^2}$$

$$4\pi (8.85 \times 10^{-12}) \times (3)^2$$

$$V_2 = 31.81 \text{ volt}$$

$$\begin{aligned} V_{\text{total}} &= V_1 + V_2 \\ &= 23.135 + 31.81 \\ &= 8.675 \text{ volt} \end{aligned}$$

### Capacitance :

A capacitor is a electronic device which consist of two conductors separated by dielectric medium.

The capacitance of two conducting planes is defined as the ratio of magnitude of charge on either side of the conductor to the potential difference between conductor

$$C = \frac{Q}{V} \quad \text{unit } C/V$$

consider a capacitance composed of two conducting plates of area 'A' separated by small dielectric medium

$$C = \frac{\epsilon A}{d} \quad \text{unit is Farad.}$$

If potential  $V$  is applied across the plate, the positive charge  $Q$  is deposited, and negative charge  $Q$  is deposited on another plate since net charge equal to zero

capacitance is used to store the electric charge

# Capacitance of conductor: (parallel plate capacitance)



Due to (+)ve and (-)ve charges the charge separated by small distance where there is dipole moment

$$C = \frac{Q}{V}$$

$$D = \epsilon E \rightarrow (1)$$

$$D = \frac{Q}{S} = \frac{Q}{A} \rightarrow (2)$$

Sub eqn (2) in (1)

$$\frac{Q}{A} = \epsilon E$$

$$Q = \epsilon A E$$

Relationship between E and V

$$E = \frac{V}{d}$$

$$Q = \epsilon A \left(\frac{V}{d}\right)$$

$$\frac{Q}{V} = \frac{\epsilon A}{d}$$

$$C = \frac{\epsilon A}{d}$$

$$C = \frac{\epsilon_0 \epsilon_r A}{d}$$

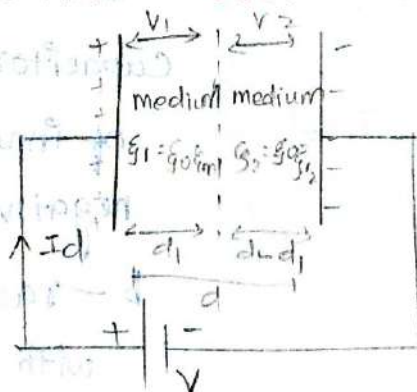
Capacitance are connected in series

$$\frac{1}{C} = \frac{1}{C_1} + \frac{1}{C_2} + \dots$$

capacitance are connected in parallel.

$$C = C_1 + C_2 + C_3 + \dots$$

## Capacitance with Two dielectric Medium



$$Q = \epsilon_1 V/A$$

$$D = \frac{Q}{A}$$

$$D_1 = \epsilon_1 E_1$$

$$D_2 = \epsilon_2 E_2$$

$$\frac{Q}{A} = \epsilon_1 E_1$$

$$\frac{Q}{A} = \epsilon_2 E_2$$

$$E_1 = \frac{Q}{\epsilon_1 A}$$

$$E_2 = \frac{Q}{\epsilon_2 A}$$

R/w btw E and V:

$$E_1 = \frac{V_1}{d_1}$$

$$E_2 = \frac{V_2}{d-d_1}$$

$$V_1 = E_1 d_1$$

$$V_2 = E_2 (d-d_1)$$

$$V_1 = \frac{E_1 d_1}{\epsilon_1 A}$$

$$V_2 = \frac{Q (d-d_1)}{\epsilon_2 A}$$

$$V = V_1 + V_2$$

$$= \frac{Q d_1}{\epsilon_1 A} + \frac{Q (d-d_1)}{\epsilon_2 A}$$

$$V = Q \left[ \frac{d_1}{\epsilon_1 A} + \frac{d-d_1}{\epsilon_2 A} \right]$$

$$C = \frac{Q}{V} = \frac{1}{\frac{d_1}{\epsilon_1 A} + \frac{d-d_1}{\epsilon_2 A}}$$

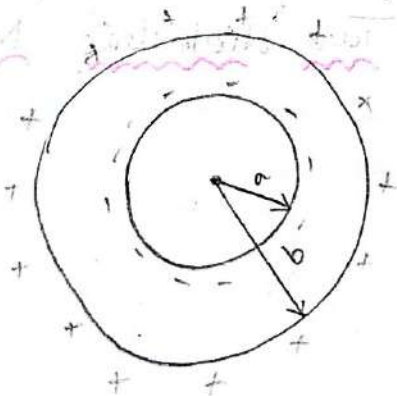
For n no. of medium

$$C = \frac{\epsilon_1 A}{d_1} + \frac{\epsilon_2 A}{d_2} + \frac{\epsilon_3 A}{d_3} + \dots$$

$$C = \frac{\epsilon_1 A}{d} + \frac{\epsilon_2 A}{(d-d_1)}$$

Capacitance on concentric sphere (or)

Capacitance on spherical



Let us assume spherical capacitor small 'a' - radius of inner spherical with negative charge

b - radius of outer spherical with (+)ve charges

Let us consider (+)ve charges and (-)ve charges are separated by small distance so it act as spherical capacitor.

$$C = \frac{Q}{V}$$

Let us assume voltage is applied to the capacitor no. of flux line are produced across the spherical capacitor

$$V = -\int \vec{E} \cdot d\vec{l} \rightarrow \textcircled{1}$$

WKT

$$\vec{E} = \frac{Q}{4\pi\epsilon_0 r^2} \vec{a}_r$$

In spherical coordinate

$$d\vec{l} = dr \vec{a}_r + r d\theta \vec{a}_\theta + r \sin\theta d\phi \vec{a}_\phi$$

$$V = -\int \frac{Q}{4\pi\epsilon_0 r^2} \vec{a}_r \cdot [dr \vec{a}_r + r d\theta \vec{a}_\theta + r \sin\theta d\phi \vec{a}_\phi]$$

$$V = \frac{-Q}{4\pi\epsilon_0} \int_b^a \frac{1}{r^2} dr$$

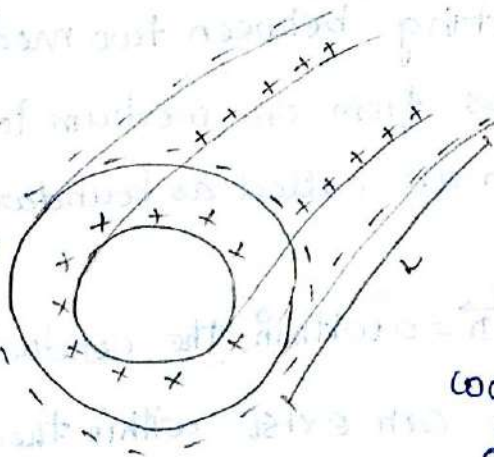
$$= \frac{-Q}{4\pi\epsilon_0} \left[ -\frac{1}{r} \right]_b^a$$

$$= \frac{Q}{4\pi\epsilon_0} \left[ \frac{1}{a} - \frac{1}{b} \right]$$

$$\frac{1}{C} = \frac{V}{Q} = \frac{1}{4\pi\epsilon_0} \left[ \frac{b-a}{ab} \right]$$

$$C = \frac{4\pi\epsilon_0 (ab)}{b-a} \quad \text{F}$$

Capacitance on coaxial cable i.e. / cylindrical capacitance



$$C = \frac{Q}{V}$$

$$V = -\int_b^a \vec{E} \cdot d\vec{l}$$

Let consider cylindrical

coordinates

$$d\vec{l} = dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z$$

For line charge distribution

$$\vec{E} = \frac{\rho l}{2\pi\epsilon_0 r} \vec{a}_r$$

$$V = -\int_b^a \frac{\rho l}{2\pi\epsilon_0 r} \vec{a}_r \cdot [dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z]$$

$$= -\frac{\rho l}{2\pi\epsilon_0} \int_b^a \frac{1}{r} dr$$

$$V = -\frac{\rho l}{2\pi\epsilon_0} [\log r]_b^a$$

$$= -\frac{\rho l}{2\pi\epsilon_0} [\log a - \log b]$$

$$= \frac{\rho l}{2\pi\epsilon_0} [\log b - \log a]$$

$$V = \frac{\rho l}{2\pi\epsilon_0} \left( \log \frac{b}{a} \right)$$

WKT  $C = \frac{Q}{V}$

$$Q = \rho l$$

$$\frac{Q}{\rho l} = C$$

$$C = \frac{\rho l}{\frac{\rho l}{2\pi\epsilon_0} \log(b/a)}$$

$$C = \frac{2\pi\epsilon_0 l}{\log(b/a)}$$

$$C = \frac{2\pi\epsilon_0 l}{\log(b/a)}$$

13m

### Boundary Condition:

The condition existing between two media when Electric field  $\vec{E}$  passes from one medium to another medium such condition are called as boundary condition.

(i)  $\vec{E} = 0$  and  $\vec{D} = 0$  within the conductor

(ii) No charge can exist within the conductor

(iii) The charges are appeared on the surface

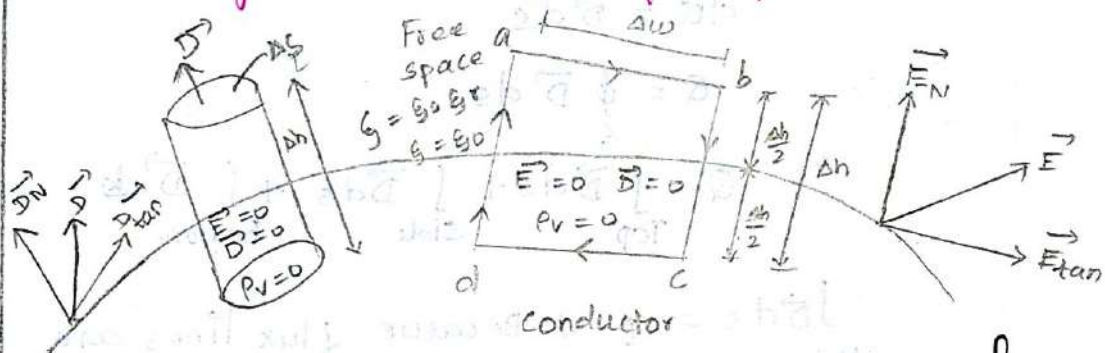
of the conductor in the form of surface charge distribution  $\rho_s$

→ volume charge density  $\rho_v = 0$  within the conductor.

Classification:-

- The boundary condition btw freespace and conductor
- The boundary condition btw dielectric and conductor
- The boundary condition btw perfect dielectric material (or) two dielectric material ( $\epsilon_1$  and  $\epsilon_2$ )

**Boundary Condition btw freespace & conductor:-**



$$\oint \vec{E} \cdot d\vec{l} = 0$$

$$\oint \vec{E} \cdot d\vec{l} = \rho_v \Delta w = 0 \quad (\because \rho_v = 0)$$

$$\vec{E} = \vec{E}_N + \vec{E}_{tan}$$

$$\int d\vec{l} = \Delta w$$

$$\int_a^b \vec{E}_{tan} \cdot d\vec{l} + \int_b^c \vec{E} \cdot d\vec{l} + \int_c^d \vec{E} \cdot d\vec{l} + \int_d^a \vec{E} \cdot d\vec{l} = 0$$

$$\int_a^b \vec{E}_{tan} \Delta w + \int_b^c \vec{E}_N \frac{\Delta h}{2} - \int_c^d (0) \frac{\Delta h}{2} + 0 +$$

at conductor  
⇒ zero

$$\int_d^a (0) \frac{\Delta h}{2} + \int_a^b \vec{E}_N \frac{\Delta h}{2} = 0$$

$$\int_a^b \vec{E}_{tan} \Delta w = 0$$

$$\vec{E}_{tan} \Delta w = 0$$

$$\vec{E}_{tan} \Delta w \neq 0$$

$$\Delta w \neq 0$$

$$\vec{E}_{tan} = 0$$

Relationship btw E and R

$$\vec{D} = \epsilon \vec{E}$$



$$\vec{D}_{\text{tan}} = \epsilon_0 \vec{E}_{\text{tan}}$$

$$\vec{E}_{\text{tan}} = \frac{\vec{D}_{\text{tan}}}{\epsilon_0} = 0$$

$$\epsilon_0 \neq 0$$

$$\boxed{\vec{D}_{\text{tan}} = 0}$$

Let us consider cylindrical Gaussian surface

$$\vec{D} = \frac{Q}{s} = \frac{q}{s} = \frac{dq}{ds}$$

$$dq = \vec{D} ds$$

$$Q = \oint \vec{D} ds$$

$$Q = \int_{\text{Top}} \vec{D} ds + \int_{\text{side}} \vec{D} ds + \int_{\text{Bottom}} \vec{D} ds$$

$$\int_{\text{Bot}} \vec{D} ds = 0 ; \text{ Because flux lines are}$$

within the conductor = 0

The flux lines leaving from lateral <sup>side</sup> surface = 0

$$\Delta h \ll \Delta s$$

$$\Delta h \rightarrow 0$$

$$\int_{\text{side}} \vec{D} \frac{\Delta h}{2} = 0$$

$$Q = \int_{\text{Top}} \vec{D} ds + 0 + 0$$

$$Q = \int_{\text{Top}} \vec{D} \cdot d\vec{s}$$

$$Q = D_N \cdot S$$

$$\vec{D}_n = D_N$$

$$\int dx = S$$

Surface charge distribution

$$\rho_s = \frac{Q}{S}$$

$$Q = S \rho_s$$

$$P_s \cdot s = D_N \cdot s$$

$$P_s = D_N$$

$$D_N = \epsilon E_N$$

$$D_N = \epsilon_0 E_N \text{ in free space}$$

$$E_N = \frac{D_N}{\epsilon_0}$$

$$E_N = \frac{P_s}{\epsilon_0}$$

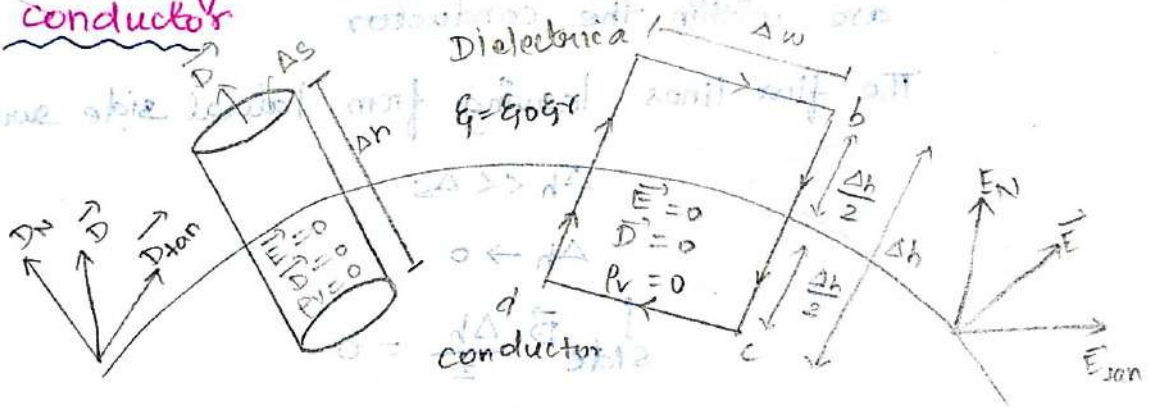
$$1. E_{tan} = 0$$

$$2. D_{tan} = 0$$

$$3. E_N = P_s / \epsilon_0$$

$$4. D_N = P_s$$

### Boundary condition between Dielectric and conductor



$$\oint \vec{E} \cdot d\vec{l} = 0$$

$$\int dl = \Delta w$$

$$\vec{E} = \vec{E}_N + \vec{E}_{tan}$$

$$\int_a^b \vec{E} \cdot d\vec{l} + \int_b^c \vec{E} \cdot d\vec{l} + \int_c^d \vec{E} \cdot d\vec{l} + \int_d^a \vec{E} \cdot d\vec{l} = 0$$

$$E_{tan} \Delta w - E_N \frac{\Delta h}{2} - (0) \frac{\Delta h}{2} + 0 + 0 + E_N \frac{\Delta h}{2} = 0$$

$$E_{tan} \Delta w = 0$$

$$E_{tan} \Delta w = 0$$

$$\Delta w \neq 0$$

$$E_{tan} = 0$$

Relationship btw  $\vec{E}$  and  $\vec{R}$

$$D = \epsilon E$$

$$D_{\text{tan}} = \epsilon E$$

$$\vec{E} = \frac{D_{\text{tan}}}{\epsilon} = 0$$

$$\epsilon \neq 0$$

$$\boxed{D_{\text{tan}} = 0}$$

Let us consider cylindrical Gaussian surface

$$\vec{D} = \frac{Q}{s} = \frac{q}{s} = \frac{dq}{ds}$$

$$dq = \vec{D} ds$$

$$Q = \int_{\text{Top}} \vec{D} ds + \int_{\text{side}} \vec{D} ds + \int_{\text{Bot}} \vec{D} ds$$

$\int_{\text{Bot}} \vec{D} ds = 0$  Because flux lines are within the conductor

The flux lines leaving from lateral side surface  $\Rightarrow$

$$\Delta h \ll \Delta s$$

$$\Delta h \rightarrow 0$$

$$\int_{\text{side}} \vec{D} \frac{\Delta h}{2} = 0$$

$$Q = \int_{\text{Top}} \vec{D} ds + 0 + 0$$

$$Q = \int_{\text{Top}} D_x dA$$

$$Q = D_N \cdot s$$

Surface charge distribution

$$P_s = \frac{Q}{s}$$

$$Q = s P_s$$

$$P_s \cdot s = D_N \cdot s$$

$$P_s = D_N$$

$$D_N \vec{=} \epsilon \vec{E}_N$$

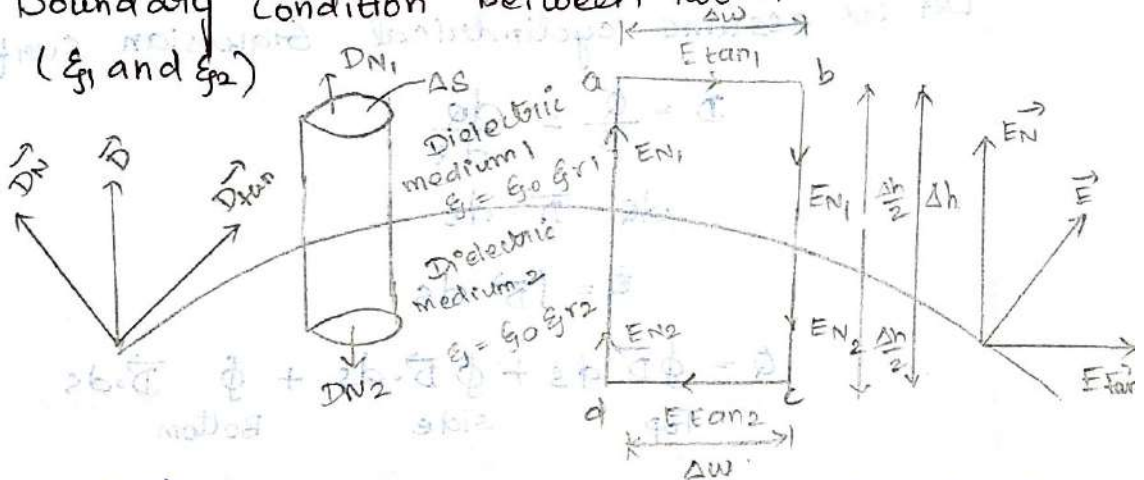
$$D_N \vec{=} \epsilon \vec{E}_N$$

$$\vec{E}_N = \frac{D_N}{\epsilon}$$

$$\vec{E}_N = \frac{P_s}{\epsilon}$$

$$\begin{aligned} \vec{E}_{tan} &= 0 \\ D_{tan} &= 0 \\ \vec{E}_N &= \frac{P_s}{\epsilon} \\ D_N &= P_s \end{aligned}$$

Boundary Condition between two dielectric medium ( $\epsilon_1$  and  $\epsilon_2$ )



$\oint \vec{E} \cdot d\vec{l} = 0$  which means this closed contour is zero. workdone in carrying a unit positive charge along a closed path is zero

$$\oint \vec{E} \cdot d\vec{l} = 0$$

$$\int_a^b \vec{E} \cdot d\vec{l} + \int_b^c \vec{E} \cdot d\vec{l} + \int_c^d \vec{E} \cdot d\vec{l} + \int_d^a \vec{E} \cdot d\vec{l} = 0$$

$$\vec{E}_{tan1} \Delta w - E_{N1} \frac{\Delta h}{2} - E_{N2} \frac{\Delta h}{2} - E_{tan2} \Delta w + E_{N2} \frac{\Delta h}{2} + E_{N1} \frac{\Delta h}{2} = 0$$

$$\vec{E}_{tan1} \Delta w - E_{tan2} \Delta w = 0$$

$$\vec{E}_{tan1} \Delta w - E_{tan2} \Delta w = 0$$

$$(\vec{E}_{tan1} - E_{tan2}) \Delta w = 0$$

$$\Delta w \neq 0$$

$$\vec{E}_{tan1} = \vec{E}_{tan2}$$

$$\vec{D}_{tan1} = \epsilon_1 \vec{E}_{tan1}$$

$$\vec{D}_{tan2} = \epsilon_2 \vec{E}_{tan2}$$

$$\frac{\vec{D}_{tan1}}{\vec{D}_{tan2}} = \frac{\epsilon_1 \vec{E}_{tan1}}{\epsilon_2 \vec{E}_{tan2}} \quad E_{tan1} = E_{tan2}$$

$$\boxed{\frac{\vec{D}_{tan1}}{\vec{D}_{tan2}} = \frac{\epsilon_1}{\epsilon_2}}$$

Let us assume cylindrical Gaussian surface

$$D = \frac{Q}{S} = \frac{dq}{dr}$$

$$dq = \vec{D} \cdot d\vec{s}$$

$$Q = \int \vec{D} \cdot d\vec{s}$$

$$Q = \oint_{\text{Top}} \vec{D} \cdot d\vec{s} + \oint_{\text{side}} \vec{D} \cdot d\vec{s} + \oint_{\text{Bottom}} \vec{D} \cdot d\vec{s}$$

The flux leaving from the lateral surface = 0.

$$\Delta h \ll \Delta s$$

$$\Delta h \rightarrow 0$$

$$Q = \int_{\text{Top}} \vec{D} \cdot d\vec{s} + \int_{\text{Bot}} \vec{D} \cdot d\vec{s}$$

$$Q = D_{N1} \Delta s + D_{N2} \Delta s$$

$$Q = \Delta s (D_{N1} - D_{N2})$$

$$Q = \int (D_{N1} - D_{N2}) \Delta s$$

$$P_s = \frac{Q}{S}$$

$$Q = P_s \cdot S$$

$$P_s \cdot S = S (D_{N1} - D_{N2})$$

when  $P_s = 0$  because in a dielectric material only bounded charges are present

$$\vec{D}_{N1} = \vec{D}_{N2}$$

$$\vec{D}_{N1} = \epsilon_1 \vec{E}_{N1}$$

$$\vec{D}_{N2} = \epsilon_2 \vec{E}_{N2}$$

$$\frac{E_{N1}}{E_{N2}} = \frac{D_{N1}/\epsilon_1}{D_{N2}/\epsilon_2}$$

$$\vec{D}_{N1} = \vec{D}_{N2}$$


$$\frac{E_{N1}}{E_{N2}} = \frac{\epsilon_2}{\epsilon_1}$$

$$\vec{E}_{tan1} = \vec{E}_{tan2}$$

$$\frac{D_{tan1}}{D_{tan2}} = \frac{\epsilon_1}{\epsilon_2}$$

$$\vec{D}_{N1} = \vec{D}_{N2}$$

$$\frac{E_{N1}}{E_{N2}} = \frac{\epsilon_2}{\epsilon_1}$$

6m 

### Energy density in Electric static field:

When a unit positive charge is moved from infinite to a point in a field, the work is done by External source and the energy is expended.

This energy gets to store in the form of potential Energy. (it means Electrostatic energy)

When external source removed, the potential Energy gets converted into Kinetic Energy

$$W_E = \frac{1}{2} \sum_{m=1}^n Q_m V_m \quad \text{Joule}$$

For line charge  $P_L$

$$W_E = \frac{1}{2} \int P_L dl \cdot V$$

For surface charge  $\rho_s$ ,

$$W_s = \frac{1}{2} \int_S \rho_s ds \cdot V$$

For volume charge  $\rho_v$ ,

$$W_E = \frac{1}{2} \int_V \rho_v dv \cdot V$$

Energy stored in terms of  $\vec{E}$  and  $\vec{D}$

Volume charge distribution along charge density

$\rho_v$  in  $C/m^3$

$$W_E = \frac{1}{2} \iiint_V \rho_v \cdot dv \cdot V \text{ Joule}$$

Gauss divergence theorem  $\nabla \cdot \vec{D} = \rho_v$

$$W_E = \frac{1}{2} \iiint_V (\nabla \cdot \vec{D}) dv \cdot V$$

( $r \rightarrow$  infinite limit)

$$= \frac{1}{2} \iiint_V \vec{D} \cdot (-\nabla v) dv$$

$$W_E = \frac{1}{2} \iiint_V \vec{D} \cdot \vec{E} dv \text{ in Joule} \rightarrow \textcircled{1}$$

$$W_E = \frac{1}{2} \iiint_V \frac{D^2}{\epsilon_0} dv$$

$$D = \epsilon_0 E$$

$$E = \frac{D}{\epsilon_0}$$

$$W_E = \frac{1}{2} \iiint_V \epsilon_0 E^2 dv$$

$$D^2 = \epsilon_0^2 E^2$$

$$\text{Diff } \textcircled{1} \quad dW_E = \frac{1}{2} \vec{D} \cdot \vec{E} dv$$

$$D^2 = \epsilon_0 E^2$$

$$\text{Energy density } \frac{dW_E}{dv} = \frac{1}{2} \vec{D} \cdot \vec{E} \text{ J/m}^3$$

Energy stored in terms of capacitance:

The capacitance store the electrostatic

Energy is equal to workdone to build up the charges

If a voltage is connected across the capacitor, the capacitor charges

Potential difference is defined as workdone per unit charge

$$V = \frac{W}{Q} = \frac{dW}{dQ}$$

$$dW = V \cdot dQ$$

$$W = \int V \cdot dQ$$

$$V = \frac{Q}{C} \quad Q = CV$$

$$W = \int \frac{Q}{C} dQ$$

$$= \frac{1}{C} \int Q dQ$$

$$= \frac{1}{C} \left[ \frac{Q^2}{2} \right]$$

$$= \frac{1}{2} \frac{1}{C} [C^2 V^2]$$

$$W = \frac{1}{2} C V^2$$

### Polarization :-

polarization is defined as dipole moment

Per unit Volume

Polarization increases the electric flux density

$$\vec{P} = \lim_{\Delta V \rightarrow 0} \frac{\sum_{i=1}^n Q_i d\vec{r}_i}{\Delta V} \quad \text{C/m}^2$$

$$\vec{D} = \epsilon_0 \vec{E} \quad \text{in free space}$$

$$\vec{D} = \epsilon_0 \vec{E} + \vec{P}$$

$$\text{Polarization } \vec{P} = \chi_e \epsilon_0 \vec{E}$$

$$\vec{D} = \epsilon_0 \vec{E} + \chi_e \epsilon_0 \vec{E}$$

$$\vec{D} = \epsilon_0 \vec{E} (1 + \chi_e) \rightarrow \textcircled{1}$$

$$\vec{D} = \epsilon_0 \epsilon_r \vec{E} \rightarrow \textcircled{2}$$

(compare ① & ②)

$$\epsilon_r = 1 + \chi_e$$

$$\chi_e = \epsilon_r - 1$$

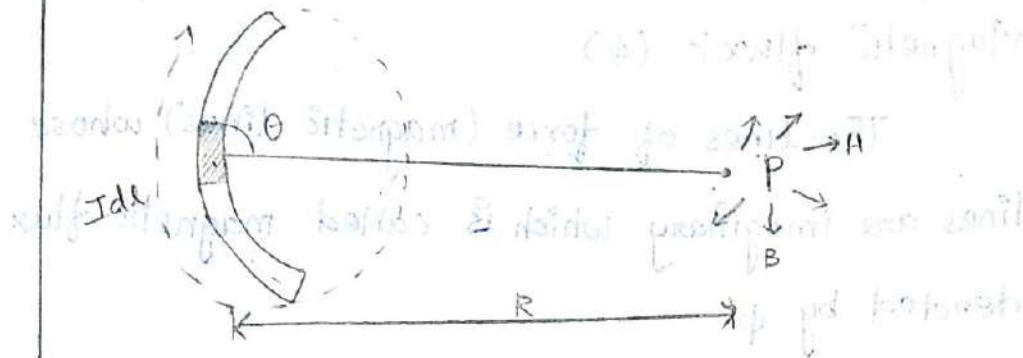
$\chi_e =$  susceptibility in dielectric medium



# UNIT-3

# Magnetostatics

## Bio-Savart Law:-



According to Bio-Savart law, the magnetic field intensity produces due to current carrying a conductor

The no. of charges moving from one end to another end of conductor which gives rise to current.

This current carrying conductor produce the magnetic field. Such a magnetic field is called steady magnetic field.

### Statement

The magnetic field intensity  $H$  is directly proportional to product of the current  $I$ , differential length  $dl$  and  $\sin$  of angle btw line joining point  $P$  to the current element and inversely proportional to the square of distance btw the point  $P$  to differential length  $dl$ . is expressed as

$$dH \propto \frac{I dl \sin \theta}{R^2}$$

$$dH = \frac{k I dl \sin \theta}{R^2}$$

$$K = \frac{1}{4\pi}$$

$$dH = \frac{I dl \sin \theta}{4\pi R^2} \quad \text{Unit is } A/m$$

Magnetic flux:- ( $\phi$ )

The lines of force (magnetic lines) whose lines are imaginary which is called magnetic flux denoted by  $\phi$

Unit is weber.

Magnetic Field Intensity:-

Magnetic strength or weakness of flux lines in terms of number of flux lines are produced due to the current is known as Magnetic Field Intensity

It is denoted as 'H'

Unit is A/m.

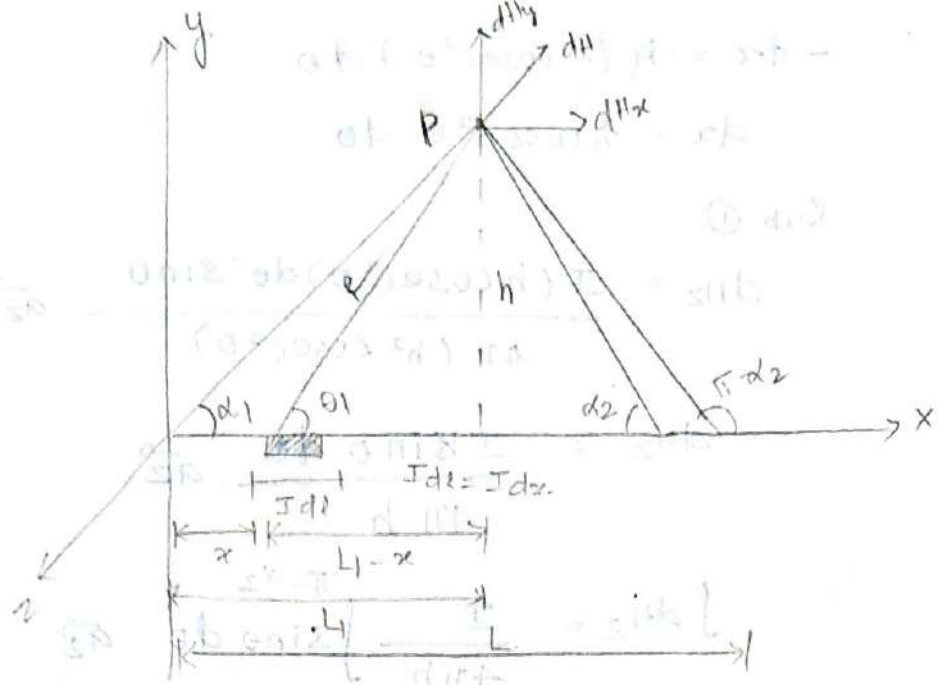
Magnetic Flux Density:-

It is defined as magnetic flux is passing through unit area denoted by the

$$B = \frac{\text{no. of flux lines}}{\text{Unit area}} = \frac{\phi}{A} \quad \text{Unit is } W/m^2$$

**Estimation of magnetic field Intensity and Density.**

(i) To find the magnetic field intensity and density at a point 'P' in current carrying conductor for finite line and infinite line.



According to Bio-savart law

$$dH = \frac{Idl \sin \theta}{4\pi R^2}$$

$$Idl = Idx$$

$$= \frac{I dx \sin \theta}{4\pi R^2}$$

$$\vec{dH} = dH_x + dH_y + dH_z$$

Let us assume the magnetic flux are moving on z direction

$$dH_z = \frac{I dx \sin \theta}{4\pi R^2} \hat{a}_z \rightarrow \textcircled{1}$$

Case (i)

H for finite line conductance

$$\sin \theta = \frac{h}{R}$$



$$R = \frac{h}{\sin \theta}$$

$$\boxed{R = h \operatorname{cosec} \theta}$$

$$\tan \theta = \frac{h}{L-x}$$

$$L-x = \frac{h}{\tan \theta} \Rightarrow h \cot \theta$$

$$-dx = h (-\operatorname{cosec}^2 \theta) d\theta$$

$$dx = h \operatorname{cosec}^2 \theta d\theta$$

Sub ①

$$dH_z = \frac{I (h \operatorname{cosec}^2 \theta) d\theta \sin \theta}{4\pi (h^2 \operatorname{cosec}^2 \theta)} \vec{a}_z$$

$$dH_z = \frac{I \sin \theta d\theta}{4\pi h} \vec{a}_z$$

$$\int dH_z = \frac{I}{4\pi h} \int_{\alpha_1}^{\pi - \alpha_2} \sin \theta d\theta \vec{a}_z$$

$$H_z = \frac{I}{4\pi h} [-\cos \theta]_{\alpha_1}^{\pi - \alpha_2}$$

$$\cos(\pi - \theta) = -\cos \theta = \frac{I}{4\pi h} [-\cos(\pi - \alpha_2) + \cos \alpha_1]$$

$$= \frac{I}{4\pi h} [-(-\cos \alpha_2) + \cos \alpha_1] \vec{a}_z$$

$$H_z = \frac{I}{4\pi h} [\cos \alpha_1 + \cos \alpha_2] \vec{a}_z \text{ A/m}$$

$$B_z = \mu H_z$$

$$\text{①} \leftarrow B_z = \frac{\mu I}{4\pi h} [\cos \alpha_1 + \cos \alpha_2] \vec{a}_z \text{ Wb/m}^2$$

Case (ii)

H infinite line conductance

For infinite line conductance

$$\alpha_1 = 0 \quad \alpha_2 = 0$$

$$\cos \alpha_1 = \cos \alpha_2 = \cos 0 = 1$$

$$H_z = \frac{I}{4\pi h} [\cos \alpha_1 + \cos \alpha_2] \vec{a}_z$$

$$= \frac{2I}{4\pi h} [\cos 0 + \cos 0] \vec{a}_z$$

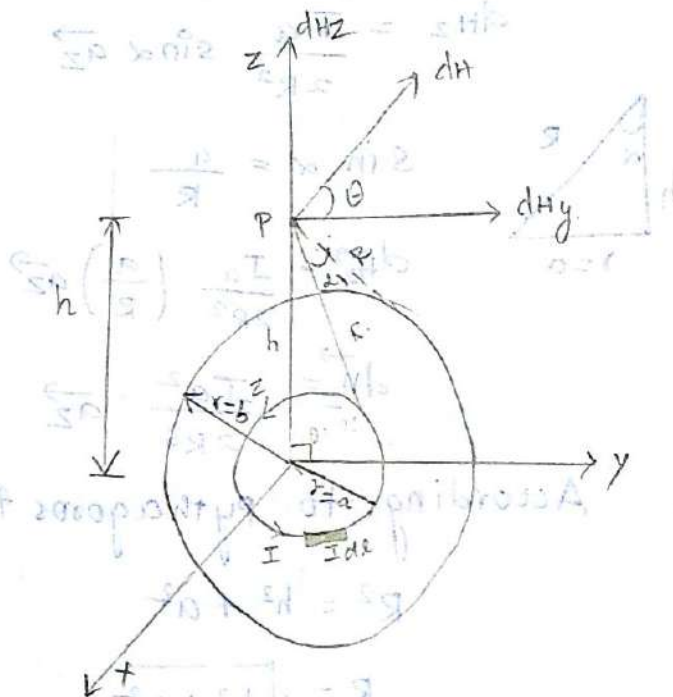
$$= \frac{I}{4\pi h} [1+1] \vec{a}_z$$

$$= \frac{I}{4\pi h} [2] \vec{a}_z \quad \text{A/m}$$

$$H_z = \frac{I}{2\pi h} \vec{a}_z \quad \text{A/m}$$

$$B_z = \frac{\mu I}{2\pi h} \vec{a}_z \quad \text{W/m}^2$$

(ii) To find Magnetic Field Intensity and density for circular conductor.



According to Bio-savart law

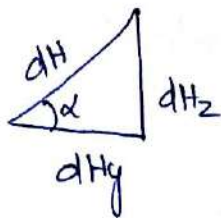
$$dH = \frac{I dl \sin \theta}{4\pi R^2}$$

The magnetic field is produced on the axis  $\theta = 90^\circ$

$$\sin \theta = \sin 90^\circ = 1$$

$$dH = \frac{I dl}{4\pi R^2}$$

Here the point P is on z axis which is right angle to the circular loop ( $\theta = 90^\circ$ )



$$\sin \alpha = \frac{dH_z}{dH}$$

$$dH_z = dH \sin \alpha$$

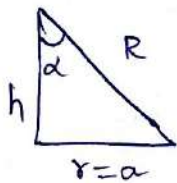
$$dH_z = \frac{I dl}{4\pi R^2} \sin \alpha \vec{a}_z$$

Circumference  $\rightarrow dl = 2\pi r$

$r = a$  in inner circular disc

$$d\vec{H}_z = \frac{I (2\pi a)}{4\pi R^2} \sin \alpha \vec{a}_z$$

$$d\vec{H}_z = \frac{I a}{2R^2} \sin \alpha \vec{a}_z$$



$$\sin \alpha = \frac{a}{R}$$

$$d\vec{H}_z = \frac{I a}{2R^2} \left(\frac{a}{R}\right) \vec{a}_z$$

$$d\vec{H}_z = \frac{I a^2}{2R^3} \vec{a}_z$$

According to Pythagoras theorem

$$R^2 = h^2 + a^2$$

$$R = \sqrt{h^2 + a^2}$$

$$R^3 = (h^2 + a^2)^{3/2}$$

$$d\vec{H}_z = \frac{I a^2}{2(h^2 + a^2)^{3/2}} \vec{a}_z$$

When  $h=0$  according to equipotential surface from ground level to the freespace because the magnetic field flux lines are produced at freespace  $H=0$

$$d\vec{H}_z = \frac{I a^2}{2(a^2)^{3/2}} \vec{a}_z$$

$$= \frac{I a^2}{2 a^3} \vec{a}_z$$

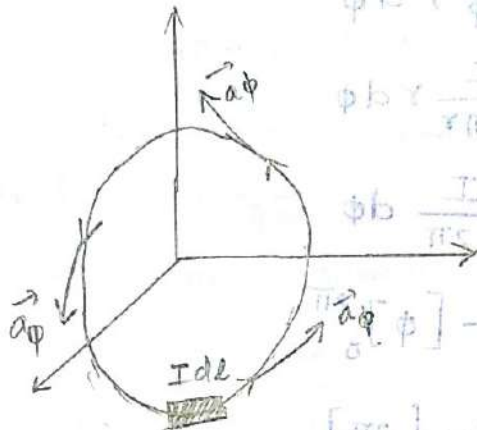
$$d\vec{H}_z = \frac{I}{2a} \vec{a}_z$$

Magnetic flux density

$$B = \mu H$$

$$B_z = \frac{\mu I}{2a} \vec{a}_z \quad \text{W/m}^2$$

(\*) Ampere's Circuital Law:-



$$\oint \vec{H} \cdot d\vec{l} = I_{\text{enclosed}}$$

The line integral of magnetic field intensity produced by an closed path is equal to the current enclosed by that path.

Proof:-

Let us consider cylindrical coordinate system according to ampere circuital law, the current carrying conductor which produces magnetic flux lines.

Assume magnetic flux lines are produced in  $\phi$  direction

$$d\vec{l} = dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z$$

$$\oint \vec{H} \cdot d\vec{l} = I = 0$$

$$I = \oint H_\phi d\phi [dr \vec{a}_r + r d\phi \vec{a}_\phi + dz \vec{a}_z]$$

$$I = \oint H_{\phi} r d\phi$$

$$= H_{\phi} r \int_0^{2\pi} d\phi$$

$$= H_{\phi} r [\phi]_0^{2\pi}$$

$$I = H_{\phi} r [2\pi]$$

$$H_{\phi} = \frac{I}{2\pi r}$$

R.H.S

$$= \oint H_{\phi} r d\phi$$

$$= \oint \frac{I}{2\pi r} r d\phi$$

$$= \oint \frac{I}{2\pi} d\phi$$

$$= \frac{I}{2\pi} [\phi]_0^{2\pi}$$

$$= \frac{I}{2\pi} [2\pi]$$

$$= I$$

L.H.S = R.H.S  
Hence proved.

Point form / Differential form of Ampere's Circuital Law

In term of magnetic field intensity,  $H$

$$\oint \vec{H} \cdot d\vec{l} = I_{\text{enclosed}}$$

According to Stokes's theorem

$$\oint \vec{H} \cdot d\vec{l} = \iint (\nabla \times \vec{H}) \cdot d\vec{s} \rightarrow \textcircled{1}$$

$$I = \iint (\nabla \times \vec{H}) \cdot d\vec{s} \rightarrow \textcircled{2}$$

$$+ \vec{p}_1 \cdot \vec{r}_1 + \vec{p}_2 \cdot \vec{r}_2 = \vec{p} \cdot \vec{r}$$

$$J = \frac{I}{A} = \frac{I}{s} = \frac{dI}{ds}$$

$$dI = J \cdot ds$$



$$I = \iint \vec{J} \cdot d\vec{s}$$

Sub (2)

$$\iint \vec{J} \cdot d\vec{s} = \iint (\nabla \times \vec{H}) \cdot d\vec{s}$$

$$\boxed{\vec{J} = \nabla \times \vec{H}}$$

In term of magnetic flux density  $B$

$$B = \frac{\phi}{A} = \frac{d\phi}{ds}$$

$$d\phi = B \cdot ds$$

$$\phi = \iint \vec{B} \cdot d\vec{s}$$

There is no circuit there is an isolated electric charge

Hence there is no magnetic field

$$\phi = \iint \vec{B} \cdot d\vec{s} = 0$$

According divergence theorem

$$\iint \vec{B} \cdot d\vec{s} = \iiint (\nabla \cdot \vec{B}) \cdot dV = 0$$

$$\boxed{\nabla \cdot \vec{B} = 0}$$

Application of ampere circuital law :-

(1) Line :-

To find magnetic field intensity for

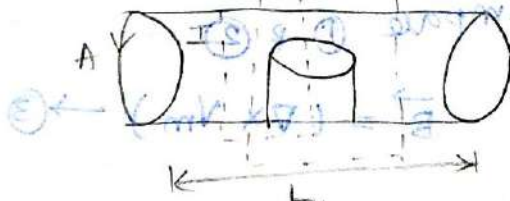
For finite line

$$\vec{H} = \frac{I}{4\pi h} [\cos \alpha_1 + \cos \alpha_2]$$

For infinite line

$$\vec{H} = \frac{I}{2\pi h}$$

(2) Solenoidal



To find magnetic field intensity

According to Ampere's Law

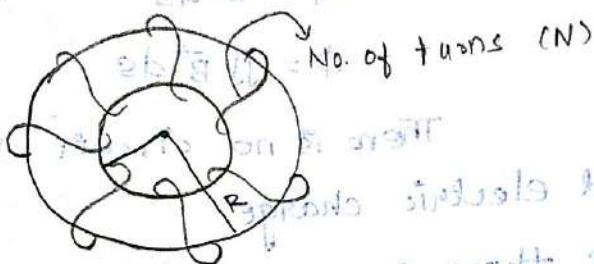
$$\oint \vec{H} \cdot d\vec{l} = I_{\text{enclosed}} \quad (1)$$

$$H \cdot L = I$$

$$H \cdot L = NI \quad \Rightarrow \quad H = \frac{NI}{L}$$

$$\boxed{\vec{H} = \frac{NI}{L}}$$

(3) Toroid ::



$$\oint \vec{H} \cdot d\vec{l} = I_{\text{enclosed}}$$

$$\vec{H} (2\pi R_m) = I$$

$$R_m = \frac{R_1 + R_2}{2} \quad \text{mean Radius}$$

$$\vec{H} (2\pi R_m) = NI$$

$$\vec{H} = \frac{NI}{2\pi R_m}$$

$$\nabla \times (\text{curl}) = 0$$

$$\nabla \cdot (\text{div}) = 0$$

$$\vec{A} = -\nabla V_m$$

Scalar and Vector Magnetic potential :-

Vectors Magnetic potential (or) poissens eqn

$$\nabla \cdot (\nabla \times V_m) = 0 \quad \rightarrow (1)$$

$$\nabla \cdot (\nabla \times \vec{H}) = 0$$

Pt from ampere's interm of 'B'

$$\nabla \cdot \vec{B} = 0 \quad \rightarrow (2)$$

Compare (1) & (2)

$$\vec{B} = (\nabla \times V_m) \quad \rightarrow (3)$$

Interm of H,  $\nabla \times \vec{H} = \vec{J} \quad \rightarrow (4)$

Relationship btw B and H

$$B = \mu H \quad \text{or} \quad H = \frac{B}{\mu}$$

$$H = \frac{B}{\mu}$$

$$\nabla \times \left( \frac{B}{\mu} \right) = J$$

$$\nabla \times B = \mu J \rightarrow \textcircled{5}$$

compare eqn ③ and ⑤

$$\nabla \times (\nabla \times V_m) = \mu J$$

Formula:  $\nabla \times (\nabla \times A) = (\nabla \cdot A) \cdot \nabla - (\nabla \cdot \nabla) A$

$$(\nabla \times V_m) \cdot \nabla - (\nabla \cdot \nabla) V_m = \mu J$$

$$(\nabla \cdot \nabla) V_m = \mu J$$

$$\nabla^2 V_m = \mu J$$

For steady state condition due to isolated electric charge

$$\vec{B} = \nabla \times V_m = 0$$

$$\nabla^2 V_m = \mu J$$

$$\nabla^2 V_m = -\mu J$$

Cartesian  $\nabla^2 V_m = \frac{\partial^2 V_m}{\partial x^2} + \frac{\partial^2 V_m}{\partial y^2} + \frac{\partial^2 V_m}{\partial z^2} = -\mu J$

Cylindrical  $\nabla^2 V_m = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2} \left( \frac{\partial^2 V_m}{\partial \phi^2} \right) + \frac{\partial^2 V_m}{\partial z^2} = -\mu J$

Spherical  $\nabla^2 V_m = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V_m}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \left( \frac{\partial^2 V_m}{\partial \phi^2} \right) = -\mu J$

Scalar Magnetic potential (or) Laplace eqn.

$$\nabla \times (\nabla V_m) = 0 \rightarrow \textcircled{6}$$

$$\nabla \cdot (\nabla \times A) = 0$$

R/w btw H and Vm

$$H = -\nabla V_m$$

Sub

$$\nabla \times (-\vec{H}) = 0 \quad \text{if } \mu = \mu_0$$

$$\nabla \times \vec{H} = 0 \quad \text{if } \mu = \mu_0$$

μ form of ampere's law  $\nabla \times \vec{H} = \vec{J}$

$$\nabla \times \vec{H} = \vec{J}$$

Current density  $\vec{J} = 0$

For an isolated electric charge

$$\nabla \cdot \vec{B} = 0$$

$$\nabla \cdot \vec{H} = \nabla \cdot (\mu \vec{H}) = \nabla \cdot (\mu \nabla \times \vec{V}_m) = 0$$

$$\nabla \cdot \vec{H} = \nabla \cdot (-\nabla V_m) = 0$$

$$\nabla \cdot (\mu \vec{H}) = 0$$

$$\mu (\nabla \cdot (-\nabla V_m)) = 0$$

$$\mu (\nabla \cdot \vec{H}) = 0$$

$$0 = \mu (\nabla \cdot (-\nabla V_m)) = 0$$

$$\nabla \cdot (\mu \nabla V_m) = 0$$

$$\nabla \cdot \vec{H} = \nabla \cdot (\mu \nabla V_m) = 0$$

$$-\nabla^2 V_m = 0$$

$$\nabla^2 V_m = 0$$

Cartesian:  $\nabla^2 V_m = \frac{\partial^2 V_m}{\partial x^2} + \frac{\partial^2 V_m}{\partial y^2} + \frac{\partial^2 V_m}{\partial z^2} = 0$

$$\nabla^2 V_m = \frac{\partial^2 V_m}{\partial x^2} + \frac{\partial^2 V_m}{\partial y^2} + \frac{\partial^2 V_m}{\partial z^2} = 0$$

Cylindrical:  $\nabla^2 V_m = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 V_m}{\partial \phi^2} + \frac{\partial^2 V_m}{\partial z^2} = 0$

$$\nabla^2 V_m = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2} \frac{\partial^2 V_m}{\partial \phi^2} + \frac{\partial^2 V_m}{\partial z^2} = 0$$

Spherical:  $\nabla^2 V_m = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V_m}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 V_m}{\partial \phi^2} = 0$

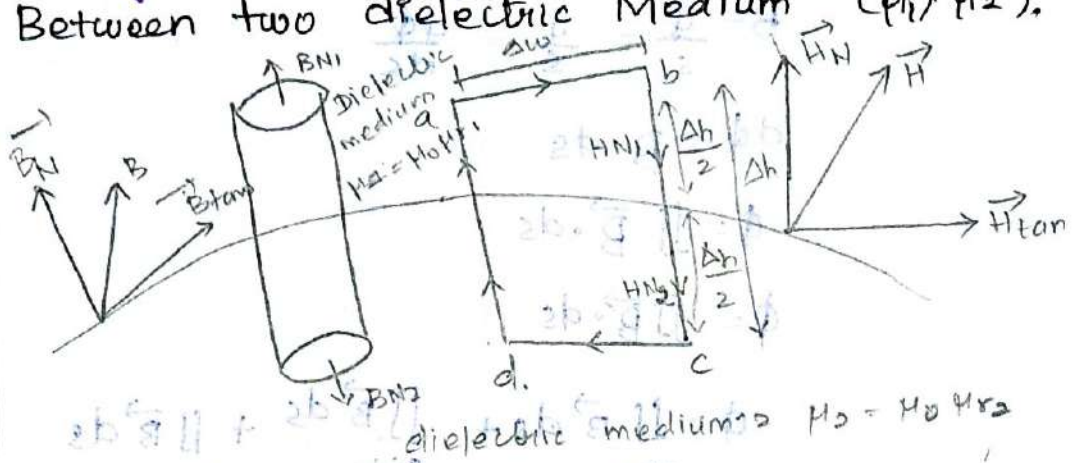
$$\nabla^2 V_m = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V_m}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 V_m}{\partial \phi^2} = 0$$

$$\nabla^2 V_m = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V_m}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 V_m}{\partial \phi^2} = 0$$

$$\nabla^2 V_m = \frac{1}{r^2} \frac{\partial}{\partial r} \left( r^2 \frac{\partial V_m}{\partial r} \right) + \frac{1}{r^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V_m}{\partial \theta} \right) + \frac{1}{r^2 \sin^2 \theta} \frac{\partial^2 V_m}{\partial \phi^2} = 0$$

# Magnetic Boundary Condition

Between two dielectric medium ( $\mu_1, \mu_2$ ).



According to ampere circuital law

$$\oint \vec{H} \cdot d\vec{l} = I_{\text{enclosed}}$$

$$\oint_A \vec{H} \cdot d\vec{l} + \int_B \vec{H} \cdot d\vec{l} + \int_C \vec{H} \cdot d\vec{l} + \int_D \vec{H} \cdot d\vec{l} = I$$

$$H_{\tan 1} \Delta w + (-HN_1) \frac{\Delta h}{2} - HN_2 \frac{\Delta h}{2} - H_{\tan 2} \Delta w +$$

$$HN_1 \frac{\Delta h}{2} + HN_2 \frac{\Delta h}{2} = I$$

$$H_{\tan 1} \Delta w - H_{\tan 2} \Delta w = I$$

$$\Delta w (H_{\tan 1} - H_{\tan 2}) = I$$

Let us consider a closed surface and

current  $I = 0$

$$\Delta w (H_{\tan 1} - H_{\tan 2}) = 0$$

$$\Delta w \neq 0$$

$$H_{\tan 1} - H_{\tan 2} = 0$$

$$H_{\tan 1} = H_{\tan 2}$$

$$B_{\tan 1} = \mu_1 H_{\tan 1}$$

$$B_{\tan 2} = \mu_2 H_{\tan 2}$$

$$\frac{B_{\tan 1}}{B_{\tan 2}} = \frac{\mu_1}{\mu_2}$$

Let us consider cylindrical Gaussian Surface

$$B = \frac{\Phi}{s} = \frac{q}{s} = \frac{d\Phi}{ds}$$

$$d\Phi = B \cdot ds$$

$$\Phi = \iint \vec{B} \cdot d\vec{s}$$

$$\Phi = \iint \vec{B} \cdot d\vec{s}$$

$$\Phi = \iint_{\text{Top}} \vec{B} \cdot d\vec{s} + \iint_{\text{Side}} \vec{B} \cdot d\vec{s} + \iint_{\text{Bottom}} \vec{B} \cdot d\vec{s}$$

$\Delta h \ll \Delta s$ ,  $\iint_{\text{Side}} \vec{B} \cdot d\vec{s} = 0$

$$\Phi = \iint_{\text{Top}} \vec{B} \cdot d\vec{s} + \iint_{\text{Bot}} \vec{B} \cdot d\vec{s}$$

$$\Phi = B N_1 \Delta s - B N_2 \Delta s$$

$\frac{d\Phi}{ds} = B$   
 $\frac{d\Phi}{ds} = \frac{q}{s}$

Let us consider steady state

$$0 = \Delta s (B N_1 - B N_2)$$

$$\Phi = 0 \quad \Delta s \neq 0$$

$$\rho_s = \frac{Q}{s}$$

$$I = \omega \boxed{B N_1 = B N_2}$$

$$I = \omega B N_1 = \mu_1 H N_1$$

$$I = \omega B N_2 = \mu_2 H N_2$$

$$\boxed{\frac{H N_1}{H N_2} = \frac{\mu_2}{\mu_1}}$$

$$\begin{aligned} H_{\tan 1} &= H_{\tan 2} \\ \frac{B_{\tan 1}}{B_{\tan 2}} &= \frac{\mu_1}{\mu_2} \\ B_{N1} &= B_{N2} \\ \frac{H_{N1}}{H_{N2}} &= \frac{\mu_2}{\mu_1} \end{aligned}$$

## Steady Magnetic field.

The force experienced by the charge moving with a velocity which produce the current due to magnetic flux is produced. which is known as steady magnetic field.

## Electric force:

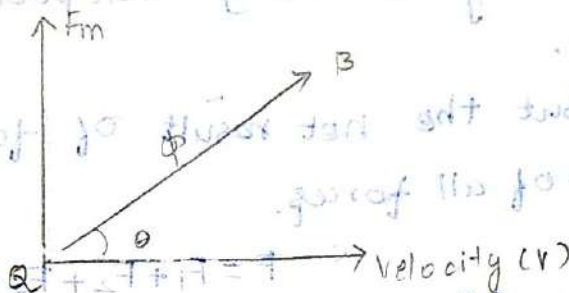
The force experienced by test charge which exists the electric field.

$$\vec{E} = \frac{F_e}{Q}$$

$$\vec{F}_e = \vec{E} \cdot Q$$

Where  $\vec{E}$  is static Electric Field.

Magnetic Force / Force on a point charge / moving charge



In a steady state magnetic field, the force experienced by the charge moving with a velocity  $v$  and produce magnetic flux density  $B'$ . Here Force is directly proportional to the magnetic flux density  $B'$  and

$\theta \rightarrow$  angle btw  $v$  and  $B'$

The magnetic force  $F_m$  is directly proportional to Both  $v$  and  $B'$

$$F \propto B' v$$

$$F_m = Q (\vec{v} \times \vec{B}')$$

$$F_m = Q |\vec{v}| |\vec{B}'| \sin \theta$$

$$\boxed{\theta = 90^\circ} \quad \sin 90^\circ = 1$$

Let consider  $\sin 90^\circ = 1$

$$F_m = Q |\vec{v}| |\vec{B}|$$

### Lorentz's Force Eqn:

The Force on moving particle, due to  
Combine electric and magnetic field.

$$\vec{F} = \vec{F}_e + \vec{F}_m$$

$$F_e \text{ for electric force} = \vec{E} Q_t$$

$$F_m \text{ for magnetic force} = Q \vec{v} \vec{B}$$

$$\vec{F} = \vec{E} Q_t + Q \vec{v} \vec{B}$$

The two or more forces  $F_1$  and  $F_2, F_3, \dots$  etc  
that they are acting independently with each  
other.

But the net result of force is equal to  
sum of all forces.

$$F = F_1 + F_2 + F_3 + \dots$$

Force acting on differential line element

(or) **current carrying conductor**

By Lorentz force eqn

$$\vec{F} = \vec{F}_e + \vec{F}_m$$

only magnetic force is charging because  
current carrying conductor  $F_e = 0$

$$F = F_m$$

$$F = Q (\vec{v} \times \vec{B})$$

$$d\vec{F} = dQ \left( \frac{d\vec{r}}{dt} \times \vec{B} \right)$$

$$= \frac{dQ}{dt} (d\vec{l} \times \vec{B})$$



$$d\vec{F} = dI (\vec{dl} \times \vec{B})$$

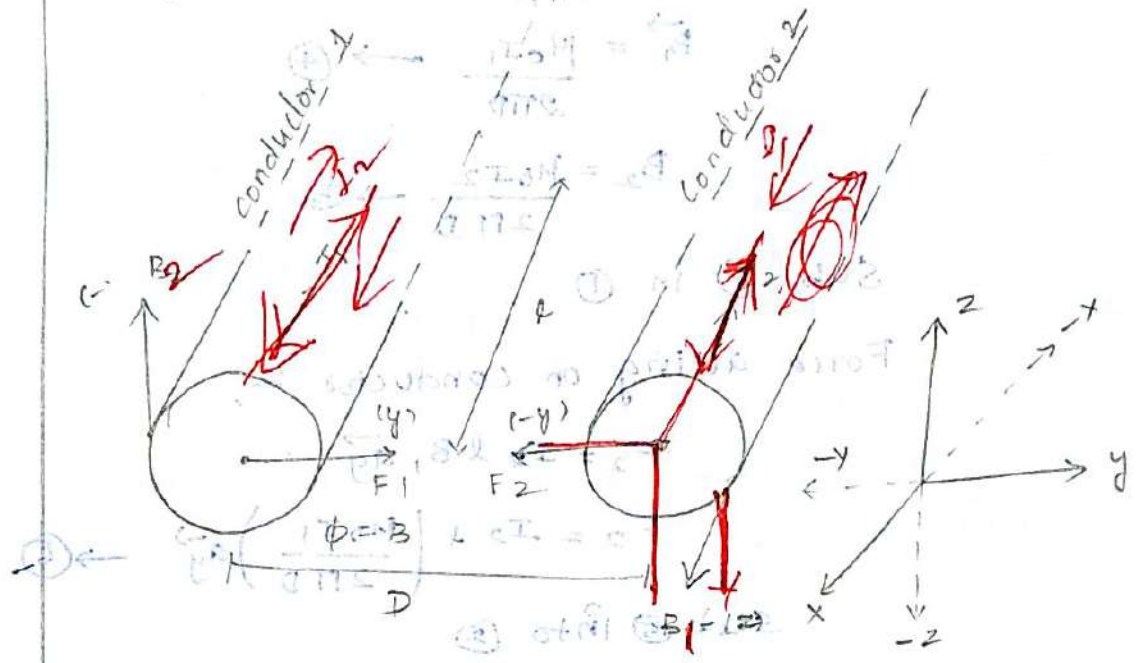
$$F = dI |\vec{dl}| |\vec{B}| \sin\theta$$

$$\theta = 90^\circ$$

$$F = I L B$$

Force acting on total length of a conductor

Force btw the two current element (or) Two parallel conductor



$a_2 \times a_1 = a_y$   
 $a_1 \times a_2 = -a_y$

Force acting on conductor 2

$$\vec{F}_2 = I_2 \vec{dl} \times \vec{B}_1$$

$$\vec{F}_2 = I_2 (-a_x) \times B_1 (-a_z)$$

$$a_x a_z = a_y$$

$$F_2 = I_2 l B_1 a_y \rightarrow \text{①}$$

Force acting on conductor 1

$$F_1 = I_1 \vec{dl} \times \vec{B}_2$$

$$= I_1 (a_x) \times B_2 (+a_z)$$

$$F_1 = I_1 l B_2 a_y \rightarrow \text{②}$$

For infinite line straight line conductor the magnetic field intensity

$$\vec{H} = \frac{I}{2\pi h}$$

$h \rightarrow$  distance =  $D$

Here the Force btw two parallel plate

Conductors separated by a distance 'D'

$$H = \frac{I}{2\pi D} \rightarrow (3)$$

R/w btw B and H

$B = \mu_0 H$  for free space.

$$\vec{B} = \frac{\mu_0 I}{2\pi D}$$

$$\vec{B}_1 = \frac{\mu_0 I_1}{2\pi D} \rightarrow (4)$$

$$\vec{B}_2 = \frac{\mu_0 I_2}{2\pi D} \rightarrow (5)$$

Sub (4) in (1)

Force acting on conductor 2

$$F_2 = I_2 \int B_1 \vec{a}_y$$

$$F_2 = I_2 \int \left( \frac{\mu_0 I_1}{2\pi D} \right) \vec{a}_y \rightarrow (6)$$

Sub (5) into (2)

Force acting on conductor 1

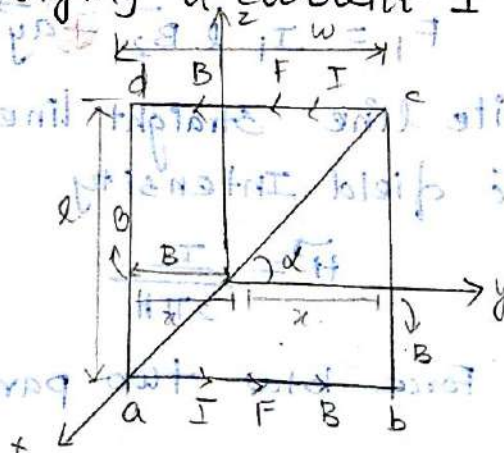
$$\vec{F}_1 = I_1 \int B_2 (-\vec{a}_y)$$

$$|\vec{F}_1| = I_1 \int \left( \frac{\mu_0 I_2}{2\pi D} \right) (\vec{a}_y)$$

$H = I$

$$|\vec{F}_1| = |\vec{F}_2|$$

Force or Torque on a closed loop conductor (or) closed loop circuit (or) Torque on a loop carrying a current I



$\alpha$  is the angle between the axis of rotation and magnetic flux line cor.  
magnetic density

$r$  - distance between reference point to the force acting on the conductor.

$L$  - Length of the conductor

$w$  - width of the conductor

Definition of Torque:

It is defined as rate of change of angle moment cor movement of rotation cor how much of force acting on a closed loop conductor. due to dispo. this force. The conductor starts to rotate with an angle  $\alpha'$  from the reference point is known as Torque.

$$\vec{\tau} = \vec{r} \times \vec{F}$$

Force acting on A to B

$$\vec{F} = I L B \sin \theta$$

Here Force are parallel line to each other

$$\theta = 0^\circ \quad \boxed{\vec{F} = 0}$$

Force acting on B to C

$$F = I L B \sin \theta$$

Here Force are acting perpendicular to each other.

$$\theta = 90^\circ \quad \sin 90^\circ = 1$$

$$\boxed{\vec{F} = I L B}$$

Force acting on C to D

$$F = I L B \sin \theta \quad \theta = 0^\circ$$

Here Force are parallel line to each other.

$$\theta = 0^\circ \quad \boxed{\vec{F} = 0}$$

Force acting on D to A

$$F = I l B \sin \theta$$

$$\theta = 90^\circ$$

$$F = I l B$$

Here Force acting on  $\perp$  to conductor

$$\text{Total Force } \vec{F} = 0 + I l B + 0 + I l B$$

$$\boxed{\vec{F} = 2 I l B}$$

$$\vec{\tau} = r \times \vec{F}$$

$$\vec{\tau} = |r| |F| \sin \alpha$$

$$= \frac{w}{2} (2 I l B) \sin \alpha$$

$$\vec{\tau} = w l I B \sin \alpha$$

$$w l = N$$

$N \rightarrow$  No. of turns

$$\vec{\tau} = N I A B \sin \alpha$$

Force acting on B to C

$$F = I l B \sin \theta$$

Here Force acting on  $\perp$

$$\text{Let } \alpha = 90^\circ$$

$$\boxed{\vec{\tau} = N I A B}$$

## Inductance

It is used to store magnetic Energy

it is defined as ratio of magnetic flux produced by the coil to the current flowing through a coil.

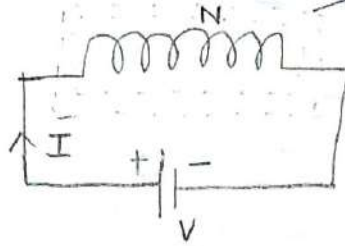
$$\boxed{L = \frac{\Phi}{I}}$$

$$\Phi \propto I$$

$N$  - No. of turns

$$I \propto N\phi$$

$$L = \frac{N\phi}{I}$$



(2m)

Statement :-

The coil consist of 'N' number of turns the current flowing through each turn which produce the magnetic flux line  $\phi$  those flux lines are imaginary lines which is stored by inductance 'L'

$$L = \frac{N\phi}{I}$$

Self Inductance :-

It is defined as flux produced by the coil due to current carrying the coil. Here the coil which linked with coil itself

Mutal Inductance :-

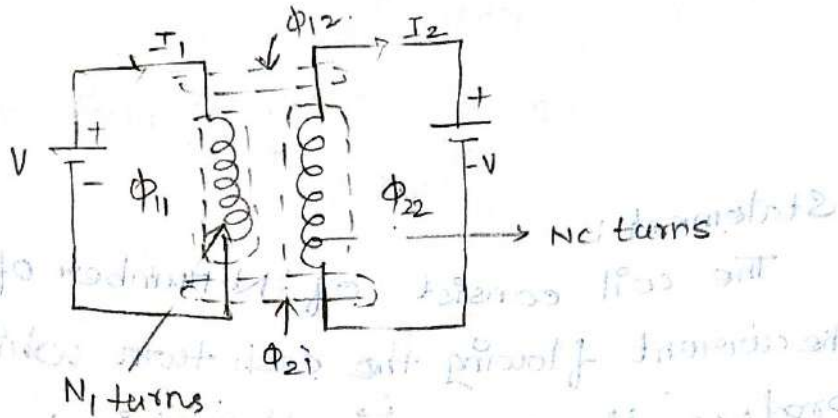
→ Let us consider the magnetic coil with difference  $M_1, M_2$  with different current  $I_1, I_2$  flowing through the coil 1 & 2

→ Due to that magnetic flux lines are produced it is coupling with each other it is known as Mutal Inductance

→ Let the coil  $N_1$  turns with inductance  $L_1$  carrying a current  $I_1$ , which produce

magnetic flux  $\Phi_{11}$  due to  $I_1$

If the coil  $N_2$  turns with inductance  $L_2$  carrying a current  $I_2$  which produce magnetic flux  $\Phi_{22}$  due to  $I_2$



Magnetic flux produced by the coil with current  $I_1$  which linked with another coil (coil 2)

Due to this magnetic flux leakage from

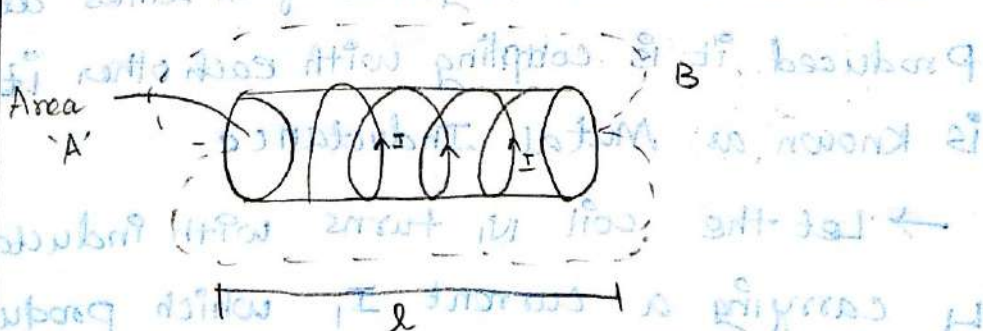
$\Phi_{12}$  
$$M_{12} = \frac{N_2 \Phi_{12}}{I_1}$$

Magnetic flux produced by the coil with current  $I_2$  which linked with another coil (coil 1)

Due to this magnetic flux leakage from  $\Phi_{21}$

$$M_{21} = \frac{N_1 \Phi_{21}}{I_2}$$

Inductance of solenoidal-



$$L = \frac{N\phi}{I}$$

Magnetic flux density

$$B = \frac{\phi}{A}$$

$$\phi = BA$$

$$L = \frac{N(BA)}{I}$$

WKT  $B = \mu H$

$$L = \frac{N\mu HA}{I}$$

Magnetic field intensity for solenoid

$$H = \frac{NI}{l}$$

$$L = \frac{N\mu A}{I} \left( \frac{NI}{l} \right)$$

$$L = \frac{N^2 \mu A I}{I l}$$

$$L = \frac{N^2 \mu A}{l}$$

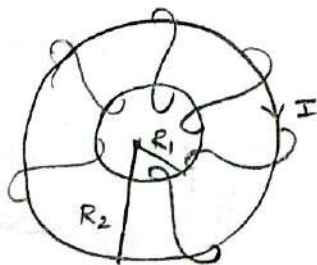
$l \rightarrow$  Inductance of solenoidal

$A \rightarrow$  area of solenoidal

$l$  - length of solenoidal

$\mu$  - permeability of magnetic field

Inductance of Toroid:



$$L = \frac{N\phi}{I}$$

Magnetic flux density  $B = \frac{\phi}{A}$

$$\Phi = \frac{BA}{\mu} \rightarrow \textcircled{2}$$

$$L = \frac{N(BA)}{I}$$

WKT,  $B = \mu H$

$$L = \frac{N\mu H A}{I} \rightarrow \textcircled{3}$$

Magnetic field Intensity for Toroid

$$H = \frac{NI}{2\pi R_m}$$

$$A = \pi r^2$$

$$LI = \frac{N\mu (\pi r^2)}{I} \left( \frac{NI}{2\pi R_m} \right)$$

$$= \frac{\mu N^2 \pi (\pi r^2) I}{2\pi R_m I}$$

$$L = \frac{\mu N^2 \pi r^2}{2R_m}$$

$L \rightarrow$  Inductance of Toroid

$r$  - radius of Toroid

$R_m$  - Mean radius  $R_m = \frac{R_1 + R_2}{2}$

$\mu$  - permeability in magnetic field

Inductance of co-axial cable / cylindrical Inductance

$a \rightarrow$  inner radius of co-axial cable

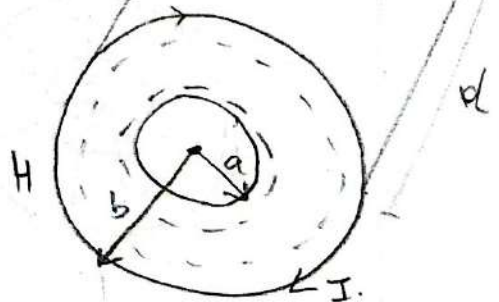
$b \rightarrow$  outer radius of co-axial cable

$d$  - co-axial cable

$\mu$  - permeability

$$L = \frac{\Phi}{I} \rightarrow \textcircled{1}$$

$$B = \mu H \rightarrow \textcircled{2}$$



Let us assume magnetic flux lines are producing  $\pi$  direction as per cylindrical co-ordinate



System differential line element  $dr, r d\phi, dz$

For infinite straight line  $H = \frac{I}{2\pi h} \rightarrow \textcircled{3}$

$$\vec{B} = \frac{\mu I}{2\pi h} a_\phi$$

$$B = \frac{\phi}{A} = \frac{d\phi}{dA} = \frac{d\phi}{ds}$$

$$d\phi = \vec{B} dA$$

$$\phi = \iint \vec{B} ds$$

$$= \iint \frac{\mu I}{2\pi h} a_\phi (dr \cdot dz a_\phi)$$

$$= \int_0^d \int_a^b \frac{\mu I}{2\pi h} dr dz [\because h=r]$$

$$= \frac{\mu I}{2\pi} \int_a^b \frac{1}{r} dr \int_0^d z dz$$

$$= \frac{\mu I}{2\pi} [\log r]_a^b [z]_0^d$$

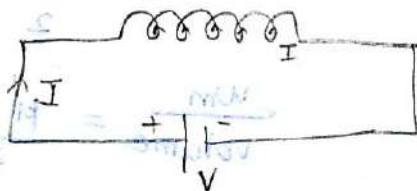
$$\phi = \frac{\mu I}{2\pi} \log \frac{b}{a} d$$

$$L = \frac{\phi}{I}$$

$$L = \frac{\mu I d \log \frac{b}{a}}{2\pi I}$$

$$L = \frac{\mu d}{2\pi} \log \frac{b}{a}$$

Energy stored in a magnetic field.



Power / Energy  $P = E = VI$

$$\frac{dw}{dt} = VI$$

$$V = L \frac{dI}{dt}$$

Sub ② in ①

$$\frac{dw}{dt} = \left( L \frac{dI}{dt} \right) I$$

$$dw = LI dI$$

$$w = \int LI dI$$

$$= L \int I dI$$

$$= L \frac{I^2}{2}$$

$$w_m = \frac{LI^2}{2}$$

In term of H and B

$$\text{Energy density } w_m = \frac{1}{2} LI^2$$

Let us consider the solenoidal

$$w_m = \frac{1}{2} \left( \frac{N^2 \mu A}{l} \right) I^2$$

$$L = \frac{N^2 \mu A}{l}$$

multiply and ÷ by l.

$$w_m = \frac{1}{2} \left( \frac{N^2 \mu A I^2}{l} \right) \left( \frac{l}{l} \right)$$

$$= \frac{N^2 I^2}{2 l^2} \left( \frac{\mu A l}{2} \right)$$

$$H \text{ for solenoidal } H = \frac{NI}{l}$$

$$w_m = \frac{\mu H^2 A l}{2}$$

(A l) = volume

$$w_m = \frac{\mu H^2 (\text{volume})}{2}$$

$$\frac{w_m}{\text{volume}} = \frac{\mu H^2}{2}$$

$$\boxed{\text{Energy density} = \frac{1}{2} \mu H^2}$$

$$= \frac{1}{2} \mu \frac{B^2}{\mu^2}$$

$$\boxed{\text{Energy Density} = \frac{1}{2} \frac{B^2}{\mu}}$$

$$= \frac{1}{2} \mu^2 H'$$

$$= \frac{1}{2} \frac{B^2}{B/H}$$

$$\boxed{\text{Energy Density} = \frac{1}{2} BH}$$

$$B = \mu H$$

$$H = \frac{B}{\mu}$$

A Ferrite Material as  $\mu_r = 10$  operate with sufficiently low flux Density  $B = 0.02$  T. Find 'H'

$$B = \mu_0 \mu_r H$$

$$B = 4\pi \times 10^7 \times 10 \times H$$

$$H = \frac{B}{4\pi \times 10^7 \times 10}$$

$$= \frac{0.02}{4 \times \pi \times 10^7 \times 10}$$

$$= \frac{0.02}{4 \times \pi \times 10^7 \times 10}$$

$$= \frac{0.02}{1.25 \times 10^9}$$

$$= 15.9 \times 10^{-12} \text{ A/m}$$

$$H = 1.592 \text{ kA/m} = 15.92 \text{ PA/m}$$

A long straight wire carries a current 5 Amps at which distance magnetic field 6 A/m

$$I = 5 \quad H = 6 \text{ A/m}$$

$$H = \frac{I}{2a} = \frac{5}{2a}$$

$$a = \frac{I}{2H} = \frac{5}{2 \times 6} = 0.416$$

2 wires carrying current in same direction  
 4 Amps and 6 Amps are placed with their axes  
 5 cm apart free space permeability calculate  
 the force between them in  $\text{N/m} \rightarrow \frac{F}{l}$

$$\frac{F}{l} = \frac{\mu_0 I_1 I_2}{2\pi d}$$

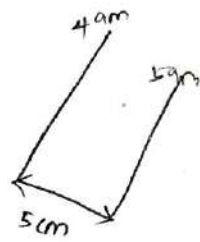
$$= \frac{4 \times 10^{-7} \times 4 \times 6}{2\pi \times 5 \times 10^{-2}}$$

$$= \frac{8 \times 10^{-7}}{5 \times 10^{-2}}$$

$$= 1.6 \times 10^{-4} \text{ N/m}$$

$$I_1 I_2$$

$$\frac{F}{l} = 1.6 \times 10^{-4} \text{ N/m}$$



A current of 3 amps flowing through inductor  
 of 100 mH what is energy stored inductor

$$W = \frac{1}{2} L I^2$$

$$= \frac{1}{2} \times 100 \times (3)^2$$

$$= \frac{1}{2} \times 100 \times 10^{-3} \times 9$$

$$= 0.45$$

$$W_m = 0.45 \text{ J}$$

If point P  $A_x = 4x + 3y + 2z$ ,  $A_y = 5x + 6y + 3z$   
 $A_z = 2x + 3y + 5z$  Determine magnetic flux density  
 and also state the nature of field.

$$\vec{B} = \nabla \times \vec{A} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ 4x+3y+2z & 5x+6y+3z & 2x+3y+5z \end{vmatrix}$$

$$= a_x \left( \frac{\partial}{\partial y} (2x+3y+5z) - \frac{\partial}{\partial z} (5x+6y+3z) \right) - a_y \left( \frac{\partial}{\partial x} (2x+3y+5z) - \frac{\partial}{\partial z} (x+3y+2z) \right) + a_z \left( \frac{\partial}{\partial x} (5x+6y+3z) - \frac{\partial}{\partial y} (3) \right)$$

$$= a_x (3-3) - a_y (2-2) + a_z (5-3)$$

$$= 0 - a_y (2-2) + 2 a_z$$

$$B = 2 a_z \neq 0$$

The field is ~~irrotational~~ rotational

It is not conservative vector field

In a perfect conducting surface in xy plane, magnetic field intensity  $\vec{H} = z \cos x a_x + z \cos x a_y$   $z > 0$  and  $z < 0$  find current density of conducting surface

$$\vec{J} = \nabla \times \vec{H}$$

$$= \begin{vmatrix} a_x & a_y & a_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ z \cos x & z \cos x & 0 \end{vmatrix}$$

$$= a_x (0 + \sin x \cos x) - a_y (0 - 0) + a_z (z \sin x - 0)$$

$$\vec{J} = -\cos x a_x - z \sin x a_z \text{ A/m.}$$

If  $\vec{A} = (3y-z) a_x + 2xz a_y$  w/m in a certain region of free space Determine  $\vec{B}, \vec{H}, \vec{J}$

$$\vec{B} = \nabla \times \vec{A}$$

$$= \nabla \times \vec{A}$$

$$\vec{B} = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ (3y-z) & 2xz & 0 \end{vmatrix}$$

$$= \vec{a}_x (0 - 2x) - \vec{a}_y (0 - (-1)) + \vec{a}_z (2z - 3)$$

$$\vec{B} = -2x\vec{a}_x - \vec{a}_y + (2z-3)\vec{a}_z \quad \text{W/m}^2$$

$$\vec{H} = \frac{\vec{B}}{\mu_0}$$

$$= \frac{-2x\vec{a}_x - \vec{a}_y + (2z-3)\vec{a}_z}{4\pi \times 10^{-7}}$$

$$\vec{H} = 795.7 \times 10^3 (-2x\vec{a}_x - \vec{a}_y + (2z-3)\vec{a}_z)$$

$$\vec{J} \Rightarrow \nabla \times \vec{H}$$

$$795.7 \times 10^3 \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ -2x & -1 & 2z-3 \end{vmatrix}$$

$$= \vec{a}_x (0-0) - \vec{a}_y (0-0) + \vec{a}_z (0-0)$$

$$\vec{J} = 795.7 \times 10^3 (0)$$

$$\vec{J} = 0 \quad \text{A/m}^2$$

In a cylindrical co-ordinate system  $\vec{B} = 2/r \vec{a}_\phi$

determine the magnetic flux passing through the

Plane surface in the range of  $0.5 \leq r \leq 2.5$

$$0 \leq z \leq 2$$

$$\Phi = ?$$

$$B = \frac{\Phi}{\pi \Delta x \Delta s} = \frac{\Phi}{\pi} \frac{d\Phi}{ds}$$

$$d\Phi = \vec{B} \cdot d\vec{s}$$

$$\Phi = \iint \vec{B} \cdot d\vec{s}$$

$$= \iint \frac{2}{r} dr dz$$

$$= \int_0^2 \int_{0.5}^{2.5} \frac{2}{r} dr \cdot dz$$

$$= \int_0^2 2 [\log r]_{0.5}^{2.5} dz$$

$$= 2 \int_0^2 [\log(2.5) - \log(0.5)] dz$$

$$= 2 [0.395 + 0.301] (2)$$

$$= 4 [0.696]$$

$$\Phi = 2.784$$

If  $\vec{B} = 2.5 \sin\left(\frac{\pi x}{2}\right) e^{-2y} \hat{a}_z$  find the total magnetic flux crossing the line where

$z=0, y \geq 0, 0 \leq x \leq 2m$

$$d\Phi = \vec{B} \cdot d\vec{s}$$

$$\Phi = \iint \vec{B} \cdot d\vec{s}$$

$$= \int_0^2 \int_0^\infty 2.5 \sin\left(\frac{\pi x}{2}\right) e^{-2y} dx dy$$

$$= \int_0^2 2.5 \left[ -\cos\left(\frac{\pi x}{2}\right) \right]_0^2 \left[ \frac{e^{-2y}}{-2} \right]_0^\infty dx$$

$$= \frac{2.5}{2} \left( \frac{2}{\pi} \right) \left[ -\cos\left(\frac{\pi}{2}\right)(2m) + \cos 0 \right]$$

$$\left[ \frac{e^{-2(\infty)}}{-2} + \frac{e^0}{-2} \right]$$

$$= 2.5 \left( \frac{2}{\pi} \right) \left[ -\cos \pi + \cos 0 \right] \left[ 0 + \frac{1}{2} \right]$$

$$= 2.5 \left( \frac{2}{\pi} \right) [2] \left[ \frac{1}{2} \right]$$

$$= 2.5 \frac{2}{\pi}$$

$$\Phi = 1.592 \text{ weber}$$

The magnetic field strength 200 turns coil carrying a current of 2 amps. the length of solenoid is 0.2m Find H

$$H = \frac{NI}{L}$$

$$= \frac{200 \times 2}{0.2}$$

$$H = 2000 \text{ A/m}$$

$\vec{B} = 0.05$  Tesla (T) and  $\mu_r = 50$ . Find H

$$H = \frac{B}{\mu_0 \mu_r}$$

$$= \frac{0.05}{4\pi \times 10^{-7} \times 50}$$

$$H = 795.77 \text{ A/m}$$

A circular coil of radius 2cm,  $B = 10 \text{ w/m}^2$  In a plane of circular coil develops  $\perp$  to the field determine the total flux around the coil.

$$\phi = \frac{B}{A} = \frac{B}{\pi r^2} = \frac{10}{3.14 \times 4 \times 10^{-4}}$$

$$B = \frac{\phi}{A}$$

$$\phi = BA$$

$$= 10 \times (2 \times 10^{-2})^2 \times 3.14$$

$$\phi = 0.0125 \text{ weber}$$

In a cylindrical co-ordinate system  $50r^2 \hat{a}_z$   $\text{wb/m}^2$  is a magnetic potential. In a certain region of free space find H, B, I



$$0 \leq r \leq 1$$

$$0 \leq \phi \leq 2\pi$$

$$\begin{aligned} \vec{B} = \nabla \times \vec{A} &= \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ 0 & 0 & 50r^2 \end{vmatrix} \\ &= \frac{1}{r} \begin{vmatrix} \vec{a}_r & r\vec{a}_\phi & \vec{a}_z \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ 0 & 0 & 50r^2 \end{vmatrix} \end{aligned}$$

$$= \frac{1}{r} [0 - r\vec{a}_\phi (100r) + 0]$$

$$= -\frac{1}{r} r \vec{a}_\phi (-100r)$$

$$\boxed{\vec{B} = -100 r \vec{a}_\phi \text{ W/m}^2}$$

$$\vec{H} = \frac{\vec{B}}{\mu_0} = \frac{-100 r \vec{a}_\phi}{4\pi \times 10^{-7}}$$

$$\boxed{\vec{H} = -7.95 r \vec{a}_\phi \times 10^7 \text{ A/m}}$$

$$\vec{J} = \nabla \times \vec{A}$$

$$= \frac{1}{r} \begin{vmatrix} \vec{a}_r & r\vec{a}_\phi & \vec{a}_z \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ 0 & -7.95 \times 10^7 r & 0 \end{vmatrix}$$

$$= \frac{1}{r} [0 - r \vec{a}_\phi (0) + \vec{a}_z (-7.95 \times 10^7)]$$

$$\boxed{\vec{J} = \frac{-7.95 \times 10^7}{r} \vec{a}_\phi \text{ A/m}^2}$$

$$I = \iint \vec{J}_z \, ds_z$$

$$I = \int_0^{2\pi} \int_0^1 -\frac{7.95 \times 10^7}{r} \vec{a}_z (r \, dr \, d\phi \, \vec{a}_z)$$

$$= -7.95 \times 10^7 \int_0^1 dr \int_0^{2\pi} d\phi$$

$$= -7.95 \times 10^7 [1] [2\pi]$$

$$I = -49.95 \times 10^7$$

$$= 499 \times 10^6$$

$$= -500 \text{ M Amps}$$

$$I \approx -500 \text{ M Amp}$$

Find  $\vec{H}$  on origin due to current element  $3\pi(a\vec{x} + 2a\vec{y} + 3a\vec{z})$  at  $(3, 4, 5)$

$$Idl = 3\pi(a\vec{x} + 2a\vec{y} + 3a\vec{z})$$

$$dH = \frac{Idl \sin\theta}{4\pi r^2}$$

$$r = \sqrt{3^2 + 4^2 + 5^2}$$

$$= \sqrt{9 + 16 + 25}$$

$$= 7.071$$

$$\theta = 90^\circ$$

$$\sin 90^\circ = 1$$

$$d\vec{H} = \frac{3\pi(a\vec{x} + 2a\vec{y} + 3a\vec{z})}{4\pi(7.071)^2}$$

$$d\vec{H} = 0.015(a\vec{x} + 2a\vec{y} + 3a\vec{z})$$

$$\vec{H} = 0.015(a\vec{x} + 2a\vec{y} + 3a\vec{z})$$

$$\vec{H} = 0.015a\vec{x} + 0.03a\vec{y} + 0.045a\vec{z} \frac{A}{m}$$

A circular loop located on  $x^2 + y^2 = 25, z = 0$  carries a direct current of 10 Amps along  $a\vec{\phi}$  determine  $\vec{H}$  at  $(0, 0, 2)$  and  $(0, 0, -2)$

$$x^2 + y^2 = a^2$$

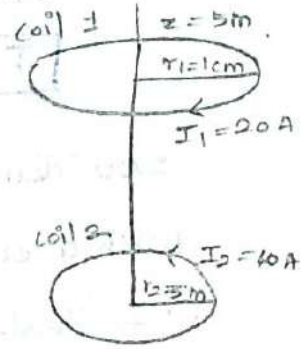
$$a = 5$$

$$I = 10A$$

$$H = \frac{I}{2a} = \frac{10}{2 \times 5} = 1 \text{ A/m}$$

A 2 circular coils are located at  $z = 0$  and  $z = 5m$  centred about the  $z$  axis. The first coil has radius of 1m and carries a current of 20 amp while

Second coil has radius of 0.5m carries the current of 40 amps calculate the H. at (0,0,3)m.



$$H = H_1 + H_2$$

$$\vec{H}_1 = \frac{I_1 r_1^2}{2(r_1^2 + z^2)^{3/2}} \vec{a}_z$$

$$= \frac{20(1)^2}{2(1^2 + 5^2)^{3/2}}$$

$$= \frac{20}{2(26)\sqrt{26}} \vec{a}_z$$

$$\vec{H}_1 = 0.079 \vec{a}_z$$

$$H_2 = \frac{I_2 r_2^2}{2(r_2^2 + z^2)^{3/2}} \vec{a}_z$$

$$= \frac{40 \times (0.5)^2}{2(0.5^2 + 0)^{3/2}}$$

$$= \frac{40(0.25)}{2(0.5)^2}$$

$$H_2 = 40 \vec{a}_z$$

$$H = H_1 + H_2$$

$$= 40 + 0.079$$

$$H = 40.079 \vec{a}_z$$

What is maximum Torque on a square loop 1000 turns in a field of flux density 2 Tesla. the loop as 10cm sides carrying the current of 3 Amps

$$T = NIAB \sin \theta$$

$$N = 1000, I = 3A, B = 1T$$

$$A = \text{side} \times \text{side}$$

$$= 10 \text{ cm} \times 10 \text{ cm}$$

$$= 10 \times 10^{-2} \times 10 \times 10^{-2}$$

$$A = 0.01$$

$$T = NIA\vec{B}$$

$$= 1000 \times 3 \times 0.01 \times 1$$

$$T = 30 \text{ N}$$

200 Turns Rectangular coil with area  $30 \text{ cm} \times 15 \text{ cm}$   
 with a current of 5 amps in uniformly field of  
 0.2 Tesla find Torque and magnetic moment

$$N = 200 \quad A = 30 \times 15 \text{ cm} = 0.045 \text{ m}^2 \quad I = 5 \quad B = 0.2$$

$$T = NIA\vec{B}$$

$$T = 9 \text{ N}$$

$$M = IA = 0.225$$

$$M = 0.225$$

A conductor 6m long lies along z direction  
 with current of 2amps  $\vec{a}_z$  find force experienced  
 by the conductor. If  $\vec{B} = 0.08 \text{ T } \vec{a}_x$

$$I = 2 \vec{a}_z \text{ A} \quad l = 6 \text{ m} \quad B = 0.08 \vec{a}_x$$

$$F = IlB \sin \theta$$

$$= 2 \times 6 \times 0.08$$

$$F = 0.96 \vec{a}_y \text{ N}$$

Calculate the inductance of solenoid  $N = 2000$  turns  
 bounded uniformly over a length 0.5m on a  
cylindrical core of diameter 4cm in free space

$$L = \frac{N^2 \mu_0 \mu_r A}{l}$$

$$L = \frac{N^2 \mu_0 A}{l}$$

$$N = 2000 \quad l = 0.5 \text{ m} \quad \mu = \mu_0 = 4\pi \times 10^{-7}$$

$$A = \pi r^2$$

$$= \pi (2 \times 10^{-2})^2$$

$$= 1.256 \times 10^{-3}$$

$$= \frac{63.13 \times 10^3}{0.5}$$

$$= 0.01264$$

Magnetisation:-

$$M = \frac{\text{Magnetic moment}}{\text{Volume}}$$

$$= \frac{Q \cdot d}{V}$$

$$= \frac{Q \cdot d}{m^2 \cdot m}$$

$$= \frac{Q \cdot d}{A \cdot d}$$

$$V = A \times d \\ = m^2 \times m$$

$$\boxed{M = \frac{Q}{A}}$$

Magnetic Susceptibility ( $\chi_m$ )

$$\chi_m = \frac{M}{H}$$

$$M = \chi_m \cdot H$$

$$B = \mu_0 H \text{ at free space}$$

$$B = \mu_0 (H + M)$$

$$= \mu_0 (H + \chi_m H)$$

$$B = \mu_0 H (1 + \chi_m) \rightarrow \textcircled{1}$$

$$B = \mu_0 \mu_r H \rightarrow \textcircled{2}$$

compare  $\textcircled{1}$  &  $\textcircled{2}$

$$\mu_r = 1 + \chi_m$$

$$\chi_m = \mu_r - 1$$

$$\boxed{\chi_m = \mu_r - 1}$$

$\mu_r \rightarrow$  Relative permeability

$\chi_m \rightarrow$  magnetic susceptibility.

# TIME VARYING FIELDS AND MAXWELL'S EQUATIONS

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# Fundamental Relations For electrostatic & Magnetostatic fields ①

- Electrostatics corresponds to stationary charges whereas magnetostatics corresponds to steady currents.

- Two important quantities that define static electric fields are electric field intensity  $E$  & electric flux density  $D$ .

$$D = \epsilon E$$

- The fundamental differential equations that govern the static electric fields are

$$\nabla \times E = 0 \quad (\text{conservative property of electrostatics})$$

$$\nabla \cdot D = \rho_v \quad (\text{Gauss law for electrostatics})$$

$\rho_v$  - Volume charge density

The 2 important quantities that define static magnetic fields are magnetic field intensity  $H$  & magnetic flux density  $B$ .

$$B = \mu H$$

The fundamental differential equations that govern the static magnetic fields are

$$\nabla \cdot B = 0 \quad (\text{Gauss law for magnetostatics})$$

$$\nabla \times H = J \quad (\text{Ampere's Circuital law})$$

$J$  - current density

## Fundamental relations

## Electrostatics

## Magnetostatics

## Sources

## Stationary charges

## Steady current

Static field condition

$$\frac{d\phi}{dt} = 0$$

$$\frac{dI}{dt} = 0$$

Field quantities

$$\mathbf{E} \ \& \ \mathbf{D}$$

$$\mathbf{H} \ \& \ \mathbf{B}$$

Constitutive parameters

$$\epsilon \ \& \ \sigma$$

$$\mu$$

Constitutive relations

$$\mathbf{D} = \epsilon \mathbf{E}$$

$$\mathbf{B} = \mu \mathbf{H}$$

Field equations in differential or point form

$$\begin{aligned} \nabla \cdot \mathbf{D} &= \rho \\ \nabla \times \mathbf{E} &= 0 \end{aligned}$$

$$\begin{aligned} \nabla \cdot \mathbf{B} &= 0 \\ \nabla \times \mathbf{H} &= \mathbf{J} \end{aligned}$$

Field equation in integral form

$$\begin{aligned} \int \mathbf{D} \cdot d\mathbf{s} &= \mathcal{Q} \\ \int \mathbf{E} \cdot d\mathbf{l} &= 0 \end{aligned}$$

$$\begin{aligned} \int \mathbf{B} \cdot d\mathbf{s} &= 0 \\ \int \mathbf{H} \cdot d\mathbf{l} &= \mathcal{I} \end{aligned}$$

Force on charge  $Q$ 

$$\mathbf{F}_e = Q\mathbf{E}$$

$$\mathbf{F}_m = Q\mathbf{v}\mathbf{B}$$

Flux

$$\begin{aligned} \psi &= \int \mathbf{D} \cdot d\mathbf{s} \\ \psi &= \mathcal{Q} = CV \end{aligned}$$

$$\begin{aligned} \phi &= \int \mathbf{B} \cdot d\mathbf{s} \\ \phi &= LI \end{aligned}$$

Potential

$$\mathbf{E} = -\nabla V$$

$$\mathbf{H} = -\nabla V_m$$

Energy density

$$W_e = \frac{1}{2} \epsilon E^2$$

$$W_m = \frac{1}{2} \mu H^2$$



$$\nabla \cdot \vec{v} = \frac{-fv}{\epsilon}$$

$$\nabla^2 A = -\mu J$$

Circuit elements

R & C

L

(3)

## Faraday's law of Electromagnetic induction

- When the magnetic flux linking a circuit changes an emf is always induced in it. The magnitude of such an emf is proportional to the rate of change of flux linkages.

- When a conductor moves through a magnetic field by cutting its flux, an emf is induced in it.

|| If when the magnetic flux cuts a stationary conductor an emf is induced.

In either case the induced emf & the rate of change of the magnetic flux are related in differential form as

$$V = \frac{-d\phi}{dt} \rightarrow \text{①}$$

V - Total EMF force in Volt     $\phi$  - total magnetic flux wb  
t - time in second.

The process of inducing an emf in a conductor in the presence of time varying magnetic field

- The -ve sign in eqn 1 indicates that the induced emf opposes the flux producing it. This is known as Lenz's law, which states that any induced emf will circulate a current in such a direction so as to oppose the cause producing it.

Equation ① is applicable to single-turn loop. For a multi-turn loop where all turns are associated with the same flux  $\Phi$ , Faraday's law may be expressed as

$$V = -N \frac{d\Phi}{dt} \rightarrow \text{②}$$

$N$  - No of turns of the loop

If every turn is not associated with the same value of flux. Then the Faraday's law may be expressed as

$$V = -\frac{d\lambda}{dt} \rightarrow \text{③}$$

$\lambda$  - total flux linkage in Weber-turns. Hence for  $N$  turns the linkage is

$$\lambda = \Phi_{m1} + \Phi_{m2} + \dots + \Phi_{mN} \rightarrow \text{④}$$

$\Phi_{m1}$  - flux associated with the 1<sup>st</sup> turn  
 $\Phi_{m2}$  - " " " 2<sup>nd</sup> turn  
 $\Phi_{mN}$  - " " " N<sup>th</sup> turn

The magnetic flux  $\phi$  passing through a loop is defined as the surface integral of normal component of magnetic flux density  $B$  over the surface area of the loop as given by (5)

$$\phi = \iint B \cdot ds \rightarrow (5)$$

The induced emf can be defined in terms of electric field intensity  $E$  as

$$V = \int E \cdot dl \rightarrow (6)$$

$l$  - closed path of integration. eqn can be expressed in terms of  $E$  &  $B$  as

$$V = \int E \cdot dl = - \frac{d}{dt} \iint B \cdot ds = - \iint \frac{dB}{dt} \cdot ds \rightarrow (7)$$

eqn (7) is known as the integral form of Faraday's law.

Transformer & motional Electromotive force

An emf is induced in a coil or conductor whenever there is a change in flux linkages.  
 $\therefore$  an emf can be produced in a closed conducting circuit by the following ways

The magnetic flux  $\phi$  passing through a loop is defined as the surface integral of normal component of magnetic flux density  $B$  over the surface area of the loop as given by

$$\phi = \iint B \cdot ds \rightarrow (5)$$

The induced emf can be defined in terms of electric field intensity  $E$  as

$$V = \int E \cdot dl \rightarrow (6)$$

$l$  - closed path of integration. eqn can be expressed in terms of  $E$  &  $B$  as

$$V = \int E \cdot dl = - \frac{d}{dt} \iint B \cdot ds = - \iint \frac{dB}{dt} \cdot ds \rightarrow (7)$$

eqn (7) is known as the integral form of Faraday's law.

Transformer is motional Electromotive force

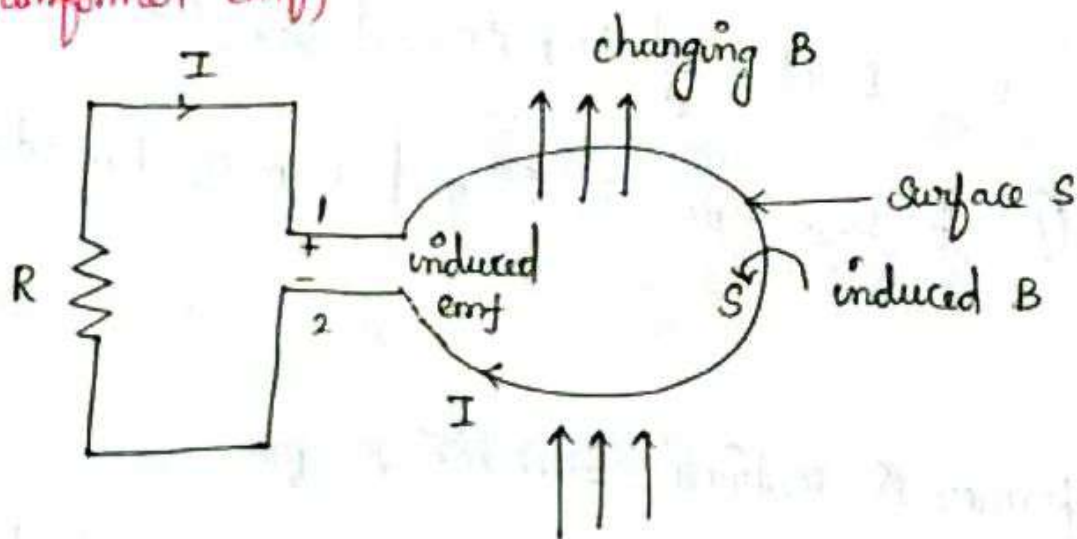
An emf is induced in a coil or conductor whenever there is a change in flux linkages.  
 $\therefore$  an emf can be produced in a closed conducting circuit by the following ways

i) By placing a stationary conductor in a time varying magnetic field (the induced emf in this case is called transformer emf) (6)

ii) By placing a moving conductor in a static magnetic field (the induced emf in this case is called motional emf)

iii) By placing a moving conductor in a time varying field.

Stationary conductor in time varying magnetic field (transformer emf)



- A stationary conductor carrying current  $I$  is placed in a time varying magnetic field of flux density  $B$

- The induced current flows in such a way to satisfy Lenz's law so that a magnetic field is produced which opposes a change in value of  $B$ .

As per Faraday's law

$$V = \oint_l E \cdot dl = - \int_s \frac{dB}{dt} ds \rightarrow (1) \quad (7)$$

- The emf given by the above equation is termed as transformer emf, which is due to the time varying current generating time varying magnetic field in a stationary loop. This effect is mainly due to transformer action.

Using Stoke's theorem

$$\oint_s (\nabla \times E) ds = - \int_s \frac{dB}{dt} ds \rightarrow (2)$$

Comparing the surface integrals on both sides

$$\nabla \times E = - \frac{dB}{dt} \rightarrow (3)$$

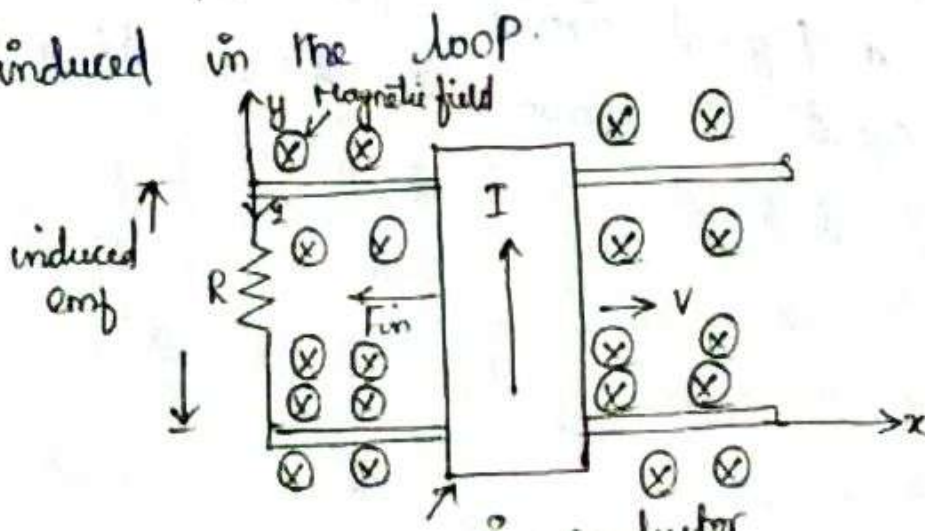
The above eqn is in a differential form which relates the field quantities at any point in space, whether or not a physical circuit exist at that point. This equation is known as the differential form of Faraday's law which states that a time varying magnetic field induces an electric field  $E$  whose curl is equal to the -ve of the time derivation of  $B$ .

Eqn (3) is one of the Maxwell's equations for time varying field & it also shows that time varying electric field is not conservative in nature i.e.  $\nabla \times E \neq 0$ . Since the work done in moving a charge on a closed path in a time varying electric field is due to energy from the time varying magnetic field laws of energy conservation are thus satisfied. Suppose if  $B$  is time independent i.e.  $\frac{dB}{dt} = 0$  the equation (1) & (3) reduced to the electrostatic equations

$$\nabla \times E = 0 \quad \& \quad \oint E \cdot dl = 0$$

Moving conductor in static magnetic field (Motional emf)

Consider a moving conductor carrying current  $I$  is placed in a static magnetic field. An emf is induced in the loop.



- The force  $F_m$  on a charge  $Q$  moving with an uniform velocity placed in a magnetic field of flux density  $B$  is given by (7)

$$F = Q(V \times B) \rightarrow (1)$$

$\therefore$  the electric field intensity is given by

$$E = \frac{F}{Q} = V \times B \rightarrow (2)$$

The field produced by the motion of the charged particle is known as motional electric field & its direction is normal to the plane containing  $V$  &  $B$ .

If we assume that a large no. of free electrons moving with an uniform velocity  $V$  is present in a conducting loop, then the emf induced in the loop is given by

$$V = \oint E \cdot dl = \oint (V \times B) \cdot dl \rightarrow (3)$$

The eqn is termed as motional emf or flux cutting emf, as it is caused by motional effect. The motional emf is present in generators & motors.

Applying Stoke's theorem

$$\int (\nabla \times E) \cdot ds = \nabla \times (V \times B) \cdot ds \rightarrow (4)$$

it will on both side



## Moving conductor in time varying magnetic field

If a moving conductor carrying current  $I$  is placed in a time varying magnetic field, then the induced emf is the sum of both transformer emf & motional emf.

$$V = \oint_l \mathbf{E} \cdot d\mathbf{l} = \text{Transformer emf} + \text{motional emf}$$

$$V = - \int_S \frac{dB}{dt} \cdot d\mathbf{s} + \oint_l (\mathbf{v} \times \mathbf{B}) \cdot d\mathbf{l}$$

Differential & integral form of Maxwell's equations

### Maxwell's equation I

From Ampere's circuital law:-

It states that line integral of magnetic field intensity  $H$  on any closed path is equal to current enclosed by that path.

$$\oint H \cdot d\mathbf{l} = I = \int_S \mathbf{J} \cdot d\mathbf{s}$$

current involves both conduction current & displacement current.

A current through resistive element is called conduction current whereas current through capacitive element is called displacement current.

current through a conductor of resistance  $R$  is

But

$$R = \frac{\rho l}{A} = \frac{l}{\sigma A}$$

$$\sigma = \frac{1}{\rho}$$

$A \rightarrow$  Area of cross section

$\sigma$  - conductivity

$\rho$  - Resistivity (11)

$$I_c = \frac{V \sigma A}{l}$$

if  $E$  is electric field then  $V = El$

$$I_c = \frac{El \sigma A}{l} = \sigma EA$$

$$\frac{I_c}{A} = \sigma E = J_c$$

current through a capacitor is

$$I_D = dQ/dt$$

$$Q = CV$$

$$I_D = C \frac{dV}{dt}$$

W.H.T

$$C = \epsilon A/d$$

$\epsilon$  - Permittivity of medium

$A$  - Area of parallel plate capacitor

$d$  - Distance b/w plates

$$I_D = \frac{\epsilon A}{d} \frac{dV}{dt}$$

$$V = Ed$$

$$I_D = \frac{\epsilon A}{d} d \frac{dE}{dt}$$

$$I_D = \epsilon A \frac{dE}{dt}$$

$$\frac{I_D}{A} = \epsilon \frac{\partial E}{\partial t} = \frac{dD}{dt}$$

(12)

$$J_D = \frac{dD}{dt}$$

$$\oint H \cdot dl = \iint (J_c + J_D) \cdot ds$$

$$\oint H \cdot dl = \iint \left( \sigma E + \frac{dD}{dt} \right) \cdot ds$$

$$\oint H \cdot dl = \iint \left( \sigma E + \epsilon \frac{dE}{dt} \right) \cdot ds$$

$$\boxed{\oint H \cdot dl = \iint \left( J + \frac{dD}{dt} \right) \cdot ds} \rightarrow \textcircled{1}$$

eqn ① is Maxwell's equations I in integral form for  
ampere's circuital law

By applying Stoke's theorem

$$\oint H \cdot dl = \iint (\nabla \times H) \cdot ds \rightarrow \textcircled{2}$$

Comparing eqn ① & ②

$$\nabla \times H = J + \frac{dD}{dt} \rightarrow \textcircled{3}$$

eqn ③ is Maxwell's eqn I in point or differential  
form

# Maxwell's equation II

(13)

Faraday's law: It states EM force induced in a circuit is equal to rate of  $\downarrow$  of magnetic flux linkage in the circuit.

$$V = -\frac{d\phi}{dt}$$

$$V = -\frac{d}{dt} (\iint B \cdot ds)$$

$$V = \oint E \cdot dl$$

$$\oint E \cdot dl = -\frac{d}{dt} \iint B \cdot ds$$

$$\oint E \cdot dl = -\iint \frac{\partial B}{\partial t} \cdot ds$$

$$= -\mu \iint \frac{\partial H}{\partial t} \cdot ds \rightarrow \textcircled{A}$$

eqn (A) is Maxwell's eqn II in integral form. By applying Stoke's theorem

$$\oint E \cdot dl = \iint (\nabla \times E) \cdot ds \rightarrow \textcircled{B}$$

Comparing eqn (A) & (B)

$$\iint (\nabla \times E) \cdot ds = -\mu \iint \frac{\partial H}{\partial t} \cdot ds$$

$$\boxed{\nabla \times E = -\mu \frac{\partial H}{\partial t} = -\frac{\partial B}{\partial t}} \rightarrow \textcircled{C}$$

eqn (C) is Maxwell's eqn II in point or differential

## Maxwell's equation III

Electric Gauss law: It states that electric flux passing through any closed surface is equal to charge enclosed by that surface.

$$\psi = Q$$

$$\iint D \cdot ds = Q \quad (\text{or}) \quad \iiint f_v dv = Q$$

$$\iint D \cdot ds = \iiint f_v dv \quad \rightarrow (7)$$

eqn (7) is Maxwell's eqn III in integral form. By applying Divergence theorem

$$\iint D \cdot ds = \iiint \nabla \cdot D dv \quad \rightarrow (8)$$

Comparing eqn (7) & (8)

$$\iiint \nabla \cdot D dv = \iiint f_v dv$$

$$\boxed{\nabla \cdot D = f_v = f} \quad \rightarrow (9)$$

eqn (9) is Maxwell's equation III in point or differential form

# Maxwell's equation IV

(15)

Magnetic Gauss law: It states that total magnetic flux through any closed surface is equal to zero

$$\phi = 0$$

$$\iint B \cdot ds = 0 \rightarrow (10)$$

eqn (10) is Maxwell's equations IV in integral form.

By applying Divergence theorem

$$\iint B \cdot ds = \iiint \nabla \cdot B \, dv \rightarrow (11)$$

Comparing eqn (10) & (11)

$$\iiint \nabla \cdot B \, dv = 0$$

$$\boxed{\nabla \cdot B = 0} \rightarrow (12)$$

eqn (12) is Maxwell's equation IV in point or differential form

\* If  $\vec{D} = 10x\vec{a}_x - 4y\vec{a}_y + kz\vec{a}_z$   $\mu C/m^2$  &  $\vec{B} = 2ay$  mT.

Find the value of k to satisfy the Maxwell's equations for region  $\sigma = 0, \rho_v = 0$

Sol  $\vec{D} = 10x\vec{a}_x - 4y\vec{a}_y + kz\vec{a}_z$   $\mu C/m^2$

$\vec{B} = 2ay$  mT,  $\sigma = 0, \rho_v = 0$

$$\left( \frac{d}{dx} \vec{a}_x + \frac{d}{dy} \vec{a}_y + \frac{d}{dz} \vec{a}_z \right) \cdot (10x \vec{a}_x - 4y \vec{a}_y + kz \vec{a}_z) =$$

(16)

$$\frac{d}{dx} (10x) - \frac{d}{dy} (4y) + \frac{d}{dz} (kz) = 0$$

$$10 - 4 + k = 0$$

$$k = -6 \mu\text{C}/\text{m}^2$$

\*) if the magnetic field  $\vec{H} = (3x \cos \beta + by \sin \alpha) \vec{a}_z$ .  
 Find current density  $\vec{J}$  if fields are invariant with time.

$$\vec{H} = (3x \cos \beta + by \sin \alpha) \vec{a}_z$$

From Maxwell's 2nd eqn

$$\nabla \times \vec{H} = \vec{J} + \frac{d\vec{D}}{dt}$$

The fields are invariant with time so  $\frac{d\vec{D}}{dt} = 0$

$$\nabla \times \vec{H} = \vec{J}$$

$$\vec{J} = \begin{vmatrix} \vec{a}_z & \vec{a}_y & \vec{a}_x \\ \frac{d}{dz} & \frac{d}{dy} & \frac{d}{dx} \\ 0 & 0 & 3x \cos \beta + by \sin \alpha \end{vmatrix}$$

$$= \vec{a}_x \left[ \frac{d}{dy} (3x \cos \beta + by \sin \alpha) \right] - \vec{a}_y \left[ \frac{d}{dx} (3x \cos \beta + by \sin \alpha) \right] + \vec{a}_z (0 \cdot 0)$$

$$\vec{J} = 6 \sin \alpha \vec{a}_x - 3 \cos \beta \vec{a}_y \text{ A/m}^2$$

\*) For 1 Ampere conductor current in copper wire,  
 find displacement current at 100 MHz. Assume for  
 copper  $\sigma = 5.8 \times 10^7 \text{ } \Omega/\text{m}$  (17)

Sol  $I_c = 1 \text{ Ampere}$

$f = 100 \text{ MHz}$

$\sigma = 5.8 \times 10^7 \text{ } \Omega/\text{m}$

Conduction current  $I_c = J_c A = 1 \text{ Ampere}$

$J_c = I/A$

$\sigma E = I/A$

$E = \frac{I}{\sigma A} = \frac{1}{5.8 \times 10^7 \times A}$

$E = \frac{0.172 \times 10^{-7}}{A} \text{ V/m}$

Displacement current  $I_D = \omega \epsilon E A$

$= \omega \epsilon_0 \epsilon_r E A$

$= 2\pi f \epsilon_0 \epsilon_r E A$

$= 2\pi \times 100 \times 10^6 \times 8.854 \times 10^{-12} \times 1 \times \frac{0.172 \times 10^{-7}}{A} \times A$

$I_D = 9.556 \times 10^{-11} \text{ Ampere}$



## Potential functions

For static EM field, the electric scalar potential is given by

$$V = \int \frac{\rho_v dv}{4\pi\epsilon R} \rightarrow (1)$$

& magnetic vector potential is

$$A = \int \frac{\mu J dv}{4\pi R} \rightarrow (2)$$

$$\text{W.K.T } B = \nabla \times A \rightarrow (3)$$

$$\nabla \times E = -\frac{dB}{dt} \rightarrow (4)$$

Sub eqn (3) in eqn (4)

$$\nabla \times E = -\frac{d}{dt} (\nabla \times A) \rightarrow (5)$$

$$\nabla \times \left( E + \frac{dA}{dt} \right) = 0 \rightarrow (6)$$

Since curl of gradient of a scalar field is identically zero, the solution to eqn (6) is

$$E + \frac{dA}{dt} = -\nabla V \rightarrow (7)$$

or

$$E = -\nabla V - \frac{dA}{dt} \rightarrow (8)$$

From eqn (3) & (8) we can determine the vector field provided the potentials  $A$  &  $V$  are known.

W.K.T  $\nabla \cdot D = f_v$  is valid for time varying

conditions. By taking the divergence of eqn (8)

(19)

$$\nabla \cdot E = \nabla \cdot \left( -\nabla V - \frac{\partial A}{\partial t} \right)$$

$$\nabla \cdot E = -\nabla^2 V - \frac{d}{dt} (\nabla \cdot A) \rightarrow (9)$$

By making use of  $D = \epsilon E$  &  $\nabla \cdot D = f_v$  eqn (9) becomes

$$\nabla \cdot E = \frac{f_v}{\epsilon} = -\nabla^2 V - \frac{d}{dt} (\nabla \cdot A)$$

$$\nabla^2 V + \frac{d}{dt} (\nabla \cdot A) = -f_v / \epsilon \rightarrow (10)$$

Taking the curl of equation (3)

$$\nabla \times \nabla \times A = \nabla \times B \rightarrow (11)$$

$$\text{W.K.T } B = \mu H \rightarrow (12)$$

Sub eqn (12) in eqn (11)

$$\nabla \times \nabla \times A = \mu (\nabla \times H)$$

$$= \mu \left( J + \frac{\partial D}{\partial t} \right) = \mu \left( J + \epsilon \frac{\partial E}{\partial t} \right)$$

$$\nabla \times \nabla \times A = \mu J + \mu \epsilon \frac{\partial E}{\partial t} \rightarrow (13)$$

Sub eqn (8) in eqn (13)

$$\nabla \times \nabla \times A = \mu J + \mu \epsilon \frac{d}{dt} \left( -\nabla V - \frac{\partial A}{\partial t} \right)$$

$$\mu J - \mu \epsilon \nabla \left( \frac{\partial V}{\partial t} \right) - \mu \epsilon \frac{\partial^2 A}{\partial t^2} \rightarrow (14)$$

Where  $D = \epsilon E$  &  $B = \mu H$  have been assumed

By applying Vector identity

$$\nabla \times \nabla \times A = \nabla (\nabla \cdot A) - \nabla^2 A \quad \text{to eqn (14)} \rightarrow (15)$$

$$\nabla^2 A - \nabla (\nabla \cdot A) = -\mu J + \mu \epsilon \nabla \left( \frac{\partial V}{\partial t} \right) + \mu \epsilon \frac{\partial^2 A}{\partial t^2} \rightarrow (16)$$

A vector field is uniquely defined when its curl & divergence are specified

$$\nabla \cdot A = -\mu \epsilon \frac{\partial V}{\partial t} \rightarrow (17)$$

The above equation relates  $A$  &  $V$  & it is called Lorentz condition for potentials. Sub eqn (17) in eqn

$$\nabla^2 V + \frac{\partial}{\partial t} \left( -\mu \epsilon \frac{\partial V}{\partial t} \right) = \frac{-\rho_v}{\epsilon}$$

$$\nabla^2 V - \mu \epsilon \frac{\partial^2 V}{\partial t^2} = -\rho_v / \epsilon \rightarrow (18)$$

Sub eqn (17) in eqn (16)

$$\nabla^2 A - \mu \epsilon \frac{\partial^2 A}{\partial t^2} = -\mu J \rightarrow (19)$$

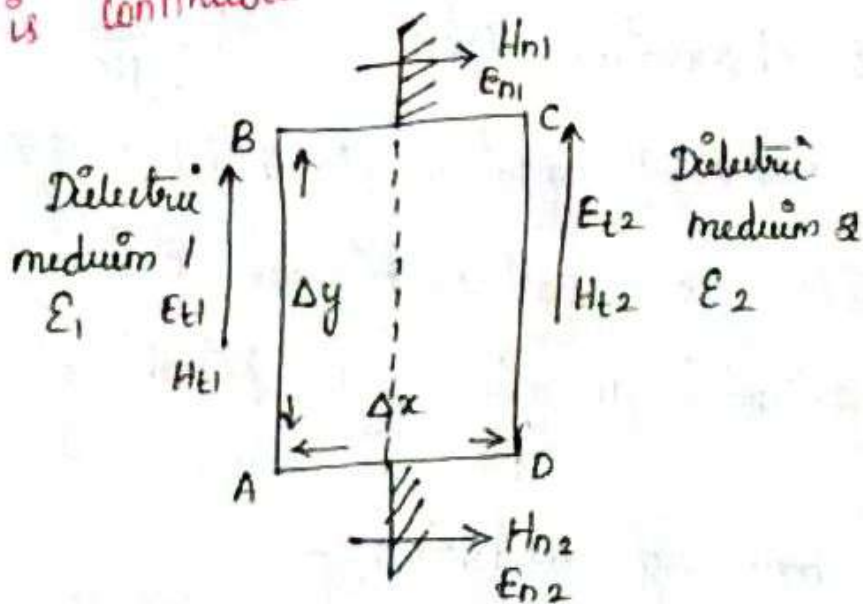
eqn (18) & (19) are called wave equation for potentials. It can be shown that the solution to equation (18)

$$(19) \quad V = \int_V \frac{[\rho_v] dv}{4\pi \epsilon R} \rightarrow (20)$$

$$V = \int_V \frac{[\rho_v] dv}{4\pi \epsilon R} \rightarrow (21)$$

3. The normal component of electric flux density  $D$  is continuous if there is no surface charge density,  $D$  is discontinuous by an amount equal to surface charge density.

A. The normal component of magnetic flux density  $B$  is continuous at the surface of discontinuity.



consider a rectangle of length  $\Delta y$  & width  $\Delta x$  at the boundary of 2 dielectric media.

$$\oint \mathbf{E} \cdot d\mathbf{l} = 0$$

Apply this to the rectangular path ABCD in which AE is just inside the medium 2

$$\oint \mathbf{E} \cdot d\mathbf{l} = E_{t1} \Delta y + E_{n1} \Delta x - E_{t2} \Delta y - E_{n2} \Delta x$$

$E_{t1}$  &  $E_{t2}$  are tangential component of  $E$  along the path AB & CD

$E_{n1}$  &  $E_{n2}$  are normal component of  $E$  along the path

The side AB & CD are brought closer together  
 the length BC & AD approaches zero.  $\Delta x \rightarrow 0$

(23)

$$E_{t1} \Delta y - E_{t2} \Delta y = \oint E \cdot dl = 0$$

$$E_{t1} = E_{t2}$$

The integral form of 1st Maxwell's equation is

$$\oint H \cdot dl = \iint (J + \frac{\partial D}{\partial t}) \cdot ds$$

Apply this to the rectangular path ABCD

$$H_{t1} \Delta y + H_{n1} \Delta x - H_{t2} \Delta y - H_{n2} \Delta x = \iint (J + \frac{\partial D}{\partial t}) \cdot \Delta x \Delta y$$

$H_{t1}$  &  $H_{t2}$  are tangential component of H along path

AB & CD

$H_{n1}$  &  $H_{n2}$  are normal component of H along the path BC & AD.

$$\text{As } \Delta x \rightarrow 0 \text{ then } H_{t1} \Delta y - H_{t2} \Delta y = 0$$

$$H_{t1} = H_{t2}$$

For a perfect conductor a HF current will flow in a thin sheet near the surface. In a current sheet a linear current density  $J_l$  flows in a sheet of depth  $\Delta x$ .

$$\text{Lt } \Delta x \rightarrow 0 \quad J \cdot \Delta x = J_l$$

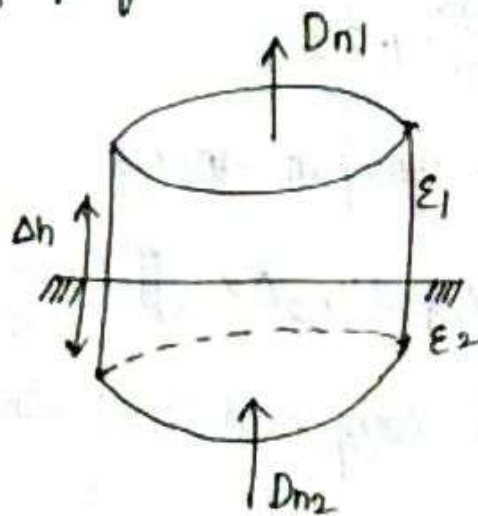
If the Maxwell's 1st equation is applied to the rectangle

$$H_{t2} \Delta y - H_{n2} \Delta x = J \Delta x \Delta y + \frac{\partial D}{\partial t} \Delta x \Delta y$$

$$H_{t1} \Delta y - H_{t2} \Delta y = J_e \Delta y$$

$$H_{t1} - H_{t2} = J_e$$

The tangential component of  $H$  is discontinuous by an amount of linear current density at the surface of perfect conductor.



Consider a pill box at boundary of a dielectric dielectric constant  $\epsilon_1$  &  $\epsilon_2$

The integral form of Maxwell's 3rd equation

$$\oint_S \mathbf{D} \cdot d\mathbf{s} = \iiint_V \rho \, dv$$

Assume there are no free charges on boundary.

Apply Gauss law to the pill box at the boundary

$$\int (D_{n1} ds - D_{n2} ds) = 0$$

$D_{n1}$  - normal component of electric flux density in medium 1  
 Normal component of electric flux density in

$$D_{n1} ds - D_{n2} ds = 0$$

(95)

$$D_{n1} = D_{n2}$$

The normal component of  $D$  is continuous if there is no surface charge density if the charges are enclosed by pill box  $\Delta h \rightarrow 0$

$$\int D \cdot ds = Q$$

$$D_{n1} ds - D_{n2} ds = Q$$

$$D_{n1} - D_{n2} = \frac{Q}{ds} = \rho_s$$

$$D_{n1} - D_{n2} = \rho_s$$

The normal component of  $D$  is discontinuous across the boundary by the amount of surface charge density. The integral form of Maxwell's 4<sup>th</sup> eqn is

$$\iint B \cdot ds = 0$$

Apply to the pill box at the boundary

$$B_{n1} ds - B_{n2} ds = 0$$

$$B_{n1} = B_{n2}$$

The normal component of magnetic flux density  $B$  is continuous across the boundary.

# Electromagnetic wave equation

(26)

## Wave equation for conducting medium

The Maxwell's equation from Faraday's law in point form is given by

$$\nabla \times \mathbf{E} = -\frac{d\mathbf{B}}{dt} = -\mu \frac{d\mathbf{H}}{dt} \rightarrow (1)$$

Taking curl on both sides

$$\nabla \times \nabla \times \mathbf{E} = -\mu \frac{d}{dt} (\nabla \times \mathbf{H}) \rightarrow (2)$$

Maxwell's eqn from ampere's law in point form is given by

$$\nabla \times \mathbf{H} = \mathbf{J} + \frac{d\mathbf{D}}{dt} = \sigma \mathbf{E} + \epsilon \frac{d\mathbf{E}}{dt} \rightarrow (3)$$

Sub eqn (3) in eqn (2)

$$\begin{aligned} \nabla \times \nabla \times \mathbf{E} &= -\mu \frac{d}{dt} \left( \sigma \mathbf{E} + \epsilon \frac{d\mathbf{E}}{dt} \right) \\ &= -\mu \sigma \frac{d\mathbf{E}}{dt} - \mu \epsilon \frac{d^2 \mathbf{E}}{dt^2} \rightarrow (4) \end{aligned}$$

But from vector identity

$$\nabla \times \nabla \times \mathbf{E} = \nabla (\nabla \cdot \mathbf{E}) - \nabla^2 \mathbf{E} \rightarrow (5)$$

$$\nabla \cdot \mathbf{E} = \frac{\nabla \cdot \mathbf{D}}{\epsilon}$$

Since there is no net charge within the conductor, the charge density  $\rho = 0$

$$\nabla \cdot \mathbf{D} = 0$$



eqn (5) becomes

$$\nabla \times \nabla \times E = -\nabla^2 E \rightarrow (6)$$

(27)

Comparing eqn (4) & (6)

$$-\nabla^2 E = -\mu\sigma \frac{dE}{dt} - \mu\epsilon \frac{d^2E}{dt^2}$$

$$\nabla^2 E = \mu\sigma \frac{dE}{dt} + \mu\epsilon \frac{d^2E}{dt^2} \rightarrow (7)$$

$$\nabla^2 E - \mu\sigma \frac{dE}{dt} - \mu\epsilon \frac{d^2E}{dt^2} = 0 \rightarrow (8)$$

This is the wave eqn in terms of electric field  $E$ .

The wave equation in terms of magnetic field  $H$  is obtained in a similar manner as follows.

The Maxwell's eqn from Ampere's law in point form is given by

$$\nabla \times H = J + \frac{dD}{dt} = \sigma E + \epsilon \frac{dE}{dt} \rightarrow (9)$$

Taking curl on both sides

$$\nabla \times \nabla \times H = \sigma \nabla \times E + \epsilon \frac{d}{dt} (\nabla \times E) \rightarrow (10)$$

But from Faraday's law

$$\nabla \times E = -\frac{dB}{dt} = -\mu \frac{dH}{dt} \rightarrow (11)$$

Sub eqn (11) in eqn (10)

$$\mu \frac{d}{dt} (\nabla \times H) = \sigma \nabla \times H + \epsilon \frac{d^2H}{dt^2} \rightarrow (12)$$

Maxwell's equation from Ampere's law in point form is given by

$$\nabla \times H = J + \frac{dD}{dt} = \sigma E + \epsilon \frac{dE}{dt} \rightarrow (3)$$

Sub eqn (3) in eqn (2)

$$\nabla \times \nabla \times E = -\mu \frac{d}{dt} \left( \sigma E + \epsilon \frac{dE}{dt} \right)$$

$$= -\mu \sigma \frac{dE}{dt} - \mu \epsilon \frac{d^2 E}{dt^2} \rightarrow (4)$$

But from vector identity

$$\nabla \times \nabla \times E = \nabla(\nabla \cdot E) - \nabla^2 E \rightarrow (5)$$

$$\nabla \cdot E = \frac{\nabla \cdot D}{\epsilon}$$

Since there is no net charge within the conductor, the charge density  $\rho = 0$

$$\nabla \cdot D = 0 \quad \nabla \cdot E = 0$$

eqn (5) becomes

$$\nabla \times \nabla \times E = -\nabla^2 E \rightarrow (6)$$

Compare eqn (4) & (6)

$$-\nabla^2 E = -\mu \sigma \frac{dE}{dt} - \mu \epsilon \frac{d^2 E}{dt^2}$$

$$\nabla^2 E = \mu \sigma \frac{dE}{dt} + \mu \epsilon \frac{d^2 E}{dt^2} \rightarrow (7)$$

$$\nabla^2 E - \mu \sigma \frac{dE}{dt} - \mu \epsilon \frac{d^2 E}{dt^2} = 0 \rightarrow (8)$$

This is the wave eqn in terms of electric field E.

The Maxwell's eqn from Ampere's law in point form given by

$$\nabla \times H = J + \frac{dD}{dt} = \sigma E + \epsilon \frac{dE}{dt} \rightarrow (9)$$

Taking curl on both sides

$$\nabla \times \nabla \times H = \sigma \nabla \times E + \epsilon \frac{d}{dt} (\nabla \times E) \rightarrow (10)$$

But from Faraday's law

$$\nabla \times E = -\frac{dB}{dt} = -\mu \frac{dH}{dt} \rightarrow (11)$$

Sub eqn (11) in eqn (10)

$$\nabla \times \nabla \times H = -\mu \sigma \frac{dH}{dt} - \mu \epsilon \frac{d^2 H}{dt^2} \rightarrow (12)$$

From vector identity

$$\nabla \times \nabla \times H = \nabla (\nabla \cdot H) - \nabla^2 H \rightarrow (13)$$

$$\nabla \cdot B = \mu \nabla \cdot H = 0$$

eqn (13) becomes

$$\nabla \times \nabla \times H = -\nabla^2 H \rightarrow (14)$$

Comparing eqn (12) & (14)

$$-\nabla^2 H = -\mu \sigma \frac{dH}{dt} - \mu \epsilon \frac{d^2 H}{dt^2}$$

$$\nabla^2 H = \mu \sigma \frac{dH}{dt} + \mu \epsilon \frac{d^2 H}{dt^2}$$

$$\nabla^2 H - \mu \sigma \frac{dH}{dt} - \mu \epsilon \frac{d^2 H}{dt^2} = 0$$

$\therefore$  This is the wave equation in terms of magnetic field

From vector identity

$$\nabla \times \nabla \times \mathbf{H} = \nabla (\nabla \cdot \mathbf{H}) - \nabla^2 \mathbf{H} \rightarrow (13)$$

$$\nabla \cdot \mathbf{B} = \mu \nabla \cdot \mathbf{H} = 0$$

eqn (13) becomes

$$\nabla \times \nabla \times \mathbf{H} = -\nabla^2 \mathbf{H} \rightarrow (14)$$

comparing eqn (12) & (14)

$$-\nabla^2 \mathbf{H} = -\mu \sigma \frac{d\mathbf{H}}{dt} - \mu \epsilon \frac{d^2 \mathbf{H}}{dt^2}$$

$$\nabla^2 \mathbf{H} = \mu \sigma \frac{d\mathbf{H}}{dt} + \mu \epsilon \frac{d^2 \mathbf{H}}{dt^2}$$

$$\nabla^2 \mathbf{H} - \mu \sigma \frac{d\mathbf{H}}{dt} - \mu \epsilon \frac{d^2 \mathbf{H}}{dt^2} = 0$$

This is the wave eqn in terms of magnetic field  $\mathbf{H}$ .

Wave equation for free space

For free space the conductivity of the medium is zero (i.e.  $\sigma = 0$ ) & there is no charge contained in it (i.e.  $\rho = 0$ )

The Maxwell's equation from Faraday's law for free space in point form is

$$\nabla \times \mathbf{E} = -\frac{d\mathbf{B}}{dt} = -\mu \frac{d\mathbf{H}}{dt} \rightarrow (1)$$

Taking curl on both sides

$$\nabla \times (\nabla \times \mathbf{E}) = -\mu \frac{d(\nabla \times \mathbf{H})}{dt} \rightarrow (2)$$

# Wave equation for free space

(3)

$$\sigma = 0 \text{ \& } \rho = 0$$

The Maxwell's eqn from Faraday's law for free space in point form is

$$\nabla \times \mathbf{E} = -\frac{\partial \mathbf{B}}{\partial t} = -\mu \frac{d\mathbf{H}}{dt} \rightarrow (1)$$

Taking curl on both sides

$$\nabla \times \nabla \times \mathbf{E} = -\mu \frac{d}{dt} (\nabla \times \mathbf{H}) \rightarrow (2)$$

The Maxwell's equation from Ampere's law for free space in point form is

$$\nabla \times \mathbf{H} = \frac{d\mathbf{D}}{dt} = \epsilon \frac{d\mathbf{E}}{dt} \rightarrow (3)$$

Sub eqn (3) in eqn (2)

$$\nabla \times \nabla \times \mathbf{E} = -\mu \frac{d}{dt} \left( \epsilon \frac{d\mathbf{E}}{dt} \right)$$

$$= -\mu \epsilon \frac{d^2 \mathbf{E}}{dt^2} \rightarrow (4)$$

From Vector identity

$$\nabla \times \nabla \times \mathbf{E} = \nabla(\nabla \cdot \mathbf{E}) - \nabla^2 \mathbf{E}$$

$$\nabla \cdot \mathbf{E} = \frac{\nabla \cdot \mathbf{D}}{\epsilon} = \frac{\rho}{\epsilon} = 0$$

$$\nabla \times \nabla \times \mathbf{E} = -\nabla^2 \mathbf{E} \rightarrow (5)$$

Compare eqn (4) & eqn (5)

$$\nabla^2 E = \mu \epsilon \frac{\partial^2 E}{\partial t^2}$$

$$\nabla^2 E - \mu \epsilon \frac{\partial^2 E}{\partial t^2} = 0 \rightarrow (6)$$

This is the wave equation for free space interms of electric field.

The wave equation for free space interms of magnetic field  $H$  is obtained in a similar manner follows.

The Maxwell's eqn from Ampere's law for free in point form is given by

$$\nabla \times H = \epsilon \frac{\partial E}{\partial t} \rightarrow (7)$$

Taking curl On both sides

$$\nabla \times \nabla \times H = \epsilon \frac{\partial}{\partial t} (\nabla \times E) \rightarrow (8)$$

The Maxwell's eqn from Faraday's law

$$\nabla \times E = -\mu \frac{\partial H}{\partial t} \rightarrow (9)$$

Sub eqn (9) in eqn (8)

$$\nabla \times \nabla \times H = -\mu \epsilon \frac{\partial}{\partial t} \left( \frac{\partial H}{\partial t} \right)$$

$$\nabla \times \nabla \times H = -\mu \epsilon \frac{\partial^2 H}{\partial t^2} \rightarrow (10)$$

From Vector identity

$$\nabla (\nabla \cdot H) - \nabla^2 H$$

$$\nabla \cdot H = \frac{1}{\mu} \nabla \cdot B = 0$$

(33)

$$\nabla \times \nabla \times H = -\nabla^2 H \rightarrow (11)$$

Compare eqn (10) & (11)

$$-\nabla^2 H = -\mu \epsilon \frac{d^2 H}{dt^2}$$

$$\nabla^2 H = \mu \epsilon \frac{d^2 H}{dt^2}$$

$$\nabla^2 H - \mu \epsilon \frac{d^2 H}{dt^2} = 0 \rightarrow (12)$$

This is the wave eqn for free space in terms of H

For free space  $\mu_r = 1$  &  $\epsilon_r = 1$

$$\nabla^2 H - \mu_0 \epsilon_0 \frac{d^2 H}{dt^2} = 0$$

or

$$\nabla^2 H = \mu_0 \epsilon_0 \frac{d^2 H}{dt^2}$$

**Solution of wave equation**

considering a plane wave propagating in x direction.

The wave eqn for free space is

$$\frac{d^2 E}{dx^2} = \mu_0 \epsilon_0 \frac{d^2 E}{dt^2}$$

The general solution of this differential equation is of the form

$$E = f_1(x - vt) + f_2(x + vt)$$

$$v_0 = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = \text{Velocity of propagation of free space.}$$

$f_1$  &  $f_2$  are any function of  $(x - v_0 t)$  &  $(x + v_0 t)$  respectively.

The solution of wave equation consist of 2 u  
 One travelling in the direction & other travelling in  
 -ve direction. consider the wave travel in +ve dir  
 alone.

$$f_2(x + v_0 t) = 0$$

The general solution of wave equation becomes

$$E = f_1(x - v_0 t) = f(x - v_0 t)$$

$$\nabla \times E = \begin{vmatrix} \vec{a}_x & \vec{a}_y & \vec{a}_z \\ \frac{d}{dx} & \frac{d}{dy} & \frac{d}{dz} \\ E_x & E_y & E_z \end{vmatrix}$$

$$= \vec{a}_z \left( \frac{dE_z}{dy} - \frac{dE_y}{dz} \right) - \vec{a}_y \left( \frac{dE_z}{dx} - \frac{dE_x}{dz} \right) + \vec{a}_x \left( \frac{dE_y}{dx} - \frac{dE_x}{dy} \right)$$

Since the wave travelling in x direction, E & H are independent of y & z.

$$E_x = H_z = 0 \quad \& \quad \frac{dE}{dy} = \frac{dE}{dz} = 0$$

$$\nabla \times E = -\frac{dE_z}{dz} \vec{a}_y + \frac{dE_y}{dx} \vec{a}_z$$



Similarly

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$$\nabla \times H = -\frac{dH_z}{dx} \vec{a}_y + \frac{dH_y}{dx} \vec{a}_z$$

$$\nabla \times H = \epsilon \frac{dE}{dt}$$

$$-\frac{dH_z}{dx} \vec{a}_y + \frac{dH_y}{dx} \vec{a}_z = \epsilon \left[ \frac{dE_y}{dt} \vec{a}_y + \frac{dE_z}{dt} \vec{a}_z \right]$$

Equating  $\vec{a}_y$  &  $\vec{a}_z$  terms

$$-\frac{dH_z}{dx} = \epsilon \frac{dE_y}{dt}$$

$$\frac{dH_y}{dx} = \epsilon \frac{dE_z}{dt}$$

From Maxwell's eqn for free space

$$\nabla \times E = -\mu \frac{dH}{dt}$$

$$\nabla \times E = -\frac{dE_z}{dx} \vec{a}_y + \frac{dE_y}{dx} \vec{a}_z$$

$$-\frac{dE_z}{dx} \vec{a}_y + \frac{dE_y}{dx} \vec{a}_z = -\mu \left[ \frac{dH_y}{dt} \vec{a}_y + \frac{dH_z}{dt} \vec{a}_z \right]$$

Equating  $\vec{a}_y$  &  $\vec{a}_z$  terms

$$\frac{dE_z}{dx} = \mu \frac{dH_y}{dt}$$

$$\frac{dE_y}{dx} = -\mu \frac{dH_z}{dt}$$

Let the solution of this eqn is given by

$$E_y = f(x - v_0 t)$$

(36)

$$\begin{aligned}\frac{dE_y}{dt} &= \frac{df}{d(x - v_0 t)} \cdot \frac{d(x - v_0 t)}{dt} \\ &= f'(x - v_0 t) (-v_0)\end{aligned}$$

$$f'(x - v_0 t) = f'$$

$$\frac{dE_y}{dt} = -v_0 f'$$

$$-\frac{dH_z}{dx} = \epsilon \frac{dE_y}{dt}$$

$$\frac{dH_z}{dx} = -\epsilon (-v_0 f') = \epsilon v_0 f' = \frac{1}{\sqrt{\mu \epsilon}} \epsilon f'$$

$$= \sqrt{\frac{\epsilon}{\mu}} f'$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} \int f' dx$$

$$H_z = \sqrt{\frac{\epsilon}{\mu}} f = \sqrt{\frac{\epsilon}{\mu}} E_y$$

$$\frac{E_y}{H_z} = \sqrt{\frac{\mu}{\epsilon}}$$

$$\frac{E_z}{H_y} = -\sqrt{\frac{\mu}{\epsilon}}$$

If  $E$  is the total electric field

$$E = \sqrt{E_y^2 + E_z^2}$$

&  $H$  is the total magnetic field

$$H = \sqrt{H_y^2 + H_z^2}$$

$$E/H = \sqrt{H/E}$$

Characteristic impedance of medium

$$\eta = \frac{E}{H} = \sqrt{\frac{H}{E}}$$

Free space  $\mu_r = \epsilon_r = 1$

$$\eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = \sqrt{\frac{4\pi \times 10^{-7}}{36\pi \times 10^9}}$$

$$\eta_0 = 120\pi \text{ or } 377 \Omega$$

### Uniform plane wave

If the phase of wave is same for all points

On a plane surface it is called as plane wave.

If the amplitude is also constant in a plane wave it is called as uniform plane wave.

The properties of uniform plane wave are given as follows

- \* At every point in E & H are  $\perp$  to each other & to the direction of travel
- \* The fields are very harmonically with time at the same frequency, everywhere in space
- \* Each field has same direction, magnitude & phase at every point in any plane  $\perp$  to the direction of wave travel

### Poynting's Theorem

- EMW an energy can be transported from transmitter to receiver. The energy stored in an electric & magnetic field is transmitted at a certain rate of energy flow which can be calculated with help of Poynting's Theorem.

### Poynting's Theorem

E is electric field expressed in V/m, H is magnetic field expressed in A/m.

- If we take the product of 2 fields it gives a new quantity which is expressed as W/unit area is called power density. As E & H both are vectors, to get power density we carry out either dot or cross product. The result of dot product is always a scalar & the result of cross product is always a vector.

# Time Harmonic fields

A time harmonic field is one that varies periodically or sinusoidally with time. Sinusoids are easily expressed in phasors.

A phasor is a complex number that contains amplitude & phase of sinusoidal oscillation. As a complex number, a phasor  $Z$  can be represented as

$$Z = x + jy = r \angle \phi \rightarrow (1)$$

or

$$Z = r e^{j\phi} = r (\cos\phi + j \sin\phi) \rightarrow (2)$$

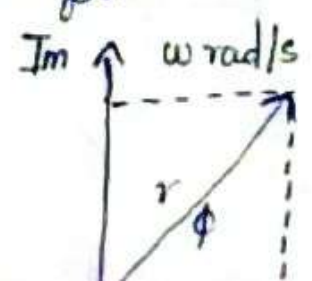
where  $j = \sqrt{-1}$ .  $x$  is real part of  $Z$ ,  $y$  is imaginary part of  $Z$ ,  $r$  is the magnitude of  $Z$ , given by

$$r = |Z| = \sqrt{x^2 + y^2} \rightarrow (3)$$

&  $\phi$  is phase of  $Z$  given by

$$\phi = \tan^{-1} \left( \frac{y}{x} \right) \rightarrow (4)$$

The phasor  $Z$  can be represented in rectangular form or in polar form.



The phasor  $z$  can be represented in rectangular form or in polar form. The 2 forms of representing  $z$ . (50)

Addition & subtraction of phasors are performed in rectangular form.

$\times$  &  $\div$  are performed in polar form

Given complex numbers

$$z = x + jy = r \angle \phi, \quad z_1 = x_1 + jy_1 = r_1 \angle \phi_1$$

The following properties should be noted

Addition:  $z_1 + z_2 = (x_1 + x_2) + j(y_1 + y_2) \rightarrow (5)$

Subtraction:  $z_1 - z_2 = (x_1 - x_2) + j(y_1 - y_2) \rightarrow (6)$

Multiplication:  $z_1 z_2 = r_1 r_2 \angle \phi_1 + \phi_2 \rightarrow (7)$

Division:  $\frac{z_1}{z_2} = \frac{r_1}{r_2} \angle \phi_1 - \phi_2 \rightarrow (8)$

Square root:  $\sqrt{z} = \sqrt{r} \angle \phi/2 \rightarrow (9)$

Complex conjugate:  $z^* = x - jy = r \angle -\phi = r e^{-j\phi} \rightarrow (10)$

To introduce time element let  $\phi = \omega t + \theta \rightarrow (11)$

$\theta$  is a function of time or space coordinates or a constant. The real & imaginary part of  $r e^{j\phi} = r e^{j\theta} e^{j\omega t}$

are given by

R.  $(r e^{j\phi}) = r \cos(\omega t + \theta) \rightarrow (12)$

In general a phasor could be a scalar or vector. If a vector  $A(x, y, z, t)$  is a time harmonic field, the phasor form of  $A$  is  $A_s(x, y, z)$  (5)

$$A = \text{Re}(A_s e^{j\omega t}) \rightarrow (14)$$

For example if  $A = A_0 \cos(\omega t - \beta x) a_y$  we can write  $A$  as

$$A = \text{Re}(A_0 e^{-j\beta x} a_y e^{j\omega t}) \rightarrow (15)$$

Comparing eqn (14) & (15) the phasor form of  $A$  is

$$A_s = A_0 e^{-j\beta x} a_y$$

From eqn (14)

$$\begin{aligned} \frac{dA}{dt} &= \frac{d}{dt} \text{Re}(A_s e^{j\omega t}) \\ &= \text{Re}(j\omega A_s e^{j\omega t}) \rightarrow (16) \end{aligned}$$

Showing that taking the time derivative of the instantaneous quantity is equivalent to  $\times$  its phasor form by  $j\omega$ . That is

$$\frac{dA}{dt} \rightarrow j\omega A_s$$

||<sup>14</sup>

$$\int A dt \rightarrow A_s / j\omega$$

The phasor concept is applied to time varying EMF.

In phasor form Maxwell's for time harmonic EM fields

in a linear, isotropic & homogeneous medium.

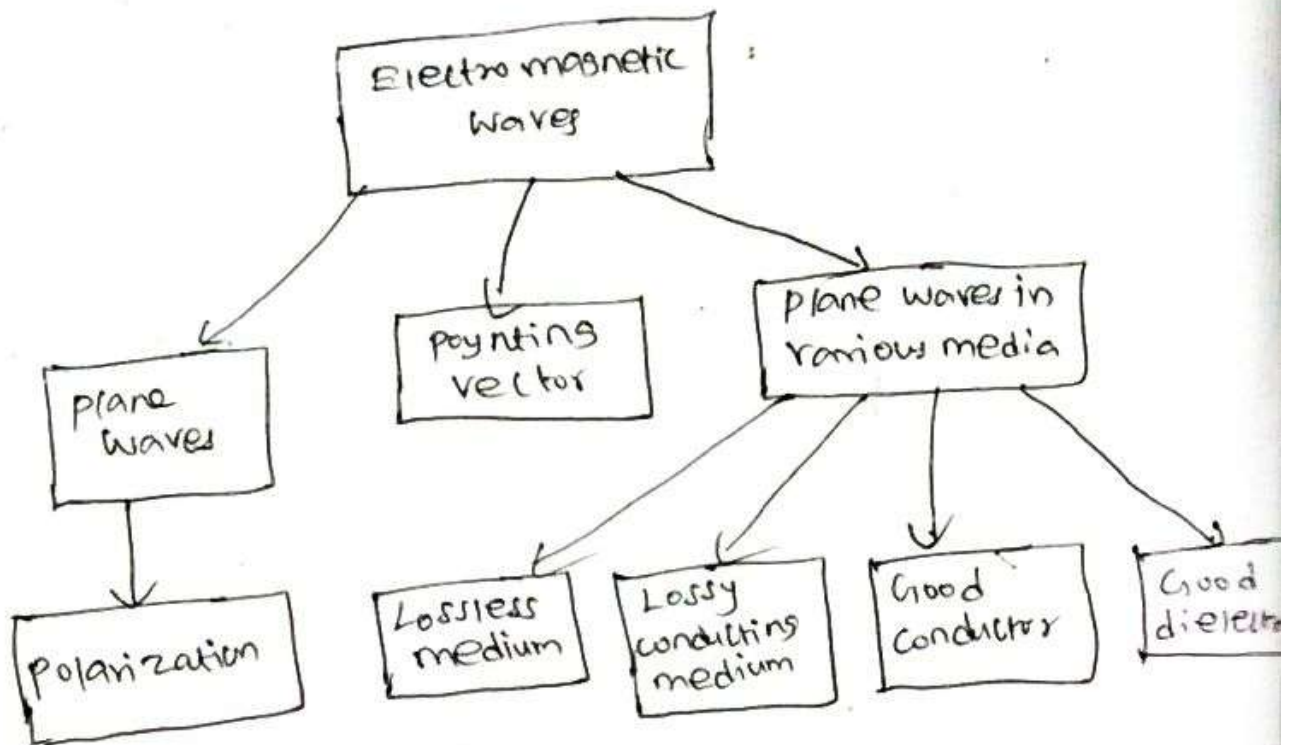


Fig. plane wave

## UNIT-V PLANE ELECTROMAGNETIC WAVES

1. Introduction - (1)
2. Plane waves in lossless media - (2)
3. Plane waves in lossy media (4)
4. Plane waves in lossy (Good dielectric/Conductor) media (9)
5. Electromagnetic Power flow & Poynting Vector (12)
6. Group velocity (16)
7. Normal incidence at a plane conducting boundary (22)
8. Normal incidence at a plane dielectric boundary (25)



# UNIT-IV PLANE ELECTROMAGNETIC WAVES ①

## Introduction:

In free space the source-free wave equation for  $E$  is

$$\nabla^2 E - \frac{1}{c^2} \frac{\partial^2 E}{\partial t^2} = 0 \rightarrow (1)$$

where,  $c = \frac{1}{\sqrt{\mu_0 \epsilon_0}} \approx 3 \times 10^8 \text{ (m/s)} = 300 \text{ (mm/s)} \rightarrow (2)$

It is the velocity of wave propagation in free space.

The solutions of equ (1) represent waves. The study of the behaviour of waves that have a one-dimensional spatial dependence (plane waves) is the main concern of this unit.

The propagation of time-harmonic plane wave fields in an unbounded homogeneous medium. Medium parameters such as intrinsic impedance, attenuation constant and phase constant.

## skin depth:

It is the depth of wave penetration into a good conductor. Electromagnetic waves carry with them electromagnetic power.

## Uniform plane wave:

It is a particular solution of Maxwell's equations with  $E$  assuming the same direction, same magnitude and same phase in infinite planes perpendicular to the direction of propagation.

It does not exist in practice because a source infinite in extent would be required to create it, and practical wave sources are always finite in extent.

The characteristics of uniform plane waves are particularly simple, and their study is of fundamental theoretical, as well as practical, importance

plane waves in lossless media:

The source free wave equation for free space becomes a homogeneous vector Helmholtz's equation.

$$\boxed{\nabla^2 E + k_0^2 E = 0} \rightarrow (3)$$

where  $k_0$  is the free space wavenumber

$$k_0 = \omega \sqrt{\mu_0 \epsilon_0} = \frac{\omega}{c} \text{ (rad/m)} \rightarrow (4)$$

In Cartesian coordinates, Eqn (3) is equivalent to three scalar Helmholtz's equations, one each for the components  $E_x$ ,  $E_y$  and  $E_z$ . Writing it for the component  $E_x$ , we have

$$\left( \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2} + k_0^2 \right) E_x = 0 \rightarrow (5)$$

Consider a uniform plane wave characterized

by wave plane surfaces perpendicular

i.e.)

$$\frac{\partial^2 E_x}{\partial x^2} = 0 \quad \text{and} \quad \frac{\partial^2 E_x}{\partial y^2} = 0 \quad (3)$$

equ (5) simplifies to

$$\frac{d^2 E_x}{dz^2} + k_0^2 E_x = 0 \quad \rightarrow (6)$$

which is an ordinary differential equation because  $E_x$ , a phasor, depends only on  $z$ .

The solution of Eq (6) is readily seen to be

$$\begin{aligned} E_x(z) &= E_x^+(z) + E_x^-(z) \\ &= E_0^+ e^{-jk_0 z} + E_0^- e^{jk_0 z} \quad \rightarrow (7) \end{aligned}$$

where  $E_0^+$  and  $E_0^-$  are arbitrary constants that must be determined by boundary conditions.

---

## Plane waves in various media:-

A media in electromagnetic is characterized by three parameters.

$\epsilon$ ,  $\mu$  and  $\sigma$

### 1. Lossless medium:-

In a lossless medium,  $\epsilon$  and  $\mu$  are real

$\sigma = 0$ , so  $\beta$  is real  $\therefore \gamma^2 = j\omega\mu(\sigma + j\omega\epsilon)$

$$\gamma^2 = j^2 \omega^2 \mu \epsilon = -(j\beta)^2 \Rightarrow \beta = \omega \sqrt{\mu \epsilon}$$

Assume the electric field with

\* only x-component,

\* NO variation along x and y-axis and

\* propagation along z-axis.

$$\text{ie } \frac{\partial \vec{E}}{\partial x} = \frac{\partial \vec{E}}{\partial y} = 0$$

Helmholtz wave equation reduces to

$$\frac{\partial^2}{\partial z^2} E_x + \beta^2 E_x = 0$$

whose solution gives wave in one dimension as follows.

$$E_x = E^+ e^{-j\beta z} + E^- e^{+j\beta z}$$

where,  $E^+$  and  $E^-$  are arbitrary constants.

$$\vec{H} = -\frac{\nabla \times \vec{E}}{j\omega\mu} = \frac{j\nabla \times \vec{E}}{\omega\mu}$$

$$= \frac{j}{\omega\mu} \begin{vmatrix} \hat{x} & \hat{y} & \hat{z} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ E^+ e^{-j\beta z} + E^- e^{+j\beta z} & 0 & 0 \end{vmatrix}$$

$$= \hat{y} \frac{j}{\omega\mu} \left\{ \frac{\partial}{\partial z} (E^+ e^{-j\beta z} + E^- e^{+j\beta z}) \right\}$$

$$\vec{H} = \frac{-j\beta (E^+ e^{-j\beta z}) + (E^- e^{+j\beta z}) j\beta}{\omega\mu} (j) \hat{y}$$

$$= \frac{-j\beta \{ (E^+ e^{-j\beta z}) - (E^- e^{+j\beta z}) \}}{\omega\mu} (j) \hat{y}$$

$$= \frac{\beta \{ (E^+ e^{-j\beta z}) - (E^- e^{+j\beta z}) \}}{\omega\mu} \hat{y}$$

$$\boxed{\vec{H} = \frac{1}{\eta} [E^+ e^{-j\beta z} - E^- e^{+j\beta z}] \hat{y}}$$

where,  $\eta$  is the wave impedance of the plane wave

$$\eta = \frac{\omega\mu}{\beta} = \sqrt{\frac{\mu}{\epsilon}} = \frac{|E_x|}{|H_y|}$$

For free space,

$$\eta_0 = \sqrt{\frac{\mu_0}{\epsilon_0}} = 120\pi = 377\Omega$$

Putting in the time dependence and taking real part, we get

$$E_x(z,t) = E^+ \cos(\omega t - \beta z) + E^- \cos(\omega t + \beta z) \quad (5)$$

For constant phase,

$$\omega t - \beta z = \text{constant} = b \text{ (say)}$$

since phase velocity:

$$v_p = \frac{dz}{dt} = \frac{d}{dt} \left( \frac{\omega t - b}{\beta} \right) = \frac{\omega}{\beta} = \frac{1}{\sqrt{\mu \epsilon}}$$
$$= \frac{1}{\sqrt{\mu_0 \epsilon_0 \epsilon_r}}$$

$$\therefore \beta = \omega \sqrt{\mu \epsilon}$$

For free space,

$$v_p = \frac{1}{\sqrt{\mu_0 \epsilon_0}} = c = 3 \times 10^8 \text{ m/s}$$

which is the speed of light in free space.

This emergence of speed of light from electromagnetic considerations is one of the main contributions from Maxwell's theory.

The magnetic field can be obtained from the source free Maxwell's curl equation.

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

## 2. Lossy conducting medium:-

(7)

If the medium is conductive with a conductivity  $\sigma$ , then the Maxwell's curl equation can be written as

$$\nabla \times \vec{E} = -j\omega \mu \vec{H}$$

$$\nabla \times \vec{H} = j\omega \epsilon \vec{E} + \sigma \vec{E} = (j\omega \epsilon + \sigma) \vec{E} = j\omega \epsilon_{\text{eff}} \vec{E}$$

$$\epsilon_{\text{eff}}(\omega) = \epsilon + \frac{\sigma}{j\omega} = \epsilon - \frac{j\sigma}{\omega} = \epsilon \left(1 - \frac{j\sigma}{\omega \epsilon}\right)$$

The effect of the conductivity has been absorbed in the complex, frequency dependent effective permittivity

$$\nabla^2 \vec{E} + \omega^2 \mu \epsilon_{\text{eff}}(\omega) \vec{E} = \nabla^2 \vec{E} + (j\gamma)^2 \vec{E} = 0$$

We can define a complex propagation constant

$$\gamma = j\omega \sqrt{\mu \epsilon_{\text{eff}}(\omega)} = \alpha + j\beta$$

where,  $\alpha$  is the attenuation constant and  $\beta$  is the phase constant

What is implication of complex wave vector?

- ✓ The wave is exponentially decaying
- ✓ The dispersion relation for a conductor

(usually non-magnetic)  $\mu$

$$\gamma = j\omega \sqrt{\mu \epsilon_{\text{eff}}(\omega)} = j\omega \sqrt{\mu_0 \epsilon_0} \sqrt{\frac{\epsilon_{\text{eff}}(\omega)}{\epsilon_0}}$$

$$= j\omega \sqrt{\mu_0 \epsilon_0} n_{\text{eff}}(\omega) = j \frac{\omega}{c} n_{\text{eff}}(\omega)$$

where,  $\eta_{eff}$  is the complex refractive index.

1-D wave equation for general lossy medium becomes

$$\frac{\partial^2 E_x}{\partial z^2} - \gamma^2 E_x = 0$$

whose solution is 1-D plane waves as follows.

$$E_x(z) = E^+ e^{-\gamma z} + E^- e^{+\gamma z}$$

$$= E^+ e^{-\alpha z} e^{-j\beta z} + E^- e^{+\alpha z} e^{j\beta z}$$

Putting the time dependence and taking real part we get

$$E_x(z,t) = E^+ e^{-\alpha z} \cos(\omega t - \beta z) + E^- e^{+\alpha z} \cos(\omega t + \beta z)$$

The magnetic field can be found out from Maxwell's equations as in the previous section.

$$H_y(z) = \frac{1}{\eta_{eff}} [E^+ e^{-\gamma z} - E^- e^{+\gamma z}]$$

where useful expression for intrinsic impedance

$$\eta_{eff} = \frac{j\omega\mu_0}{\gamma} = \frac{j\omega\mu_0}{j\omega\sqrt{\mu_0\epsilon_{eff}(\omega)}}$$

$$\eta_{eff} = \sqrt{\frac{\mu_0}{\epsilon_{eff}(\omega)}}$$



the electric field and magnetic field are no longer in phase as  $\epsilon_{eff}$  is complex (9)

Poynting vector (or) power flow for this wave inside the lossy conducting medium is

$$\vec{S} = \vec{E} \times \vec{H}^*$$

$$= E^+ e^{-\alpha z} e^{-j\beta z} \hat{x} \times \left( \frac{E^+ e^{-\alpha z} e^{-j\beta z}}{\eta_{eff}} \right)^* \hat{y}$$

$$= \frac{|E^+|^2 e^{-2\alpha z} e^{-j\beta z} e^{+j\beta z}}{\eta_{eff}^*} \hat{z}$$

$$\vec{S} = \frac{|E^+|^2}{\eta_{eff}^*} e^{-2\alpha z} \hat{z}$$

It is decaying in terms of square of an exponential function.

### 3. Good dielectric / Conductor:-

✓ Note that  $\sigma/\omega\epsilon$  is defined as loss tangent of a medium

✓ A medium with  $\sigma/\omega\epsilon < 0.01$  is said to be a good insulator

✓ whereas a medium with  $\sigma/\omega\epsilon > 100$  is said to be a good conductor.

For good dielectric,

$$\sigma \ll \omega \epsilon \therefore \gamma = j\omega \sqrt{\mu \epsilon} \left( \sqrt{1 - \frac{j\sigma}{\omega \epsilon}} \right)$$

It can be approximated using Taylor's series expansion obtain  $\alpha$  and  $\beta$  as follows

$$\alpha = \frac{\sigma}{2} \sqrt{\frac{\mu}{\epsilon}}$$

$$\beta = \omega \sqrt{\mu \epsilon}$$

For a good conductor,

$$\sigma \gg \omega \epsilon$$

$$\therefore \gamma \cong (1+j) \sqrt{\frac{\omega \mu \sigma}{2}}$$

$$\Rightarrow \alpha = \beta = \sqrt{\frac{\omega \mu \sigma}{2}}$$

Skin effect:

1) The field do attenuate as they travel in a good dielectric medium

2)  $\alpha$  in a good dielectric is very small in comparison to that of a good conductor

3) As the amplitude of the wave varies with  $e^{-\alpha z}$

4) The wave amplitude reduces to 1/e or 37% times over a distance of

$$\delta = \frac{1}{\alpha} = \frac{1}{\beta} = \sqrt{\frac{2}{\omega \mu \sigma}} = \sqrt{\frac{2}{2\pi f \mu \sigma}} \quad (11)$$

$$\delta = \frac{1}{\sqrt{\pi f \mu \sigma}}$$

which is also known as skin depth.

This means that in a good conductor

- a) higher the frequency, lower is the skin depth
- b) higher is the conductivity, lower is the skin depth &
- c) higher is the permeability, lower is the skin depth.

Let us assume an EM wave which has x-component and travelling along the z-axis.

Then, it can be expressed as

$$E_x(z, t) = E_0 e^{-\alpha z} e^{-j(\beta z - \omega t)}$$

taking the real part we have,

$$E_x(z, t) = E_0 e^{-\alpha z} \cos(\omega t - \beta z)$$

Substituting the values of  $\alpha$  and  $\beta$  for good conductors, we have,

$$E_x(z, t) = E_0 e^{-\sqrt{\pi f \mu \sigma} z} \cos(\omega t - \sqrt{\pi f \mu \sigma} z)$$

Now using the point form of ohm's law for conductors, we can write.

$$J_x = \sigma E_x(z, t) = \sigma E_0 e^{-\sqrt{\pi f \mu \sigma} z} \cos(\omega t - \sqrt{\pi f \mu \sigma} z)$$

What is the phase velocity and wavelength inside a good conductor?

$$v_p = \frac{\omega}{\beta} = \omega \delta, \quad \lambda = \frac{2\pi}{\beta} = 2\pi \delta.$$

### ELECTROMAGNETIC POWER FLOW & POYNTING VECTOR

The rate of energy flow per unit area in a plane wave is described by a vector termed as Poynting vector.

which is basically curl of electric field intensity vector and magnetic field intensity vector.

$$\vec{S} = \vec{E} \times \vec{H}^*$$

The magnitude of Poynting vector is the power flow per unit area and it points along the direction of wave propagation vector.

The average power per unit area is often called the intensity of EM wave and it is given by

$$\vec{S}_{avg} = \frac{1}{2} \text{Re} (\vec{E} \times \vec{H}^*)$$

Let us try to derive the point form of Poynting theorem from two Maxwell's curl equations.

$$\nabla \times \vec{E} = -\mu \frac{\partial \vec{H}}{\partial t}$$

(13)

$$\nabla \times \vec{H} = \epsilon \frac{\partial \vec{E}}{\partial t} + \vec{J}$$

From vector analysis,

$$\nabla \cdot (\vec{E} \times \vec{H}) = \vec{H} \cdot (\nabla \times \vec{E}) - \vec{E} \cdot (\nabla \times \vec{H})$$

$$= \vec{H} \cdot \left(-\mu \frac{\partial \vec{H}}{\partial t}\right) - \vec{E} \cdot \left(\epsilon \frac{\partial \vec{E}}{\partial t} + \vec{J}\right)$$

We can further simplify.

$$\vec{A} \cdot \frac{\partial \vec{A}}{\partial t} = \frac{1}{2} \frac{\partial}{\partial t} (\vec{A} \cdot \vec{A})$$

$$\therefore \nabla \cdot (\vec{E} \times \vec{H}) = -\frac{\mu}{2} \frac{\partial}{\partial t} (\vec{H} \cdot \vec{H}) - \frac{\epsilon}{2} \frac{\partial}{\partial t} (\vec{E} \cdot \vec{E}) - \vec{E} \cdot \vec{J}$$

Basically a point relation.

It should be valid at every point in space at every instant of time.

The power is given by the integral of this relation of Poynting vector over a volume as follows:

$$\int_V \nabla \cdot (\vec{E} \times \vec{H}) dV = \oint_S (\vec{E} \times \vec{H}) \cdot d\vec{S}$$

$$= \oint_S \vec{S} \cdot d\vec{S}$$

$$\oint_S \vec{S} \cdot d\vec{S} = -\frac{\mu}{2} \int_V \frac{\partial}{\partial t} (\vec{H} \cdot \vec{H}) dV - \frac{\epsilon}{2} \int_V \frac{\partial}{\partial t} (\vec{E} \cdot \vec{E}) dV - \int_V \vec{E} \cdot \vec{J} dV$$

we can interchange the volume integral and partial derivative with respect to time

$$\oint_S \vec{s} \cdot d\vec{s} = \frac{\partial}{\partial t} \int_V \frac{1}{2} \mu H^2 dV - \frac{\partial}{\partial t} \int_V \frac{1}{2} \epsilon E^2 dV - \int_V \sigma E^2 dV$$

This is the integral form of Poynting vector and power flow in EM fields

Poynting theorem states that "the power coming out of the closed volume is equal to the total decrease in EM energy per unit time" i.e. power loss from the volume which constitutes of

✓ rate of decrease in magnetic energy store in the volume

✓ rate of decrease in electric energy store in the volume

✓ ohmic power loss (energy converted into heat energy per unit time) in the volume.

Now going back to the last four points of plane wave:

k) the direction of propagation is in the same direction as of Poynting vector

Note that the direction of Poynting vector is also in the z-direction same as that of wave vector.

The average value of the Poynting vector

$$\vec{S}_{avg} = \frac{1}{2} \operatorname{Re}(\vec{E} \times \vec{H}^*)$$

$$= \frac{1}{2} \operatorname{Re}\left(\frac{|\vec{E}_0|^2 \hat{z}}{\eta_0}\right) = \frac{|\vec{E}_0|^2 \hat{z}}{2\eta_0}$$

Stored Electric energy:  $W_e = \frac{1}{2} \epsilon_0 E^2$

Stored magnetic energy:

$$W_m = \frac{1}{2} \mu_0 H^2 = \frac{1}{2} \mu_0 \frac{E^2}{\eta_0^2} = \frac{1}{2} \mu_0 \frac{\epsilon_0}{\mu_0} E^2$$

$$= \frac{1}{2} \epsilon_0 E^2 = W_e$$

Group velocity:

The relation between phase velocity ( $u$ ) and the phase constant  $\beta$  is

$$u_p = \frac{\omega}{\beta} \text{ (m/s)} \rightarrow \textcircled{1}$$

For plane wave in a lossless medium  $\beta = \omega \sqrt{\mu \epsilon}$  is a linear function of  $\omega$ .

The phase velocity  $u_p = \frac{1}{\sqrt{\mu \epsilon}}$  is a constant

(15)

\* The instantaneous value of the Poynting vector is given by

$$\vec{E}^2 / \eta_0 \quad (\text{on } H^2 \eta_0)$$

\* The average value of the Poynting vector is given by  $E^2 / 2\eta_0$  (on  $H^2 \eta_0 / 2$ ).

\* The stored electric energy is equal to the stored magnetic energy at any instant.

Let us assume a plane wave travelling in the +z direction in free space, then

$$\vec{E} = \vec{E}_0 e^{-j\beta z} = \vec{E}_0 e^{-j\beta z}$$

$$\vec{H} = \frac{\hat{z} \times \vec{E}_0}{\eta_0} e^{-j\beta z}$$

The instantaneous value of the Poynting

vector,

$$(iv) \vec{S} = \vec{E} \times \vec{H}^* = (\vec{E}_0 e^{-j\beta z}) \times \left( \frac{\hat{z} \times \vec{E}_0}{\eta_0} e^{j\beta z} \right)$$

$$= \frac{1}{\eta_0} (\vec{E}_0) \times (\hat{z} \times \vec{E}_0)$$

$$\vec{S} = \frac{\hat{z} (\vec{E}_0 \cdot \vec{E}_0) - \vec{E}_0 (\vec{E}_0 \cdot \hat{z})}{\eta_0} = \frac{\hat{z} (\vec{E}_0 \cdot \vec{E}_0)}{\eta_0} = \frac{|\vec{E}_0|^2 \hat{z}}{\eta_0}$$



that is independent of frequency. (17)

The phenomenon of signal distortion in the signal caused by a dependence of the phase velocity on frequency is called dispersion.

A group velocity is the velocity of propagation of the wave-packet envelope (of a group of frequencies).

Consider the simplest case of a wave packet that consists of two travelling waves having equal amplitude and slightly different angular frequencies  $\omega_0 + \Delta\omega$  and  $\omega_0 - \Delta\omega$  ( $\Delta\omega \ll \omega_0$ )

Let the phase constants corresponding to the two frequencies be  $\beta_0 + \Delta\beta$  and  $\beta_0 - \Delta\beta$  we have

$$E(z,t) = E_0 \cos[(\omega_0 + \Delta\omega)t - (\beta_0 + \Delta\beta)z] +$$

$$E_0 \cos[(\omega_0 - \Delta\omega)t - (\beta_0 - \Delta\beta)z] \rightarrow \text{①}$$

$$E(z,t) = 2E_0 \cos(\omega_0 t - \beta_0 z) \cos(\Delta\omega t - \Delta\beta z)$$

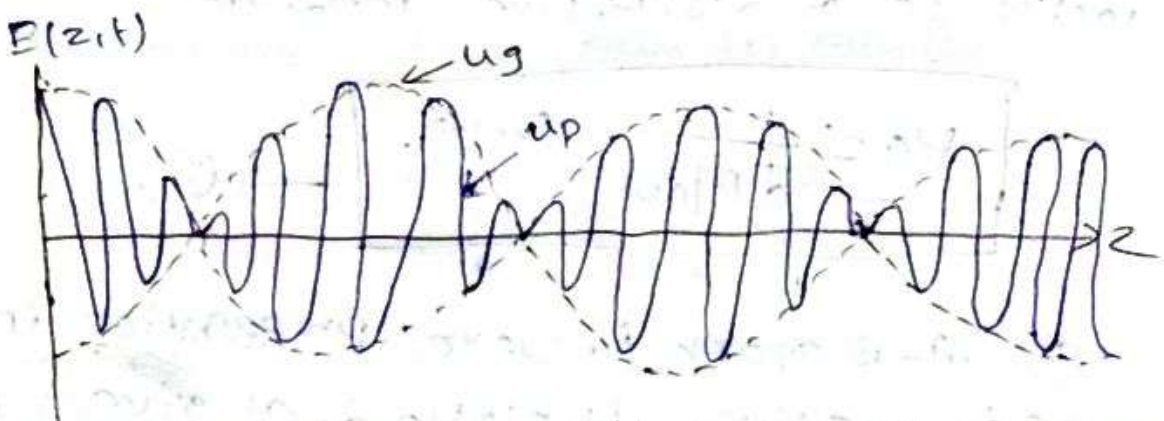


Fig. Sum of two harmonic traveling waves of equal amplitude and slightly different frequencies.

Since,  $\Delta\omega \ll \omega_0$ , the eqn (2) represents a rapidly oscillating wave having an angular frequency  $\omega_0$  and an amplitude that varies slowly with an angular frequency  $\Delta\omega$ .

The wave inside the envelope propagates with a phase velocity found by setting  $\omega_0 t - \beta_0 z = \text{constant}$ .

$$v_p = \frac{dz}{dt} = \frac{\omega_0}{\beta_0}$$

The group velocity ( $v_g$ ) can be determined by setting the argument of the first cosine factor in eqn (2) equal to a constant

$$t \Delta\omega - z \Delta\beta = \text{constant}$$

From which we obtain

$$v_g = \frac{dz}{dt} = \frac{\Delta\omega}{\Delta\beta} = \frac{1}{\Delta\beta/\Delta\omega}$$

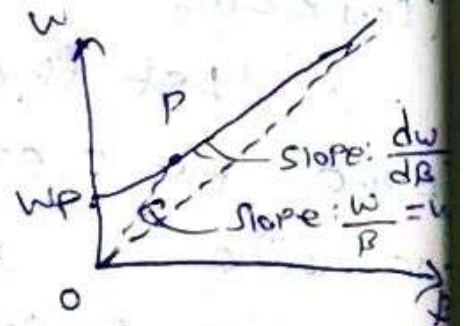


Fig.  $\omega$ - $\beta$  graph for ionized gas

In the limit that  $\Delta\omega \rightarrow 0$ ,

We have the formula for computing the group velocity in a dispersive medium.

$$v_g = \frac{1}{d\beta/d\omega} \text{ (m/s)} \rightarrow (3)$$

In  $\omega$ - $\beta$  graph for wave propagation in a ionized medium is plotted as given by

$$B = \omega \sqrt{\mu \epsilon_0} \sqrt{1 - \left(\frac{f_p}{f}\right)^2} \quad (19)$$

$$= \frac{\omega}{c} \sqrt{1 - \left(\frac{\omega_p}{\omega}\right)^2} \rightarrow (4)$$

At  $\omega = \omega_p$  (the cutoff angular frequency),  
 $B = 0$ . For  $\omega > \omega_p$ , wave propagation is  
 possible, and

$$u_p = \frac{\omega}{B} = \frac{c}{\sqrt{1 - \left(\frac{\omega_p}{\omega}\right)^2}} \rightarrow (5)$$

Sub eqn (4) in eqn (3), we have

$$u_g = c \sqrt{1 - \left(\frac{\omega_p}{\omega}\right)^2} \rightarrow (6)$$

$u_p \geq c$  and  $u_g \leq c$  then  $u_p u_g = c^2$ .

A general relation between the group  
 and phase velocities may be obtained by  
 combining eqn (1) & (3). From eqn (1) we have

$$\frac{dB}{d\omega} = \frac{d}{d\omega} \left( \frac{\omega}{u_p} \right) = \frac{1}{u_p} - \frac{\omega}{u_p^2} \frac{du_p}{d\omega}$$

Sub of the above eqn in eqn (3)

$$u_g = \frac{u_p}{1 - \frac{\omega}{u_p} \cdot \frac{du_p}{d\omega}} \rightarrow (7)$$

From eqn (7) we see three possible cases.

a) No dispersion:-

$\frac{d u_p}{d \omega} = 0$  ( $u_p$  independent of  $\omega$ ,  $\beta$  a linear function of  $\omega$ )

$$u_g = u_p$$

b) Normal dispersion:-

$\frac{d u_p}{d \omega} < 0$  ( $u_p$  decreasing with  $\omega$ )

$$u_g < u_p$$

c) Anomalous dispersion:-

$\frac{d u_p}{d \omega} > 0$  ( $u_p$  increasing with  $\omega$ )

$$u_g > u_p$$

Prblm: A narrow band signal propagates in a lossy dielectric medium which has a loss tangent of 0.2 at 550 (kHz), the carrier frequency of the signal. The dielectric constant of the medium is 2.5.

a) Determine  $q$  and  $R$ . b) Determine  $u_p$  and  $u_g$ .

Is the medium dispersive?

Solution:

a) since loss tangent  $\epsilon''/\epsilon' = 0.2$  and  $\epsilon'' = \frac{\sigma}{\omega \epsilon'}$

$$\begin{aligned} \epsilon'' &= 0.2 \epsilon' = 0.2 \times 2.5 \epsilon_0 \\ &= 4.42 \times 10^{-12} \text{ (F/m)} \end{aligned}$$

$$\alpha = \frac{\omega \epsilon''}{2} \sqrt{\frac{\mu}{\epsilon'}} = \pi (550 \times 10^3) \times 14.42 \times 10^{-12} \times \frac{377}{\sqrt{2.5}}$$

$$\alpha = 1.82 \times 10^{-3} \text{ (NP/m)}$$

$$\beta = \omega \sqrt{\mu \epsilon'} \left[ 1 + \frac{1}{8} \left( \frac{\epsilon''}{\epsilon'} \right)^2 \right]$$

$$= 2\pi (550 \times 10^3) \frac{\sqrt{2.5}}{3 \times 10^8} \left[ 1 + \frac{1}{8} (0.2)^2 \right]$$

$$\beta = 0.0182 \times 1.005 = 0.0183 \text{ (rad/m)}$$

b) Phase velocity

$$v_p = \frac{\omega}{\beta} = \frac{1}{\sqrt{\mu \epsilon'} \left[ 1 + \frac{1}{8} \left( \frac{\epsilon''}{\epsilon'} \right)^2 \right]} \approx \frac{1}{\sqrt{\mu \epsilon'}} \left[ 1 - \frac{1}{8} \left( \frac{\epsilon''}{\epsilon'} \right)^2 \right]$$

$$= \frac{3 \times 10^8}{\sqrt{2.5}} \left[ 1 - \frac{1}{8} (0.2)^2 \right]$$

$$v_p = 1.888 \times 10^8 \text{ (m/s)}$$

c) Group velocity

$$\frac{d\beta}{d\omega} = \sqrt{\mu \epsilon'} \left[ 1 + \frac{1}{8} \left( \frac{\epsilon''}{\epsilon'} \right)^2 \right]$$

$$v_g = \frac{1}{(d\beta/d\omega)} \approx \frac{1}{\sqrt{\mu \epsilon'}} \approx v_p$$

Thus a low-loss dielectric is nearly nondispersive. Here we have assumed  $\epsilon''$  to be independent of frequency.

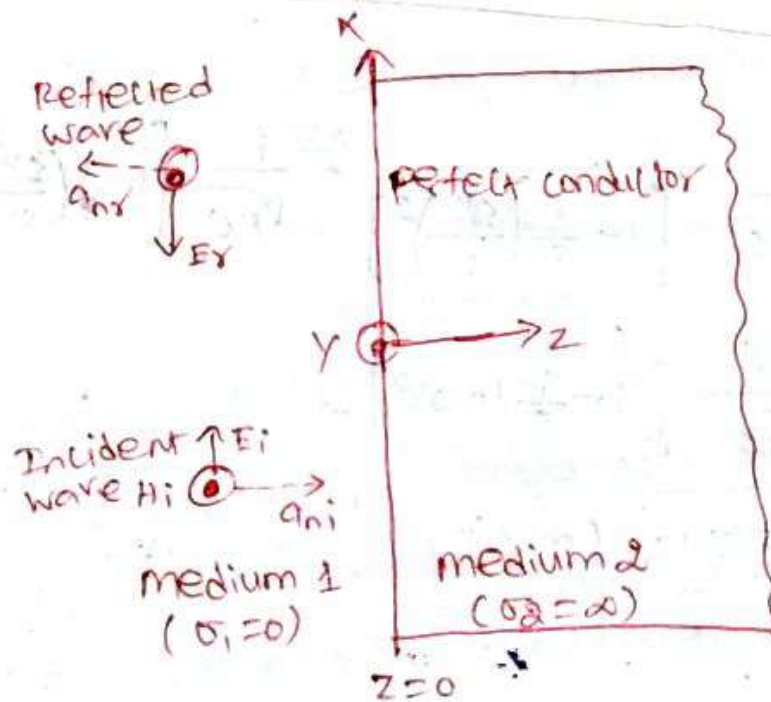
For a high-loss dielectric,  $\epsilon''$  will be a function of  $\omega$  and may have a magnitude comparable to  $\epsilon'$ .

+

Normal Incidence at a Plane Conducting Boundary

k) The incident wave travels in a lossless medium

\*) The boundary is an interface with a perfect conductor.



Incident wave (inside medium 1)

$$\vec{E}_i(z) = \hat{a}_x E_{i0} e^{-j\beta_1 z}$$

$$\vec{H}_i(z) = \hat{a}_y \frac{E_{i0}}{\eta_1} e^{-j\beta_1 z}$$

where,  $E_{i0}$  is the magnitude of  $\vec{E}_i$   
 $\beta_1$  is the phase constant  
 $\eta_1$  is the intrinsic impedance of medium 1

(23)

Inside wave in medium 2, both electric and magnetic fields vanish,  $\vec{E}_2 = 0$ ,  $\vec{H}_2 = 0$

No wave is transmitted across the boundary into the  $z > 0$

Reflected wave (inside medium 1)

$$\vec{E}_y(z) = a_x \hat{E}_{y0} e^{+j\beta_1 z}$$

$$\vec{H}_x(z) = \frac{1}{\eta_1} a_x \hat{a}_x \times \vec{E}_y(z)$$

$$= \frac{1}{\eta_1} (-a_z) \times \vec{E}_y(z)$$

$$\vec{H}_x(z) = -a_y \frac{1}{\eta_1} E_{y0} e^{+j\beta_1 z}$$

Total wave in medium 1

$$\vec{E}_1(z) = \vec{E}_i(z) + \vec{E}_r(z)$$

$$= a_x \hat{a}_x (E_{i0} e^{-j\beta_1 z} + E_{r0} e^{+j\beta_1 z})$$

Continuity of tangential component of the E field at the boundary  $z=0$

$$\vec{E}_1(0) = a_x \hat{a}_x (E_{i0} + E_{r0}) = E_2(0) = 0$$

$$\Rightarrow E_{r0} = -E_{i0}$$

$$\therefore \vec{E}_1(z) = a_x \hat{a}_x E_{i0} (e^{-j\beta_1 z} - e^{+j\beta_1 z}) = -a_x \hat{a}_x j 2 E_{i0} \sin \beta_1 z$$

$$\therefore \vec{H}_1(z) = \vec{H}_i(z) + \vec{H}_r(z)$$

$$= a_y \hat{a}_y 2 \frac{E_{i0}}{\eta_1} \cos \beta_1 z$$

$$\therefore \vec{E}_1(z) = \hat{a}_x E_{10} (e^{-j\beta_1 z} - e^{+j\beta_1 z})$$

$$= -\hat{a}_x j 2 E_{10} \sin \beta_1 z$$

$$\therefore \vec{H}_1(z) = \vec{H}_1^+(z) + \vec{H}_1^-(z)$$

$$= \hat{a}_y 2 \frac{E_{10}}{\eta_1} \cos \beta_1 z$$

The space-time behavior of the total field in medium 1

$$\vec{E}_1(z, t) = \text{Re} [\vec{E}_1(z) e^{j\omega t}] = \hat{a}_x 2 E_{10} \sin \beta_1 z \sin \omega t$$

$$\vec{H}_1(z, t) = \text{Re} [\vec{H}_1(z) e^{j\omega t}] = \hat{a}_y 2 \frac{E_{10}}{\eta_1} \cos \beta_1 z \cos \omega t$$

zeros of  $\vec{E}_1(z, t)$  }  
 maxima of  $\vec{H}_1(z, t)$  } occur at  $\beta_1 z = -n\pi$  (or)

$$z = -n \frac{\lambda}{2}, \quad n = 0, 1, 2, \dots$$

maxima of  $\vec{E}_1(z, t)$  }  
 zeros of  $\vec{H}_1(z, t)$  } occur at  $\beta_1 z = -(2n+1) \frac{\pi}{2}$

$$\text{(or)} \quad z = -(2n+1) \frac{\lambda}{4}, \quad n = 0, 1, 2, \dots$$

$$\vec{E}_1(z, t) = \text{Re} [\vec{E}_1(z) e^{j\omega t}]$$

$$= \hat{a}_x 2 E_{10} \sin \beta_1 z \sin \omega t$$

$$\vec{H}_1(z, t) = \text{Re} [\vec{H}_1(z) e^{j\omega t}]$$

$$= \hat{a}_y 2 \frac{E_{10}}{\eta_1} \cos \beta_1 z \cos \omega t$$

The total wave in medium 1 is not a travel wave.  
 Standing wave.



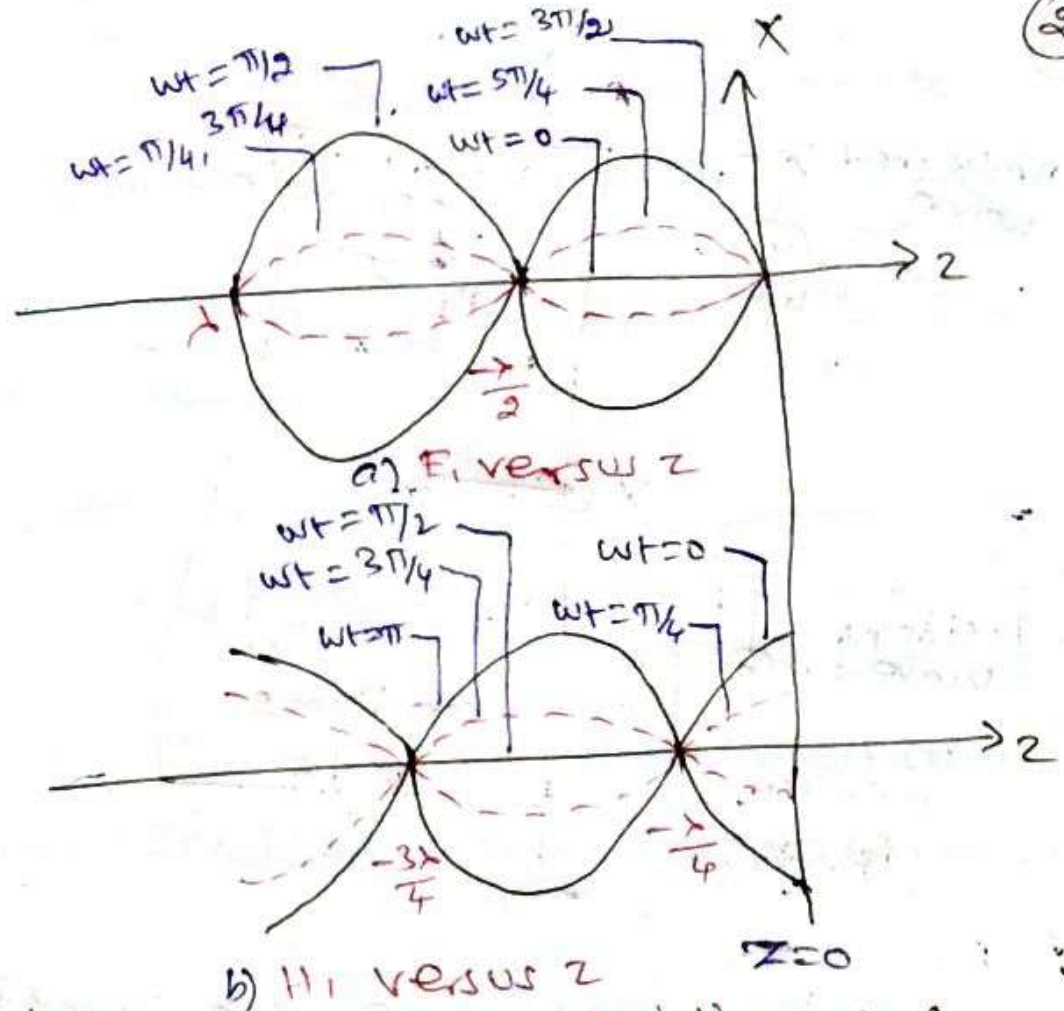


Fig Standing waves of  $E_1 = a_x E_1$  and  $H_1 = a_y H_1$  for several values of  $\omega t$ . Note following three points

- (i)  $E_1$  vanishes on the conducting boundary
- (ii)  $H_1$  a maximum on the conducting boundary
- (iii) the standing waves of  $E_1$  and  $H_1$  are in time quadrature (90° phase difference)

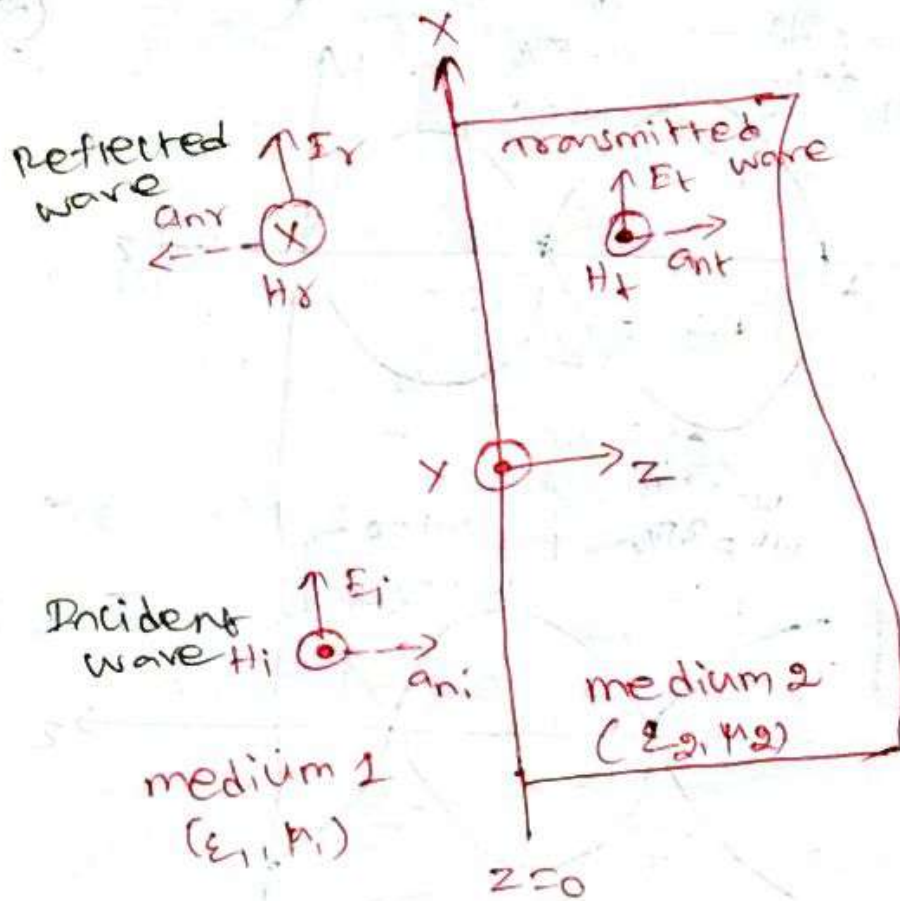
Normal incidence at a plane dielectric boundary

$$\sigma_1 = \sigma_2 = 0, \quad \epsilon_1 \neq \epsilon_2$$

Incident wave (inside medium 1)

$$\vec{E}_i(z) = a_x \hat{x} E_{i0} e^{-j\beta_1 z}$$

$$\vec{H}_i(z) = a_y \hat{y} \frac{E_{i0}}{\eta_1} e^{-j\beta_1 z}$$



Reflected wave (inside medium 1)

$$\vec{E}_r(z) = \hat{a}_x E_{r0} e^{-j\beta_1 z}$$

$$\vec{H}_r(z) = (-\hat{a}_z) \times \frac{1}{\eta_1} \vec{E}_r(z) = -\hat{a}_y \frac{E_{r0}}{\eta_1} e^{-j\beta_1 z}$$

Transmitted wave (inside medium 2)

$$\vec{E}_t(z) = \hat{a}_x E_{t0} e^{-j\beta_2 z}$$

$$\vec{H}_t(z) = \hat{a}_z \times \frac{1}{\eta_2} \vec{E}_t(z) = \hat{a}_y \frac{E_{t0}}{\eta_2} e^{-j\beta_2 z}$$

The tangential components (the x-components) of the electric and magnetic field intensities must be continuous (at interface  $z=0$ )

$$E_{1tan} = E_{2tan}, \quad H_{1tan} = H_{2tan}$$

$$\vec{E}_T(0) + \vec{E}_R(0) = \vec{E}_T(0) \Rightarrow E_{T0} + E_{R0} = E_{T0} \quad (27)$$

$$\vec{H}_I(0) + \vec{H}_R(0) = \vec{H}_T(0) \Rightarrow \frac{1}{\eta_1} (E_{T0} - E_{R0}) = \frac{E_{T0}}{\eta_2}$$

$$\therefore E_{R0} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} E_{T0} \quad E_{T0} = \frac{2\eta_2}{\eta_2 + \eta_1} E_{I0}$$

Reflection coefficient (+ or -)  $\leq 1$

$$\eta_1 = \eta_2 \therefore \Gamma = 0$$

$$\Gamma = \frac{E_{R0}}{E_{I0}} = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1}$$

$\eta_2 = 0$  (Short)  $\Gamma = -1$   $E/H, E \neq 0$  perfect conductor!!

$\eta_2 = \infty$  (open)  $\Gamma = 1$   $H(I) = 0$  NO current!!

$$E_{T0} = \frac{2\eta_2}{\eta_2 + \eta_1} E_{I0}$$

$$\tau = \frac{E_{T0}}{E_{I0}} = \frac{2\eta_2}{\eta_2 + \eta_1}$$

$$1 + \Gamma = \tau$$

Transmission coefficient (+ always)

If medium 2  $\rightarrow$  perfect conductor  $\eta_2 = 0$

$$\Gamma = -1, \tau = 0 \Rightarrow E_{R0} = -E_{I0}, E_{T0} = 0$$

$\rightarrow$  totally reflected. Standing wave produced in medium 1.

If medium 2 is not a perfect conductor, partial reflection will result.

$$\begin{aligned} \vec{E}_T(z) &= \vec{E}_I(z) + \vec{E}_R(z) = \hat{a}_x E_{I0} (e^{-j\beta_1 z} + \Gamma e^{j\beta_1 z}) \\ &= \hat{a}_x E_{I0} [(1 + \Gamma) e^{-j\beta_1 z} + \Gamma (e^{j\beta_1 z} - e^{-j\beta_1 z})] \end{aligned} \quad (z < 0)$$

$$= \hat{a}_x E_{i0} \left[ \underbrace{(1 + \Gamma) e^{-j\beta_1 z}}_{\text{Traveling}} + \underbrace{\Gamma (j2 \sin \beta_1 z)}_{\text{Standing}} \right]$$

$$\Rightarrow \vec{E}_1(z) = \hat{a}_x E_{i0} e^{-j\beta_1 z} (1 + \Gamma e^{j2\beta_1 z}) \quad (z < 0)$$

$$|\vec{E}_1(z)| = E_{i0} \left\{ (1 + \Gamma e^{j2\beta_1 z})(1 + \Gamma e^{-j2\beta_1 z}) \right\}$$

$$|\vec{E}_1(z)| = E_{i0} (1 + \Gamma^2 + 2\Gamma \cos 2\beta_1 z)^{1/2}$$

For dissipationless media  $\eta_1, \eta_2, \Gamma, \Pi$  are real

However,  $\Gamma$  can be positive (or) negative

$$(i) \Gamma = \frac{\eta_2 - \eta_1}{\eta_2 + \eta_1} > 0 \quad (\eta_2 > \eta_1)$$

✓ maximum value of  $|\vec{E}_1(z)|$  is  $E_{i0}(1 + \Gamma)$

which occurs when  $2\beta_1 z_{\max} = -2n\pi$  ( $n=0, 1, 2, \dots$ )

$$\therefore z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2}, \quad n=0, 1, 2, \dots$$

✓ minimum value of  $|\vec{E}_1(z)|$  is  $E_{i0}(1 - \Gamma)$

which occurs when  $2\beta_1 z_{\min} = -(2n+1)\pi$  ( $n=0, 1, 2, \dots$ )

$$\therefore z_{\min} = -\frac{(2n+1)\pi}{2\beta_1} = -\frac{(2n+1)\lambda_1}{4}, \quad n=0, 1, 2, \dots$$

$$(ii) \Gamma < 0 \quad (\eta_2 < \eta_1)$$

✓ maximum value of  $|\vec{E}_1(z)|$  is  $E_{i0}(1 - \Gamma)$

$$\text{at } z_{\min} = -\frac{(2n+1)\pi}{2\beta_1} = -\frac{(2n+1)\lambda_1}{4}, \quad n=0, 1, 2, \dots$$

✓ minimum value of  $|\vec{E}_1(z)|$  is  $E_{i0}(1-\Gamma)$ . (2)

$$\text{at } z_{\max} = -\frac{n\pi}{\beta_1} = -\frac{n\lambda_1}{2}, \quad n=0,1,2,\dots$$

$$S = \frac{|E|_{\max}}{|E|_{\min}} = \frac{1+|\Gamma|}{1-|\Gamma|}$$

Standing wave ratio (SWR)

$$|\Gamma| = \frac{S-1}{S+1}$$

$$(-1 \leq \Gamma \leq 1, 1 \leq S \leq \infty)$$

if  $\Gamma = 0, S = 1$ , no reflection, full power transmission

if  $\Gamma = 1, S = \infty$ , total reflection, no power transmission